# IE0

## DESCRIPCIÓN

## SISTEMA Y PROCEDIMIENTO DE GESTIÓN DE CARGA DE BATERÍAS DE UN VEHÍCULO ELÉCTRICO

## **OBJETO DE LA INVENCIÓN**

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La presente invención describe un sistema y un procedimiento para gestionar de una manera óptima la carga y descarga de un conjunto de baterías, 10 más particularmente, un conjunto de baterías para un vehículo eléctrico.

## ANTECEDENTES DE LA INVENCIÓN

El funcionamiento de un vehículo eléctrico requiere de un almacén de energía, normalmente constituido por un grupo de baterías interconectadas según diferentes combinaciones serie/paralelo según cada caso. Aunque las características de las baterías que se utilizan en un mismo coche eléctrico son similares, puede haber pequeñas diferencias entre ellas debido a los procesos de fabricación, principalmente en cuanto al nivel de energía eléctrica que son capaces de almacenar o a parámetros como la corriente de auto descarga. Estas diferencias, aun siendo pequeñas, provocan que en la repetición de ciclos de carga y descarga algunas baterías acumulen tensión por encima del resto y que otras se descarguen más de lo debido.

Sin embargo, para alargar lo máximo posible su vida útil es aconsejable mantener en todo momento las baterías dentro de las especificaciones del fabricante, que normalmente consisten en unos niveles de tensión de celda máximo y mínimo. Esto es especialmente importante en ciertos tipos de baterías, como por ejemplo en las baterías de ión de litio, cuya vida útil puede verse reducida enormemente si se sobrecargan o se sobredescargan. Este problema es especialmente grave en vista del elevado precio de las baterías.

En consecuencia, es necesario controlar de un modo sencillo el estado de cada una de las baterías de forma individual, de forma que el mal funcionamiento de una de ellas sea rápidamente detectado y no comprometa el funcionamiento del conjunto completo. Además, es también muy importante conocer en todo momento el nivel de energía disponible dentro de cada batería para poder estimar el alcance de un vehículo eléctrico.

Por otro lado, es critico controlar la adecuada conexión de las baterías entre si, ya que, debido a las elevadas intensidades que circulan en un vehículo 10 eléctrico, la existencia de pequeñas resistencias causadas por mal contacto eléctrico entre las baterías pueden provocar perdida de potencia y sobrecalentamiento.

Además, el sistema debe controlar también las situaciones catastróficas, por ejemplo un consumo de corriente de las baterías por encima del máximo marcado para el motor eléctrico, como el que se produciría en caso de un cortocircuito accidental a la salida del bloque de las baterías. Estas situaciones no siempre pueden controlarse adecuadamente con un fusible convencional, ya que el tiempo de actuación de un fusible de gran tamaño es muy largo, y además necesita una corriente muy elevada respecto a la corriente normal de trabajo para que pueda fundirse.

Aunque existen actualmente algunos sistemas que realizan algunas de estas funciones, ninguno de ellos ha resuelto aún el problema de modo satisfactorio.

### **DESCRIPCIÓN DE LA INVENCIÓN**

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El sistema de la presente invención resuelve la problemática anterior 30 gracias a un sistema que mide la carga de cada batería, su temperatura y la corriente de carga/descarga que la atraviesa, y en función de estos datos decide si es necesario disipar parte de la energía aportada a una batería

particular durante una operación de carga, o bien tomar otro tipo de medidas. Para ello, el sistema de gestión de carga de baterías en un vehículo eléctrico de la presente invención comprende:

- a) Un sensor de la corriente que atraviesa las baterías, preferentemente un sensor de corriente por efecto Hall. Este sensor de corriente permite controlar el estado de carga de las baterías, y detectar situaciones tales como una sobrecorriente por encima de los niveles máximos recomendados.
- b) Una pluralidad de tarjetas de control de baterías, cada una de las 10 cuales está conectada hasta a 12 baterías y hasta a 3 sensores de temperatura. Los sensores de temperatura sirven para comprobar la temperatura en la zona de las baterías y, adicionalmente, aseguran que en el proceso de balanceo por cargas resistivas de las baterías no se produzcan sobrecalentamientos en las tarjetas de control de baterías. En este contexto, un proceso de balanceo 15 consiste en disipar en calor la energía sobrante de una batería particular durante una operación de carga. Cada tarjeta de control de baterías comprende un conversor A/D que convierte las tensiones de las baterías de analógico a digital, una pluralidad de resistencias de potencia que disipan energía sobrante durante 20 procesos de balanceo, y un microcontrolador que controla su funcionamiento. Preferentemente, el conversor A/D es un modelo LTC6802, aunque podría ser cualquier otro modelo.
- c) Un coordinador de tarjetas de control de baterías conectado al
   microcontrolador de cada tarjeta de control de baterías y a un sensor de la corriente que atraviesa las baterías, preferentemente un sensor de corriente de efecto Hall. El coordinador de tarjetas recibe los datos acerca del nivel de carga de las baterías y su temperatura (de las tarjetas de control de baterías), así como acerca de la corriente que las atraviesa (del sensor de corriente), y determina si
   es necesario un proceso de balanceo para seguir dentro de las especificaciones del fabricante de la batería.

d) Un bus con aislamiento galvánico, que interconecta las baterías, el controlador y el sensor de corriente. De este modo, es posible solucionar las diferencias de tensión que generan las propias baterías que se pretenden controlar.

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## **BREVE DESCRIPCIÓN DE LOS DIBUJOS**

La Fig. 1 muestra un esquema de las diferentes partes que componen el sistema de la invención.

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## **REALIZACIÓN PREFERENTE DE LA INVENCIÓN**

Se describe a continuación un ejemplo de sistema de acuerdo con la presente invención haciendo referencia a la figura adjunta, en la que se ha representado un conjunto de 15 baterías (B) que alimenta un motor (M) de un vehículo eléctrico y que está controlado por medio del sistema de la invención.

El sistema de este ejemplo comprende 3 tarjetas (T) de lectura respectivamente conectadas a conjuntos de 5 baterías (B1...B5; B6... B10;
B11...B15). Nótese que se trata únicamente de un ejemplo simplificado de la invención, ya que realmente cada tarjeta (T) es conectable hasta a 12 baterías (B). Cada tarjeta (T) de lectura está conectada además a 3 sensores de temperatura, que por simplicidad no se muestran en la figura. Los sensores de temperatura obtienen datos acerca de la temperatura en la zona de las baterías y de la propia tarjeta (T) de lectura.

El sistema comprende además un coordinador (C) de tarjetas conectado a través de un bus (B) aislado galvánicamente a las diferentes tarjetas (T) de lectura y al sensor (I) de corriente. De este modo, cada tarjeta (T) recibe periódicamente datos acerca de las baterías (B) a las que está conectada, en concreto nivel de carga y temperatura, y envía estos datos al coordinador (C) de tarjetas a través del bus (B) aislado galvánicamente. El coordinador (C), teniendo

también en cuenta la corriente de carga que recibe del sensor (I) de corriente, realiza los cálculos necesarios para determinar si es necesario balancear la carga de alguna de las baterías (B) y, en ese caso, se comunica con la tarjeta (T) correspondiente para enviar las órdenes adecuadas.

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Más concretamente, durante los procesos de descarga, el sistema verifica el estado de las baterías (B) comprobando posibles excesos de temperatura y controlando, a través de la medida simultánea de la tensión y de la corriente, el estado de la carga de cada batería (B) y el valor de su resistencia interna gracias a la variación de la tensión en función de la corriente consumida en la descarga. Además, durante la descarga el sistema comprueba la correcta integridad de todas las interconexiones de las baterías (B) entre sí al formar el circuito serie de baterías (B), ya que en caso de haber una conexión defectuosa que tuviera un cierto valor ohmico esta pequeña resistencia sería visualizada como un aumento 15 de la resistencia interna de esa misma batería (B).

Este sistema también asegura que durante el proceso de descarga ninguna de las baterías (B) llegue a tener una tensión por debajo de la tensión mínima de trabajo recomendada por el fabricante. Además, esta información de la cantidad de energía disponible en cada batería (B) durante la descarga puede utilizarse para forzar un modo de consumo reducido (por ejemplo, mediante la limitación de la velocidad máxima) en el vehículo eléctrico cuando la carga de las baterías (B) cae por debajo de un umbral determinado.

Por otro lado, durante procesos de carga, la función del sistema es comprobar que ninguna de las baterías (B) se carga por encima de la tensión máxima recomendada por el fabricante. En caso de detectarse esta situación, se activarán las líneas de salida necesarias para detener la carga y de esta forma evitar la sobrecarga de una batería (B).

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Adicionalmente, durante el proceso de carga se realiza un proceso de balanceado consistente en activar una corriente de descarga individual en

aquellas baterías (B) cuyo nivel de carga sea demasiado elevado. La carga sobrante se disipa por medio de unas resistencias colocadas sobre las propias tarjetas para aquellas baterías (B) que tengan una tensión por encima del resto de baterías (B). De esta forma, mediante la descarga selectiva de aquellas baterías (B) con tensiones más altas se consigue alcanzar en los pasos finales del proceso de balanceo una tensión uniforme y equivalente a la tensión máxima

de carga recomendada por el fabricante en todo el grupo de baterías (B).

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El sistema también es capaz de detectar cortocircuitos accidentales a la salida de las baterías, que pueden producirse de forma excepcional en situaciones como una mala manipulación de la zona del motor o en un accidente de circulación, complementando así la funcionalidad que puede proporcionar un fusible convencional. Para ello, el sensor (I) de corriente del sistema monitoriza de forma periódica la corriente consumida de las baterías (B), y en caso de detectar una corriente mayor a un determinado umbral durante un determinado tiempo, acciona una línea de salida que permite cortar con el procedimiento mecánico adecuado (normalmente un relé) la salida de las baterías (B) que va hacia el motor (M), salvaguardando de esta forma la integridad de las baterías (B).

### REIVINDICACIONES

1. Sistema de gestión de carga de baterías (B) de un vehículo eléctrico, caracterizado porque comprende:

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- una pluralidad de tarjetas (T) de lectura de baterías (B), cada una de las cuales está conectada a hasta 12 baterías (B) y a hasta 3 sensores de temperatura, y que comprende un conversor A/D que convierte las tensiones de las baterías (B) de analógico a digital, una pluralidad de resistencias de potencia que disipan energía sobrante durante operaciones de balanceo, y un microcontrolador que controla su funcionamiento;

- un coordinador (C) de tarjetas (T) de control de baterías (B) conectado al microcontrolador de cada tarjeta (T) de control de baterías (B) y a un sensor (I) de la corriente que atraviesa las baterías (B), que recibe la corriente, temperatura y carga de cada batería (B) y determina si es necesario una operación de balanceo para no superar la tensión máxima recomendada de alguna de las baterías (B); y

- un bus (B) con aislamiento galvánico que interconecta las baterías (B), el controlador (C) y el sensor (I) de corriente.

2. Sistema de acuerdo con la reivindicación anterior,

donde el sensor (I) de corriente es un sensor de corriente por efecto Hall.

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3. Coche eléctrico que comprende el sistema de gestión de baterías de cualquiera de las reivindicaciones 1-2.

4. Procedimiento de gestión de carga de baterías (B) de un vehículo
 30 eléctrico mediante el sistema de cualquiera de las reivindicaciones 1-2, caracterizado porque comprende verificar que ninguna de las baterías (B) se carga por encima de una tensión máxima recomendada, realizando descargas

individuales en baterías (B) cuyo nivel de carga sea excesivo.

5. Procedimiento de gestión de carga de baterías (B) según la reivindicación
 4, que además comprende monitorizar la resistencia interna de las baterías (B) y
 5 determinar la aparición de una conexión defectuosa en caso de detectar un aumento excesivo de alguna de ellas.

 Procedimiento de gestión de carga de baterías (B) según cualquiera de las reivindicaciones 4-5, que además comprende verificar que ninguna de las
 baterías (B) llegue a tener una tensión por debajo de la tensión mínima recomendada.

7. Procedimiento de gestión de carga de baterías (B) según cualquiera de las reivindicaciones 4-6, que además comprende limitar la velocidad del vehículo
 15 cuando la carga de las baterías (B) cae por debajo de un umbral inferior.

Procedimiento de gestión de carga de baterías (B) según la reivindicación
 que además comprende detener la carga cuando alguna de las baterías (B) supera la tensión máxima recomendada.

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9. Procedimiento de gestión de carga de baterías (B) según cualquiera de las reivindicaciones 4-6, que además comprende desconectar el motor (M) cuando la corriente por las baterías (B) supera un determinado umbral

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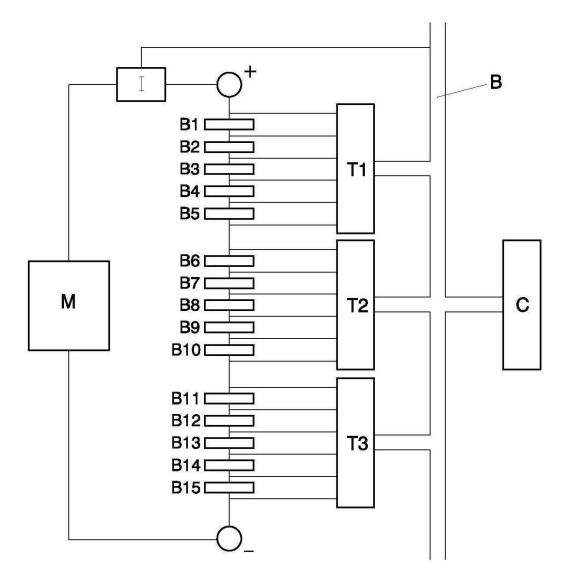
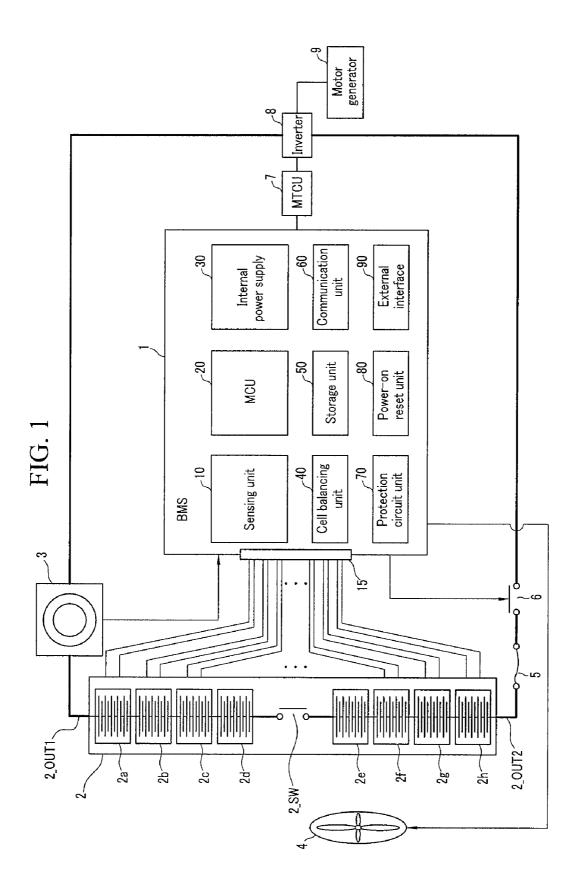


FIG. 1

# IE1



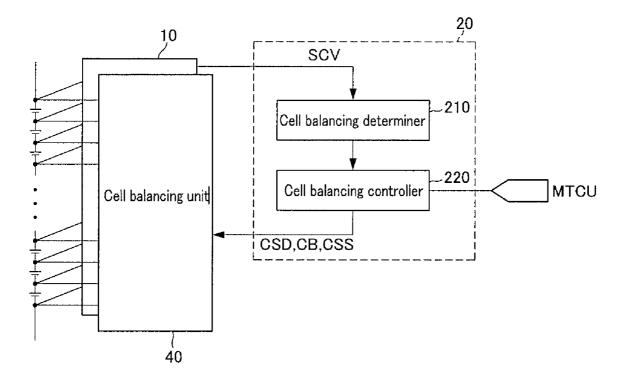
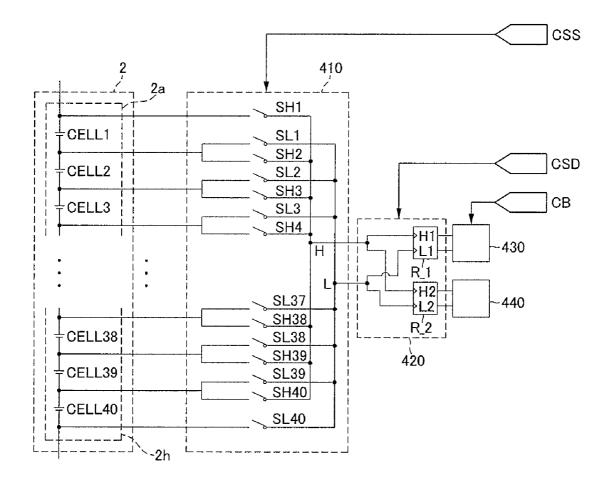


FIG. 2







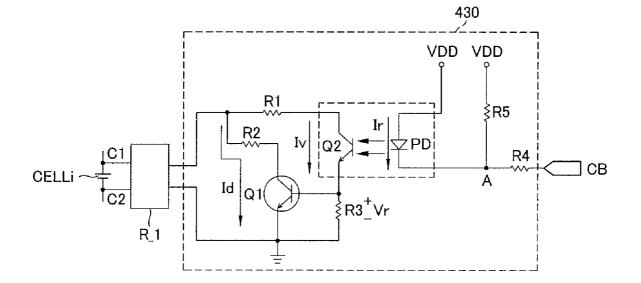
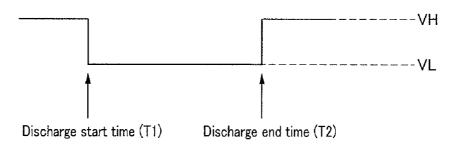


FIG. 5



## FIG. 6

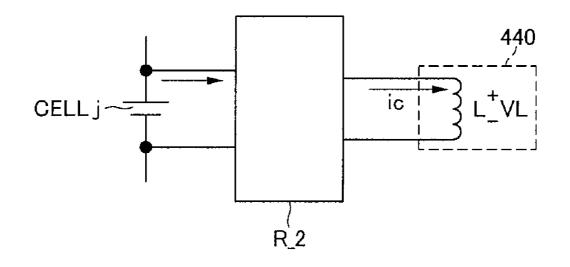
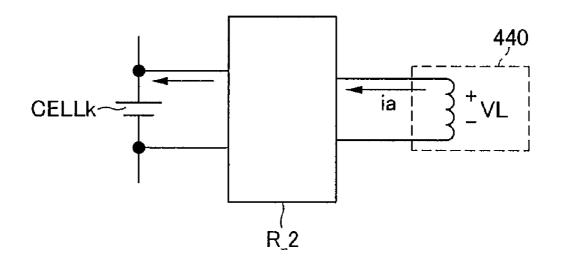
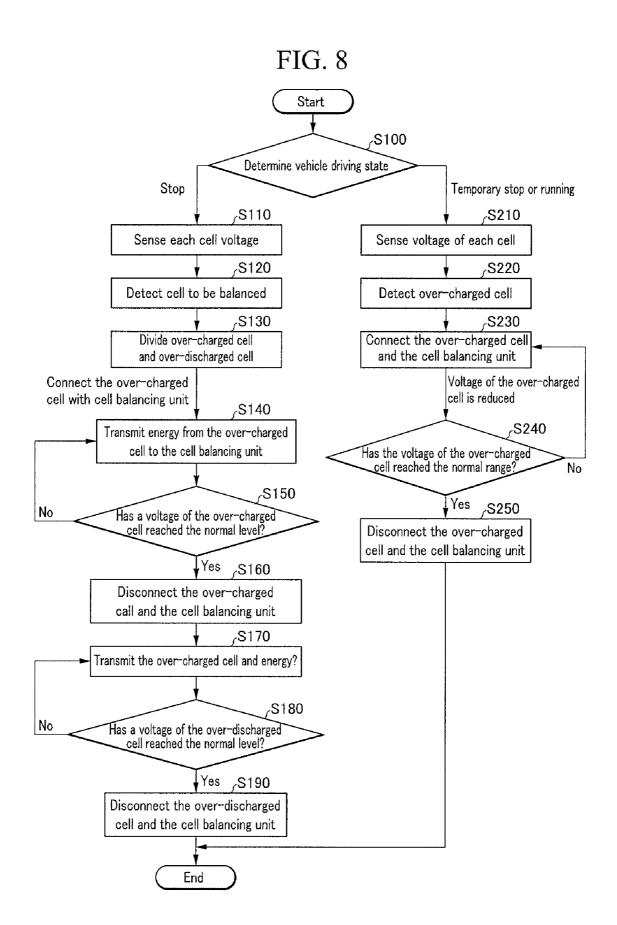


FIG. 7





## FIG. 9

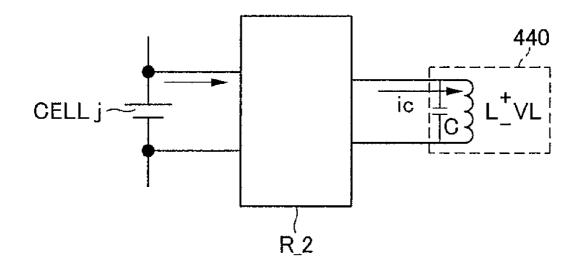
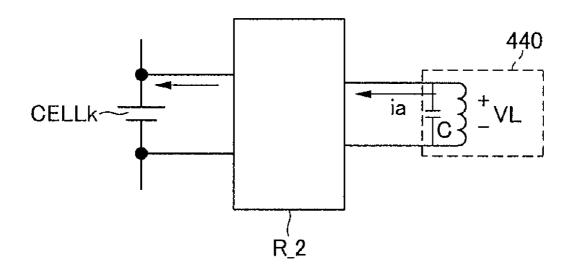


FIG. 10



### BATTERY MANAGEMENT SYSTEM AND DRIVING METHOD THEREOF

CROSS-REFERENCE TO RELATED APPLICATION

[0001]

### BACKGROUND OF THE INVENTION

[0002] 1. Field of the Invention

**[0003]** The present invention relates to a battery management system, and more particularly, to a battery management system for a vehicle and a driving method thereof.

[0004] 2. Description of the Related Art

**[0005]** Vehicles using an internal combustion engine powered by gasoline or diesel have caused serious air pollution. Accordingly, various efforts for developing electric or hybrid vehicles have recently been undertaken to reduce air pollution.

**[0006]** An electric vehicle uses an electric motor operating by electrical energy output by a battery. Because the electric vehicle mainly uses a battery formed by one battery pack including a plurality of rechargeable/dischargeable secondary cells, there are no emission gasses and the electric vehicle makes less noise.

**[0007]** A hybrid vehicle commonly refers to a gasolineelectric hybrid vehicle that uses gasoline to power an internalcombustion engine and a battery to power an electric motor. **[0008]** In order to enhance output power of a vehicle using a battery as a power source, the number of rechargeable battery cells has increased, and a battery management system (BMS) is required for efficiently managing a plurality of cells connected to each other.

**[0009]** When a plurality of cells are serially connected, balance between the cells is important. In order to maintain a balance between the cells, a voltage difference between a plurality of cells forming a battery is maintained within an allowable range. The balance between the cells will be referred to as "cell balancing." The cell balancing is closely related to cycle life and the output power of the battery. When the cells are not balanced, the cells deteriorate such that the cycle-life of the battery is reduced, thereby reducing the output power.

**[0010]** In addition, a conventional cell balancing method uses a discharge path that is formed according to control of a main controller. In this case, the discharge path includes a resistance. The main controller compares a voltage of each cell with a predetermined voltage by using a comparator so as to determine whether the cells are balanced. Therefore, a lot of time is required for determining cell balancing. In addition, constituent elements of a circuit for cell balancing are increased as the number of cells forming the battery is increased, and accordingly, wiring is additionally required for information exchange between a main controller and the cells. Further, when the number of cells increases, another main controller is required.

#### SUMMARY OF THE INVENTION

**[0011]** A battery management system for managing a battery is provided. The battery supplies power to a vehicle and includes a plurality of cells. The battery management system

includes a sensing unit for measuring a voltage of each of the plurality of cells. The battery management system detects at least one first cell among the plurality of cells needing to be balanced according to a measured voltage of each of the plurality of cells, and performs a cell balancing operation on said at least one first cell by using different methods depending upon an operational state of the vehicle.

**[0012]** According to an exemplary embodiment of the present invention, the battery management system further includes a cell balancing determiner for detecting said at least one first cell by comparing the measured voltage of each of the plurality of cells with a normal range. In addition, the battery management system includes a cell balancing controller for controlling a voltage of said at least one first cell to be discharged when the vehicle is in a temporary stop state, a running state, or a stop state. Furthermore, the battery management system includes a cell balancing unit for discharging the voltage of said at least one first cell.

**[0013]** According to an exemplary embodiment of the present invention, when the vehicle is in the stop state, the cell balancing determiner detects an over-charged cell or an overdischarged cell from among the plurality of cells, and the cell balancing unit receives energy from the over-charged cell and transmits the energy received from the over-charged cell to the over-discharged cell.

[0014] According to an exemplary embodiment of the present invention, the cell balancing unit includes a plurality of first cell balancing switches having first ends respectively coupled to the plurality of cells and second ends coupled to a first output terminal; a plurality of second cell balancing switches having first ends respectively coupled to the plurality of cells and second ends coupled to a second output terminal; a first cell balancing driver for discharging said at least one first cell when the vehicle is in the temporary stop state or the running state; a second cell balancing driver for transmitting the energy received from the over-charged cell to the over-discharged cell when the vehicle is in the stop state; a first relay for connecting the first cell balancing driver, the first output terminal, and the second output terminal; and a second relay for connecting the second cell balancing driver, the first output terminal, and the second output terminal.

**[0015]** According to an exemplary embodiment of the present invention, when the vehicle is in the temporary stop state or the running state, the cell balancing controller turns on the first relay and turns on first switches and second switches among the plurality of first cell balancing switches that correspond to said at least one first cell.

[0016] According to an exemplary embodiment of the present invention, the first relay has a first relay first end and a first relay second end. The first cell balancing driver includes a first resistor having a first resistor first end and a first resistor second end. The first resistor first end is coupled to the first relay first end. A second resistor has a second resistor first end and a second resistor second end. The second resistor first end is coupled to the first relay first end and to the first resistor first end. A first transistor has a first transistor first electrode, a first transistor second electrode, and a first transistor control electrode. The first transistor first electrode is coupled to the second transistor second end. The first transistor second electrode is coupled to the first relay second end. A second transistor has a second transistor first electrode and a second transistor second electrode. The second transistor first electrode is coupled to the first resistor second end. A third

resistor has a third resistor first end and a third resistor second end. The third resistor first end is coupled to the second transistor second electrode and to the first transistor control electrode. The third transistor second end is coupled to the first transistor second electrode and to the first relay second end. A photodiode forms a photocoupler with the second transistor. The photodiode has a photodiode anode electrode and a photodiode cathode electrode. A first power source is coupled to the photodiode anode electrode. A fourth resistor has a fourth resistor first end and a fourth resistor second end. The fourth resistor first end is coupled to the photodiode cathode electrode. A fifth resistor has a fifth resistor first end and a fifth resistor second end. The fifth resistor first end is coupled to the first power source. The fifth resistor second end is coupled to the fourth resistor first end and to the photodiode cathode electrode.

**[0017]** According to an exemplary embodiment of the present invention, the cell balancing controller transmits a discharge control signal to the fourth transistor second end to permit a driving current to flow to the photodiode.

**[0018]** According to an exemplary embodiment of the present invention, a bias current flows to the second transistor in response to the driving current, and the first transistor is turned on by a bias voltage applied to the first transistor control electrode and to the first transistor second electrode in response to the bias current and the third resistor.

**[0019]** According to an exemplary embodiment of the present invention, when the first transistor is turned on, charges of said at least one first cell are discharged by a discharge path having the second resistor and the first transistor so that a voltage of said at least one first cell is decreased.

**[0020]** According to an exemplary embodiment of the present invention, wherein when the vehicle is in the stop state, the cell balancing controller turns on first switches and second switches among the plurality of first cell balancing switches and the plurality of second cell balancing switches that correspond to the over-charged cell, and turns off the first switches and the second switches when a voltage of the over-charged cell reaches the normal range.

**[0021]** According to an exemplary embodiment of the present invention, after turning off the first switches and the second switches when the vehicle is in the stop state, the cell balancing controller turns on third switches among the plurality of first cell balancing switches corresponding to the over-discharged cell, and turns on fourth switches among the plurality of second cell balancing switches corresponding to the over-discharged cell.

**[0022]** According to an exemplary embodiment of the present invention, when the vehicle is in the stop state, the cell balancing controller turns off the third switches and the fourth switches when the voltage of the over-discharged cell reaches the normal range.

**[0023]** According to an exemplary embodiment of the present invention, the second relay has a second relay first end and a second relay second end. The second cell balancing driver includes an inductor having an inductor first end and an inductor second end. The inductor first end is coupled to the second relay first end. The inductor second end is coupled to the second relay second end. The inductor stores energy transmitted through the first switches, the second switches, and the second relay.

**[0024]** According to an exemplary embodiment of the present invention, the energy stored in the inductor is trans-

mitted to the over-discharged cell through the third switches, the fourth switches, and the second relay.

**[0025]** According to an exemplary embodiment of the present invention, the second cell balancing driver further comprises a capacitor coupled in parallel with the inductor.

**[0026]** According to an exemplary embodiment of the present invention, the battery management system further includes a connection port unit coupled to the plurality of cells. The connection port unit couples the battery management system and the plurality of cells.

**[0027]** According to an exemplary embodiment of the present invention, the battery management system is formed of one chip.

**[0028]** According to an exemplary embodiment of the present invention, a driving method of a battery management system for managing a battery is provided. The battery supplies power to a vehicle and includes a plurality of cells. The driving method includes a) determining a driving state of the vehicle; b) sensing a voltage of each of the plurality of cells; c) determining a cell needing to be balanced among the plurality of cells; and d) performing a cell balancing operation on the cell by using different methods depending upon an operational state of the vehicle.

**[0029]** According to an exemplary embodiment of the present invention, in a), when the vehicle is in a stop state, c) comprises detecting an over-charged cell and an over-discharged cell, and d) comprises performing the cell balancing operation on the over-charged cell and the over-discharged cell by transmitting energy from the over-charged cell to the over-discharged cell.

**[0030]** According to an exemplary embodiment of the present invention, in a), when the vehicle is in a temporary stop state or a running state, c) comprises detecting an overcharged cell, and d) comprises performing the cell balancing operation by discharging the over-charged cell.

#### BRIEF DESCRIPTION OF THE DRAWINGS

**[0031]** FIG. 1 schematically shows a battery, a battery management system (BMS), and a peripheral device of the BMS according to an exemplary embodiment of the present invention.

**[0032]** FIG. **2** shows a sensing unit, a cell balancing unit, and a main control unit (MCU) according to an exemplary embodiment of the present invention.

**[0033]** FIG. **3** shows a BMS that performs cell balancing according to a driving state of a vehicle.

**[0034]** FIG. **4** shows connection between a cell that needs to be balanced among a plurality of cells, a first relay, and a first cell balancing driver.

**[0035]** FIG. **5** shows a discharge control signal according to an exemplary embodiment of the present invention.

**[0036]** FIG. **6** shows a case in which energy from an overcharged cell is transmitted to a cell balancing driver.

**[0037]** FIG. **7** shows a case in which energy from the cell balancing driver is transmitted to an over-discharged cell.

**[0038]** FIG. **8** is a flowchart showing a BMS driving method according to an exemplary embodiment of the present invention.

**[0039]** FIG. **9** and FIG. **10** depict a cell balancing driver according to another exemplary embodiment of the present invention.

#### DETAILED DESCRIPTION

**[0040]** Throughout this specification and the claims that follow, when it is described that an element is "coupled" to another element, the element may be "directly coupled" to the other element or "electrically coupled" to the other element through a third element. In addition, unless explicitly described to the contrary, the word "comprise" and variations such as "comprises" and "comprising" will be understood to imply the inclusion of stated elements but not the exclusion of any other elements.

**[0041]** FIG. 1 schematically shows a battery, a BMS 1, and a peripheral device of the BMS 1 according to an exemplary embodiment of the present invention. The BMS 1 according to an exemplary embodiment of the present invention is formed of one chip.

[0042] The battery 2 includes a plurality of sub-packs 2a to 2h having a plurality of battery cells coupled in series to each other, and output terminal 2\_OUT1, an output terminal 2\_OUT2, and a safety switch 2\_SW provided between the sub-pack 2d and the sub-pack 2e. While eight sub-packs 2a to 2h are exemplified, and one sub-pack is a group of a plurality of battery cells in an exemplary embodiment of the present invention, it is not limited thereto. The safety switch 2 SW is manually turned on/off to guarantee safety for a worker when performing operations for the battery or replacing the battery. In the present exemplary embodiment, the safety switch 2\_SW is provided between the sub-pack 2d and the sub-pack 2e, but it is not limited thereto. The output terminal 2 OUT1 and the output terminal 2\_OUT2 are coupled to the inverter 8. [0043] The current sensor 3 measures an output current value of the battery 2 and outputs the measured output current value to the sensing unit 10 of the BMS 1. The current sensor 3 may be provided as a Hall current transformer (Hall CT) using a Hall element to measure a current value and outputting an analog current signal corresponding to the measured current value. The sensor 3 transmits information on the measured current value of the battery 2 to the BMS 1.

[0044] The cooling fan 4 cools down heat generated by charging and discharging the battery 2 in response to a control signal of the BMS 1, and prevents the battery 2 from being deteriorated by a temperature increase and prevents the charging and discharging efficiency from being deteriorated. [0045] The fuse 5 prevents an overflowing current, which may be caused by a disconnection or short circuit of the battery 2, from being transmitted to the battery 2. That is, when an overcurrent is generated, the fuse 5 is disconnected so as to interrupt the current from overflowing.

**[0046]** The main switch **6** turns on/off the battery **2** in response to the control signal of the BMS **1** when an unusual phenomenon, including an overflowed voltage, an over-current, and a high temperature, occurs.

[0047] The BMS 1 includes a sensing unit 10, a main control unit (MCU) 20 an internal power supply 30, a cell balancing unit 40, a storage unit 50, a communication unit 60, a protective circuit unit 70, a power-on reset unit 80, and an external interface 90. In addition, a plurality of cells are respectively electrically coupled to the BMS 1 through a connection port unit 15. The connection port unit 15 according to an exemplary embodiment of the present invention includes a number of connection terminals that is greater than the number of cells. Hereinafter, a plurality of connection terminals connected to the cells are referred to as cell connection terminals.

**[0048]** The sensing unit **10** is electrically coupled to a plurality of cells that form the battery **2** through a plurality of cell connection terminals. The sensing unit **10** measures each battery cell voltage and each battery terminal voltage and measures battery current by using an electrical signal transmitted from the Hall sensor **3**. Further, the sensing unit **10** measures a cell temperature and a battery temperature. The sensing unit **10** transmits the battery cell voltage, the battery terminal voltage is a voltage between the output terminals **2**\_OUT**1** and **2**\_OUT**2**, and the battery current is a current flowing to the battery.

[0049] The MCU 20 estimates a state of charge (SOC) and a state of health (SOH) based on the battery current, the battery terminal voltage, the respective battery cell voltages, the cell temperature, and an ambient temperature, which are transmitted from the sensing unit 10. The MCU 20 generates information that informs a state (i.e., SOC and SOH) of the battery 2 and transmits the information to the MoTor Control Unit (MTCU) 7 of the vehicle. The MTCU 7 controls charge or discharge of the battery 2 based on the SOC and the SOH. The MCU 20 according to an exemplary embodiment of the present invention can recognize the number of cells when the cell terminals of the connection port unit 15 are respectively coupled to the corresponding cells. The MCU 20 generates signals for controlling the cell balancing unit 40 corresponding to the recognized number of cells. A configuration and operation of the cell balancing unit 40 will be described in further detail later.

**[0050]** The internal power supply **30** supplies power by using a backup battery to transmit a voltage (e.g., a bias voltage) to the BMS **1** for operation.

[0051] The cell balancing unit 40 balances a SOC of each cell. That is, the cell balancing unit 40 controls a SOC of each cell to be within a reference range by charging/discharging a cell whose SOC is not within the reference range. Herein, the reference range may be set according to a battery terminal voltage, a SOC, a SOH, and temperature. In addition, the reference range may be set within a predetermined range that is set on the basis of an average value of the plurality of cell voltages. The cell balancing unit 40 operates according to a control signal of the MCU 20 that has received information on each cell voltage, or directly receives information on each cell voltage from the sensing unit 10, determines whether a cell is balanced or not based on the information, and balances each cell. The cell balancing unit 40 is electrically coupled to the respective cells through the plurality of cell connection terminals. The cell balancing unit 40 according to an exemplary embodiment of the present invention may use a different cell balancing method according to a driving state of the vehicle. The BMS and a driving method thereof according to an exemplary embodiment of the present invention will be described in further detail later with reference to FIG. 2, FIG. 3, FIG. 4, and FIG. 5.

**[0052]** When the BMS 1 is turned off, the storage unit **50** stores data including a current SOC and a current SOH. Herein, the storage unit **50** may be provided as a non-volatile electrically erasable programmable read-only memory (EE-PROM).

**[0053]** The communication unit **60** communicates with the MTCU **7** of the vehicle. The protective circuit unit **70** uses firmware to protect the battery **2** from shocks, over-flowed currents, and low voltages.

[0054] The power-on reset unit 80 resets the overall system when the power source of the BMS 1 is turned on. The external interface 90 couples auxiliary devices for the BMS 1, such as the cooling fan 4 and the main switch 6. While the cooling fan 4 and the main switch 6 are shown as the auxiliary devices in the present exemplary embodiment, it is not limited thereto.

[0055] The MTCU 7 determines a torque state based on information on an accelerator, a brake, and a vehicle speed, and controls an output of the motor-generator 9 so that the output corresponds to the torque information. That is, the MTCU 7 controls the motor-generator 9 to output an output corresponding to the torque information by controlling a switching operation of the inverter 8. In addition, the MTCU 7 receives the SOC of the battery 2 from the MCU 20 through the communication unit 60 and controls the SOC level of the battery 2 to be a target level (e.g., 55%). For example, when the SOC level transmitted from the MCU 20 is less than 55%, the MTCU 7 controls the switching operation of the inverter 8 so as to output power toward the battery 2 and charge the battery 2. In this case, the battery pack current has a negative (-) value. When the SOC level is greater than 55%, the MTCU 7 controls the switching operation of the inverter 8 so as to output the power toward the motor-generator 9 and discharge the battery 2. In this case, the battery pack current has a positive (+) value.

**[0056]** The inverter **8** controls the battery **2** to be charged or discharged in accordance with a control signal of the MTCU **7**.

**[0057]** The motor generator **9** uses the electrical energy of the battery **2** to drive the vehicle based on the torque information transmitted from the MTCU **7**.

[0058] As described, the BMS 1 according to an exemplary embodiment of the present invention transmits battery information and a charge/discharge control signal to the MTCU 7 and the MTCU 7 performs charge or discharge, accordingly. In addition, in the BMS 1, each cell is balanced by using the sensing unit 10, the cell balancing unit 40, and the MCU 20. [0059] Hereinafter, a configuration and operation for performing cell balancing according to an exemplary embodiment of the present invention will be described in further detail with reference to FIG. 2, FIG. 3, FIG. 4, and FIG. 5. The BMS and the plurality of cells are electrically coupled through the respective cell connection terminals. In FIG. 2 and FIG. 3, the connection port unit 15 is omitted for convenience of description.

[0060] FIG. 2 shows the sensing unit 10, the cell balancing unit 40, and the MCU 20 of the BMS 1 according to an exemplary embodiment of the present invention. These constituent elements are directly associated with the cell balancing operation. Wires connected to the lateral ends of each cell are coupled to the sensing unit 10 and the cell balancing unit 40, and the sensing unit 10 is coupled to the MCU 20, and therefore a measured voltage can be transmitted therethrough. In addition, the cell balancing unit 40 is coupled to the MCU 20. The MCU 20 receives information on a driving state of the vehicle from the MTCU 7. According to an exemplary embodiment of the present invention, the MCU 20 receives information on a cell voltage from the sensing unit 10, determines the necessity of the cell balancing operation, and controls the cell balancing operation, but this is not restrictive. Alternatively, the cell balancing unit 40 may individually include a micro control unit (MCU), may receive cell voltage information from the sensing unit 10 and the vehicle driving state from the MTCU 7, may determine the necessity of the cell balancing operation, and may control the cell balancing operation. That is, in an alternatively embodiment, the cell balancing determiner 210 and a cell balancing controller 220 can be included in the cell balancing unit 40.

[0061] The sensing unit 10 transmits a cell voltage signal SCV generated by measuring a voltage of each cell to the cell balancing determiner 210. The cell balancing determiner 210 compares a voltage of the cell voltage signal SCV with a normal range. When the voltage of the cell voltage signal SCV is not in the normal range, the cell balancing determiner 210 determines that the cell balancing operation needs to be performed. When the voltage of the cell voltage signal SCV is less than the normal range, the corresponding cell is determined to be a discharge target cell, and when the voltage of the cell voltage signal SCV is greater than the normal range, the corresponding cell is determined to be a charge target cell. The cell balancing controller 220 receives information on a cell that needs to be balanced from the cell balancing determiner 210, and transmits a cell balancing control signal CSD to the cell balancing unit 40. The cell balancing controller 220 generates the cell balancing control signal CSD according to the information on the vehicle driving state, transmitted from the MTCU 7, and information on the cell that needs to be balanced, and transmits the cell balancing control signal CSD to the cell balancing unit 40.

**[0062]** The vehicle driving state or operational state according to an exemplary embodiment of the present invention can be divided into a stop state, a running state, and a temporary stop state. The stop state means that the vehicle is parked and operation of the vehicle is stopped, and the temporary stop state means that the vehicle is temporarily stopped while it is running.

[0063] The BMS 1 according to an exemplary embodiment of the present invention detects a cell that needs to be balanced and performs the cell balancing operation when the vehicle driving state is set to the stop state or the temporary stop state. A different cell balancing method is used in accordance with a vehicle driving state, and a control signal CSD is also changed, accordingly. The different cell balancing method will be described in further detail later with reference to FIG. 3. The cell balancing controller 220 generates a connection control signal CSS that controls a connection between the cell that is determined to be balanced and the cell balancing unit 40 and transmits the connection control signal CSS to the cell balancing unit 40, and generates a discharge control signal CB for controlling discharge of a cell and transmits the discharge control signal CB to the cell balancing unit 40.

**[0064]** FIG. **3** shows a BMS that performs a cell balancing operation according to a vehicle driving state according to an exemplary embodiment of the present invention. The battery **2** includes a plurality of cells (e.g., CELL**1** to CELL**40**), and every five cells forms one pack. Therefore, the battery **2** is formed of eight packs **2***a* to **2***h*.

[0065] The cell balancing unit 40 includes a cell balancing multiplexer 410, a cell balancing connector 420, a first cell balancing driver 430, and a second cell balancing driver 440. [0066] The cell balancing multiplexer 410 includes a plurality of cell balancing switches SH1 to SH40 and SL1 to SL40, and the plurality of cell balancing switches SH1 to SH40 and SL1 to SL40 are turned on/off according to the connection control signal CSS. A first end of the cell balancing switch SH1 is coupled to a first end of the cell CELL1, and a second end of the cell balancing switch SH1 is electrically coupled to a first output terminal H. A first end of the cell balancing switch SL1 is coupled to a second end of the cell cell balancing switch SL1 is electrically coupled to a second end of the cell balancing switch SL1 is electrically coupled to a second end of the cell balancing switch SL1 is electrically coupled to a second end of the cell balancing switch SL1 is electrically coupled to a second output terminal L.

[0067] In a like manner to the above, each of the cell balancing switches SH1 to SH40 has one end electrically coupled to one end of the corresponding cell and the other end electrically coupled to the first output terminal H. In addition, each of the cell balancing switches SL1 to SL40 has one end coupled to one end of the corresponding cell and the other end electrically coupled to the second output terminal L. The plurality of cell balancing switches SH1 to SH40 and SL1 to SL40 are turned on/off in response to the connection control signal CSS transmitted from the cell balancing controller 220. The connection control signal CSS according to an exemplary embodiment of the present invention refers to a plurality of signals respectively corresponding to the plurality of cell balancing switches SH1 to SH40 and SL1 to SL40.

**[0068]** Particularly, for example, when the cell balancing determiner **210** determines that the cell CELL3 needs to be balanced, the cell balancing controller **220** transmits a connection control signal CSS that turns on the cell balancing switches SH3 and SL3 to the cell balancing switches SH3 and SL3. Then, the cell balancing switches SH3 and SL3 are turned on, and the first output terminal H and the second output terminal L are respectively coupled to the lateral ends of the cell CELL3.

[0069] The cell balancing connector 420 is coupled to the first and second output terminals H and L of the cell balancing multiplexer 410, the first cell balancing driver 430, and the second cell balancing driver 440. The cell balancing connector 420 includes first and second relays  $R_1$  and  $R_2$  and output terminals H1 and H2 of the first and second relays  $R_1$  and  $R_2$  are coupled to the first output terminal H, and input terminals L1 and L2 of the first and second relays  $R_1$  and  $R_2$  are coupled to the second output terminal L. The first and second relays  $R_1$  and  $R_2$  are coupled to the second output terminal L. The first and second relays  $R_1$  and  $R_2$  operate according to a cell balancing control signal CSD.

[0070] In further detail, the cell balancing control signal CSD turns on the first relay  $R_1$  when the vehicle is in the temporary stop state or the running state, and electrically couples the first cell balancing driver 430 with a cell that needs to be balanced. When the first cell balancing driver 430 is coupled with the cell, a discharge path is formed by using a resistor and reduces a voltage of the cell to thereby control the voltage of the cell to be included within the normal range. In this case, a discharge control signal CB that controls the start and end of discharge of the cell that needs to be balanced is transmitted to the first cell balancing driver 430.

[0071] When the vehicle is in the stop state, the cell balancing control signal CSD turns on the second relay R\_2, and electrically couples the second balancing driver 440 with a cell that needs to be balanced.

**[0072]** When the vehicle is in the stop state, the second cell balancing driver **440** receives energy from a cell having a relatively high voltage among cells that need to be balanced, and transmits the energy to a cell having a relatively low voltage among the cells that need to be balanced. Then, the voltage of the cell having the relatively high voltage is

reduced, and the voltage of the cell having the relatively low voltage is increased such that each of the respective cells has a voltage within the normal range.

**[0073]** The first cell balancing driver **430** will be described in further detail with reference to FIG. **4**. FIG. **4** shows a case in which a cell CELLi among the plurality of cells needs to be balanced. In this case, the cell CELLi is coupled with the first cell balancing driver **430** through the first relay R\_1.

**[0074]** As shown in FIG. **4**, the first cell balancing driver **430** includes a plurality of resistors R1 to R5, a second transistor that forms a photocoupler, a photodiode PD, and a first transistor. The first cell balancing driver **430** is controlled by the discharge control signal CB.

**[0075]** In further detail, the cell balancing determiner **210** determines that the cell CELLi needs to be balanced, and transmits information on the cell CELLi to the cell balancing controller **220**. The information on the cell CELLi includes identification information according to a location of the cell CELLi. Then, the cell balancing controller **220** generates a connection control signal CSS according to the identification information of the cell CELLi.

[0076] According to the connection control signal CSS, cell balancing switches SHi and SLi of the cell balancing multiplexer 410 are turned on. In addition, the cell balancing controller 220 generates a cell balancing control signal CSD for selecting one of the first cell balancing driver 430 or the second cell balancing driver 440 in accordance with a vehicle driving state transmitted from the MTCU 7, and transmits the cell balancing control signal CSD to the cell balancing connector 420. When the current vehicle driving state is set to the temporary stop state, the cell balancing control signal CSS turns on the first relay R\_1 to operate the first cell balancing driver 430. Then, as shown in FIG. 4, the cell CELLi and the first cell balancing driver 430 are electrically coupled.

[0077] In this case, the discharge control signal CB is generated by the cell balancing controller 220, and controls the cell CELLi to be discharged until a voltage of the cell CELLi reaches the normal range. The cell balancing determiner 210 compares the voltage of the cell CELLi, transmitted from the sensing unit 10, with the normal range in real time, and notifies the cell balancing controller 220 that the cell balancing operation is completed when the voltage of the cell CELLi is within the normal range. Then, the cell balancing controller 220 controls the discharge control signal CB to stop the discharging of the cell CELLi. The discharge control signal CB according to an exemplary embodiment of the present invention is changed to a first level (i.e., a low level) at a discharge start time, and is maintained at the first level while the discharge is continued. When the cell balancing operation is completed, the discharge control signal is changed to a second level that is higher than the first level.

[0078] FIG. 5 shows a discharge control signal CB according to an exemplary embodiment of the present invention. As shown in FIG. 5, when the discharge control signal CB of the first level VL is transmitted to the first cell balancing driver 430 at the discharge start time T1, a voltage VA between a voltage of a power source VDD and the voltage of the first level VL is applied to a node A of a resistor R4 and a resistor R5. The voltage VA is determined by a resistance ratio between the resistors R4 and R5. A level of the voltage VA according to an exemplary embodiment of the present invention is set to provide a driving current to the photodiode PD. [0079] Therefore, when the discharge control signal CB of the first level is applied, a driving current Ir is generated in the

photodiode PD due to a voltage difference between the voltage of the power source VDD and the voltage VA. The photodiode PD and the second transistor Q2 that forms the photocoupler are turned on in correspondence with the driving current Ir flowing to the photodiode PD, and a bias current Iv is generated. When the bias current Iv flows, the resistor R3 is applied with the bias current Iv, and the first transistor Q1 is turned on. Through the turned-on first transistor Q1, a discharge current Id flows. Then, a discharge path formed from a first end C1 of the cell CELLi to a second end C2 is formed through the turned-on relay R\_1 and the resistor R2, and the voltage of the cell CELLi is quickly reduced through the discharge current Id. When the voltage of the cell CELLi is reduced through the discharge and thus it is included within the normal range, the discharge control signal CB is changed to the second level VH at a time T2, and the discharge current Id is not generated. The first and second transistors according to an exemplary embodiment of the present invention are respectively bipolar junction transistors (BJT) having a collector electrode and an emitter electrode as two electrodes and a base electrode as a control electrode. The first and second transistors are n-channel type transistors. However, this is not restrictive, and another transistor performing the same operation can also be used.

**[0080]** As described, the BMS according to an exemplary embodiment of the present invention controls the voltage of the cell CELLi to be included with the normal range by forcing a discharge so as to perform the cell balancing operation when the vehicle driving state is set to the temporary stop state.

**[0081]** In the case that the vehicle driving state is set to the stop state, cell balancing by using the BMS and the driving method thereof according to an exemplary embodiment of the present invention will be described in further detail with reference to FIG. **6** and FIG. **7**.

**[0082]** Unlike the temporary stop or the running state, a discharge does not need to be forcibly performed when the vehicle is in the stop state. That is, the cell balancing operation does not need to be quickly completed. Therefore, the BMS and the driving method thereof according to an exemplary embodiment of the present invention move energy of a cell having a relatively high voltage to a cell having a relatively low voltage so as to include voltages of a plurality of cells within the normal range.

**[0083]** FIG. **6** and FIG. **7** show an operation of the second cell balancing driver **440** of the cell balancing unit **40** according to an exemplary embodiment of the present invention. A cell having a voltage that is greater than an average voltage of the voltages of the plurality of cells is referred to as an overcharged cell, and a cell having a voltage that is less than the average voltage is referred to as an over-discharged cell. The normal range may correspond to a range including an error range based on the average voltage, or it may correspond to the average voltage.

[0084] FIG. 6 shows a case in which energy of an overcharged cell CELLj is transmitted to the cell balancing driver 440. FIG. 7 shows a case in which the cell balancing driver 440 transmits energy to an over-discharged cell CELLk.

**[0085]** The cell balancing determiner **210** detects a cell that needs to be balanced from among the plurality of cells and transmits the detected cell to the cell balancing controller **220**, and the cell balancing controller **220** divides an over-charged cell CELL<sub>j</sub> having a voltage that is greater than the normal

range and an over-discharged cell CELLk having a voltage that is less than the normal range among cells that need to be balanced.

[0086] In addition, the cell balancing controller 220 transmits a connection control signal CSS to the cell balancing multiplexer 410 so as to turn on the cell balancing switches SHj and SLj, and transmits a cell balancing control signal CSD to the second relay R\_2 so as to turn on the second relay R\_2. Then, as shown in FIG. 6, the lateral ends of an inductor L of the second cell balancing driver 440 are coupled to the over-charged cell CELLj and a current ic flows to the inductor L, and accordingly a voltage VL is increased and a voltage of the over-charged cell CELLj is decreased. When the voltage of the over-charged cell CELLj reaches the normal range, the cell balancing determiner 210 informs the cell balancing controller 220 that the voltage of the over-charged cell CELL j has reached the normal range. The cell balancing controller 220 disconnects the cell balancing switches SHj and SLj and the second relay R\_2 by using the connection control signal CSS and the cell balancing control signal CSD.

[0087] In addition, the cell balancing controller 220 turns on the cell balancing switches SHk and SLk and the second relay R\_2 by using the connection control signal CSS and the cell balancing control signal CSD, and electrically couples the over-discharged cell CELLk and the inductor L. Then, a voltage of the inductor L becomes greater than a voltage of the over-discharged cell CELLk so that a current ia flows to the over-discharged cell CELLk. Accordingly, energy stored in the inductor L is transmitted to the over-discharged cell CELLk and the voltage of the cell CELLk is increased. When the voltage of the cell CELLk reaches the normal range, the cell balancing determiner 210 informs the cell balancing controller 220 that the voltage of the cell CELLk has reached the normal range. The cell balancing controller 220 turns off the cell balancing switches SHk and SLk by using the connection control signal CSS and the cell balancing control signal CSD, and disconnects the cell CELLk and the inductor L.

**[0088]** Although in FIG. **6** and FIG. **7** the second cell balancing driver **440** includes only the inductor L according to an exemplary embodiment of the present invention, it is not restrictive. For example, as depicted in FIG. **9** and FIG. **10**, the second cell balancing driver **440** may further include a capacitor C for maintaining a voltage of the inductor L at a constant level.

**[0089]** As described, the plurality of cell voltages can be balanced to the average voltage by using the second cell balancing driver **440** including the inductor L.

**[0090]** A driving method of the BMS according to an exemplary embodiment of the present invention will be described with reference to FIG. **8**.

**[0091]** FIG. **8** is a flowchart showing a driving method of the BMS according to an exemplary embodiment of the present invention.

[0092] As shown in FIG. 8, the BMS 1 determines a driving state of the vehicle in step S100. When it is determined in step S100 that the vehicle is in the stop state, the BMS 1 senses a voltage of respective cells in step S110. The BMS 1 detects cells that need to be balanced to the average voltage based on the sensing result in step S120. The BMS 1 divides an over-charged cell and an over-discharged cell from among the detected cells in step S130. The BMS 1 controls the cell balancing unit 40 to receive energy from the over-charged cell in step S140. In addition, the BMS 1 determines whether a voltage of the over-charged cell reaches the normal range in

step S150. When it is determined in step S150 that the voltage of the over-charged cell reaches the normal range, the BMS 1 disconnects the over-charged cell and the cell balancing unit 40 in step S160.

[0093] When it is determined in step S150 that the voltage of the over-charged cell has not reached the normal range, the BMS 1 controls the cell balancing unit 40 to receive the energy from the over-charged cell. After step S160, the BMS 1 connects the cell balancing unit 40 and the over-discharged cell, and the cell balancing unit 40 transmits the received energy to the over-discharged cell in step S170. The BMS 1 determines whether a voltage of the over-discharged cell reaches the normal range in step S180. When it is determined in step S180 that the voltage of the over-discharged cell has reached the normal range, the BMS 1 disconnects the overdischarged cell and the cell balancing unit 40 in step S190. When it is determined in step S180 that the voltage of the over-discharged cell has not reached the normal range, the BMS 1 controls the cell balancing unit 40 to transmit the energy to the over-discharged cell.

[0094] When it is determined in step S100 that the vehicle is in the temporary stop state or the running state, the BMS 1 senses a voltage of respective cells in step S210. The BMS 1 detects an over-charged cell that needs to be balanced based on the sensing result in step S220. The BMS 1 connects the detected over-charge cell and the cell balancing unit 40 in step S230. Then, the over-charged cell is discharged, and thus a voltage of the over-charged cell is reduced. The BMS 1 determines whether the reduced voltage has reached the normal range in step S240. When it is determined in S250 that the reduced voltage of the over-charged cell has reached the normal range, the BMS 1 disconnects the over-charged cell and the cell balancing unit 40 in step S250.

**[0095]** As described, the cell balancing operation can be performed in accordance with a driving state of the vehicle by using the BMS and the driving method thereof according to an exemplary embodiment of the present invention. According to an exemplary embodiment of the present invention, equalization and balancing between cells can be efficiently performed. In addition, because the BMS is formed of one chip, there is no need for increasing the amount of wiring or replacing the main controller even through the number of cells is changed.

**[0096]** While this invention has been described in connection with what is presently considered to be practical exemplary embodiments, it is to be understood that the invention is not limited to the disclosed embodiments, but, on the contrary, is intended to cover various modifications and equivalent arrangements included within the spirit and scope of the appended claims.

What is claimed is:

1. A battery management system for managing a battery having a plurality of cells for supplying power to a vehicle, the battery management system comprising:

- a sensing unit for measuring a voltage of each of the plurality of cells,
- wherein the battery management system detects at least one first cell among the plurality of cells needing to be balanced according to a measured voltage of each of the plurality of cells, and performs a cell balancing operation on said at least one first cell by using different methods depending upon an operational state of the vehicle.

2. The battery management system of claim 1, further comprising:

- a cell balancing determiner for detecting said at least one first cell by comparing the measured voltage for each of the plurality of cells with a normal range;
- a cell balancing controller for controlling a voltage of said at least one first cell to be discharged when the vehicle is in a temporary stop state, a running state, or a stop state; and
- a cell balancing unit for discharging the voltage of said at least one first cell.

3. The battery management system of claim 2, wherein when the vehicle is in the stop state, the cell balancing determiner detects an over-charged cell or an over-discharged cell from among the plurality of cells, and the cell balancing unit receives energy from the over-charged cell and transmits the energy received from the over-charged cell to the over-discharged cell.

4. The battery management system of claim 3, wherein the cell balancing unit comprises:

- a plurality of first cell balancing switches having first ends respectively coupled to the plurality of cells and second ends coupled to a first output terminal;
- a plurality of second cell balancing switches having first ends respectively coupled to the plurality of cells and second ends coupled to a second output terminal;
- a first cell balancing driver for discharging said at least one first cell when the vehicle is in the temporary stop state or the running state;
- a second cell balancing driver for transmitting the energy received from the over-charged cell to the over-discharged cell when the vehicle is in the stop state;
- a first relay for connecting the first cell balancing driver, the first output terminal, and the second output terminal; and
- a second relay for connecting the second cell balancing driver, the first output terminal, and the second output terminal.

5. The battery management system of claim 4, wherein when the vehicle is in the temporary stop state or the running state, the cell balancing controller turns on the first relay and turns on first switches and second switches among the plurality of first cell balancing switches and the plurality of second cell balancing switches that correspond to said at least one first cell.

6. The battery management system of claim 5, wherein the first relay has a first relay first end and a first relay second end, and wherein the first cell balancing driver comprises:

- a first resistor having a first resistor first end and a first resistor second end, the first resistor first end being coupled to the first relay first end;
- a second resistor having a second resistor first end and a second resistor second end, the second resistor first end being coupled to the first relay first end and to the first resistor first end;
- a first transistor having a first transistor first electrode, a first transistor second electrode, and a first transistor control electrode, the first transistor first electrode being coupled to the second transistor second end, the first transistor second electrode being coupled to the first relay second end;
- a second transistor having a second transistor first electrode and a second transistor second electrode, the second transistor first electrode being coupled to the first resistor second end;

- a third resistor having a third resistor first end and a third resistor second end, the third resistor first end being coupled to the second transistor second electrode and to the first transistor control electrode, the third transistor second end being coupled to the first transistor second electrode and to the first relay second end;
- a photodiode for forming a photocoupler with the second transistor, the photodiode having a photodiode anode electrode and a photodiode cathode electrode;
- a first power source coupled to the photodiode anode electrode;
- a fourth resistor having a fourth resistor first end and a fourth resistor second end, the fourth resistor first end being coupled to the photodiode cathode electrode; and
- a fifth resistor having a fifth resistor first end and a fifth resistor second end, the fifth resistor first end being coupled to the first power source, the fifth resistor second end being coupled to the fourth resistor first end and to the photodiode cathode electrode.

7. The battery management system of claim 6, wherein the cell balancing controller transmits a discharge control signal to the fourth transistor second end to permit a driving current to flow to the photodiode.

**8**. The battery management system of claim **7**, wherein a bias current flows to the second transistor in response to the driving current, and the first transistor is turned on by a bias voltage applied to the first transistor control electrode and to the first transistor second electrode in response to the bias current and the third resistor.

**9**. The battery management system of claim **8**, wherein when the first transistor is turned on, charges of said at least one first cell are discharged by a discharge path having the second resistor and the first transistor so that a voltage of said at least one first cell is decreased.

10. The battery management system of claim 4, wherein when the vehicle is in the stop state, the cell balancing controller turns on first switches and second switches among the plurality of first cell balancing switches and the plurality of second cell balancing switches that correspond to the over-charged cell, and turns off the first switches and the second switches when a voltage of the over-charged cell reaches the normal range.

11. The battery management system of claim 10, wherein after turning off the first switches and the second switches when the vehicle is in the stop state, the cell balancing controller turns on third switches among the plurality of first cell balancing switches corresponding to the over-discharged cell, and turns on fourth switches among the plurality of second cell balancing switches corresponding to the over-discharged cell.

12. The battery management system of claim 11, wherein when the vehicle is in the stop state, the cell balancing con-

troller turns off the third switches and the fourth switches when the voltage of the over-discharged cell reaches the normal range.

- **13**. The battery management system of claim **12**, wherein: the second relay has a second relay first end and a second relay second end;
- the second cell balancing driver comprises an inductor having an inductor first end and an inductor second end, the inductor first end being coupled to the second relay first end, the inductor second end being coupled to the second relay second end; and
- the inductor stores energy transmitted through the first switches, the second switches, and the second relay.

14. The battery management system of claim 13, wherein the energy stored in the inductor is transmitted to the overdischarged cell through the third switches, the fourth switches, and the second relay.

15. The battery management system of claim 14, wherein the second cell balancing driver further comprises a capacitor coupled in parallel with the inductor.

**16**. The battery management system of claim **1**, further comprising a connection port unit coupled to the plurality of cells, the connection port unit coupling the battery management system and the plurality of cells.

17. The battery management system of claim 1, wherein the battery management system is formed of one chip.

**18**. A driving method of a battery management system for managing a battery having a plurality of cells supplying power to a vehicle, the driving method comprising:

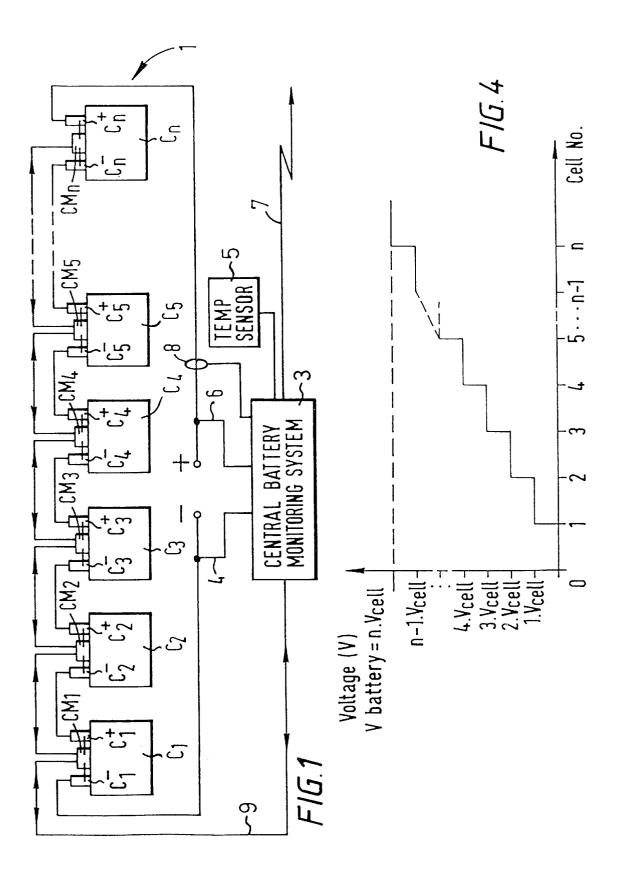
- a) determining a driving state of the vehicle;
- b) sensing a voltage for each of the plurality of cells;
- c) determining a cell among the plurality of cells needing to be balanced; and
- d) performing a cell balancing operation on the cell by using different methods depending upon an operational state of the vehicle.

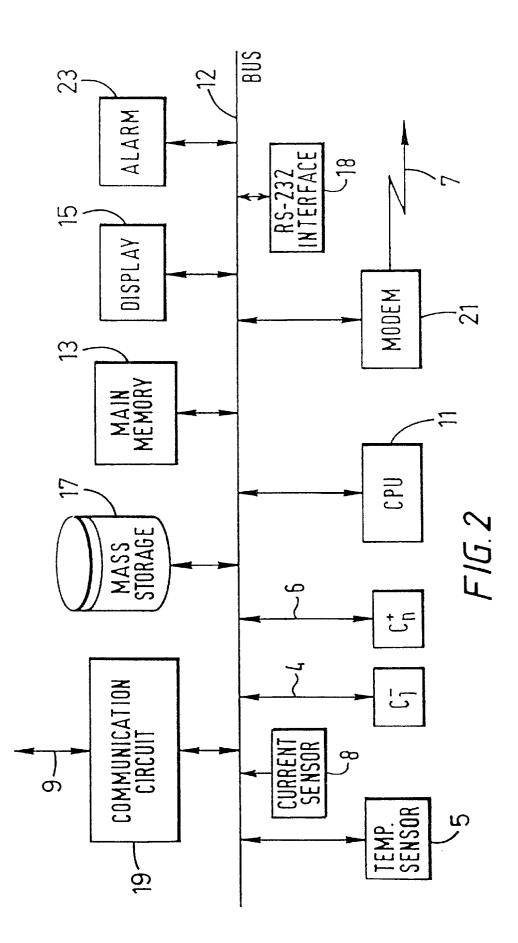
**19**. The driving method of claim **18**, wherein when the vehicle is in a stop state, determining a cell needing to be balanced comprises detecting an over-charged cell and an over-discharged cell, and performing a cell balancing operation comprises performing the cell balancing operation on the over-charged cell and the over-discharged cell by transmitting energy from the over-charged cell to the over-discharged cell.

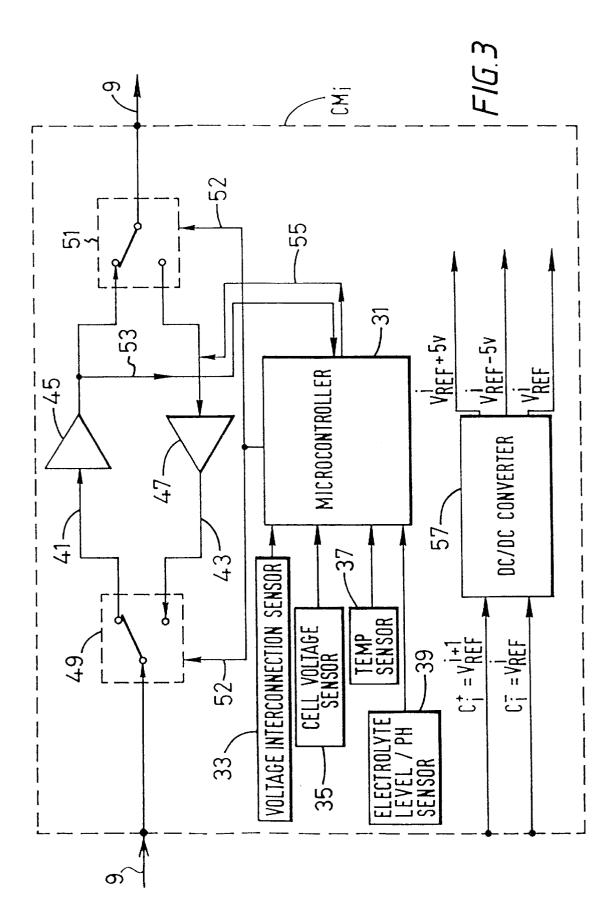
**20**. The driving method of claim **18**, wherein when the vehicle is in a temporary stop state or a running state, determining a cell needing to be balanced comprises detecting an over-charged cell, and performing a cell balancing operation comprises performing the cell balancing operation by discharging the over-charged cell.

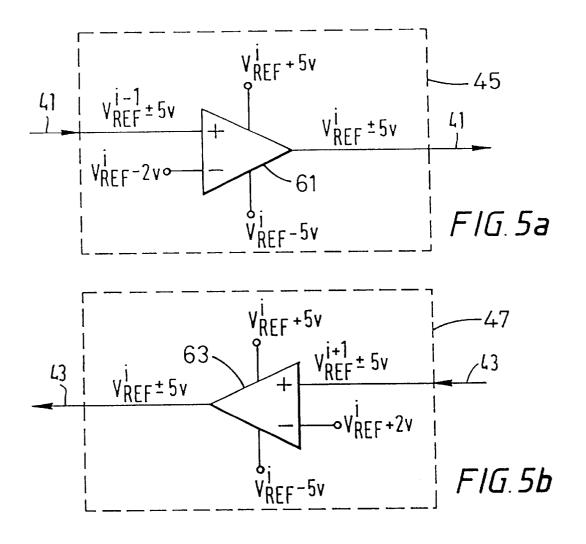
\* \* \* \* \*

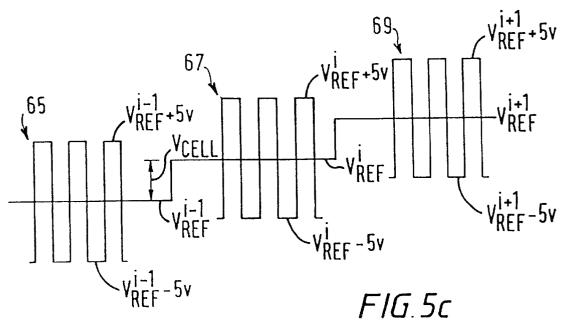
# IE2

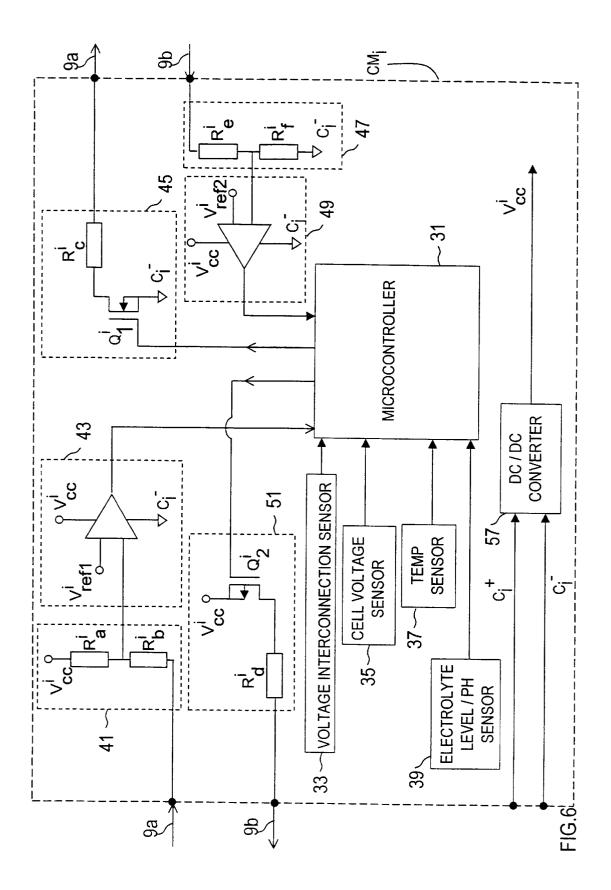












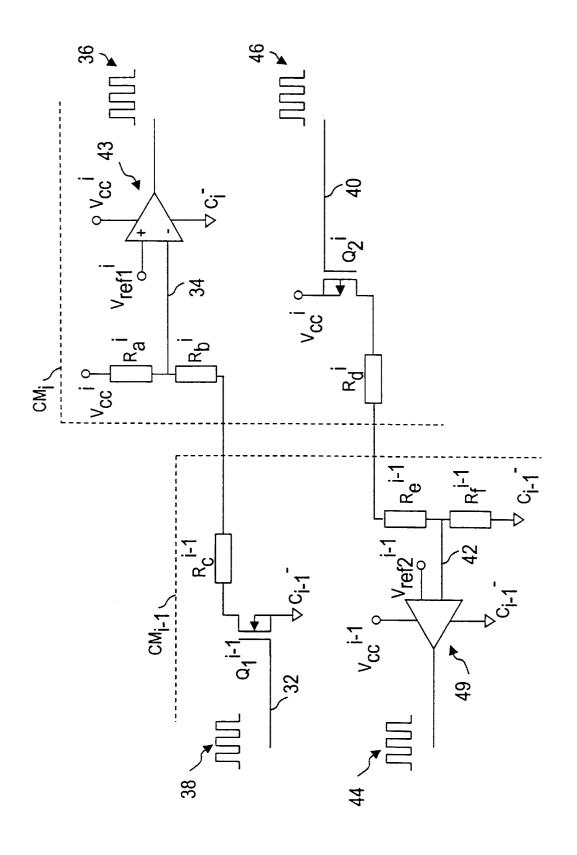
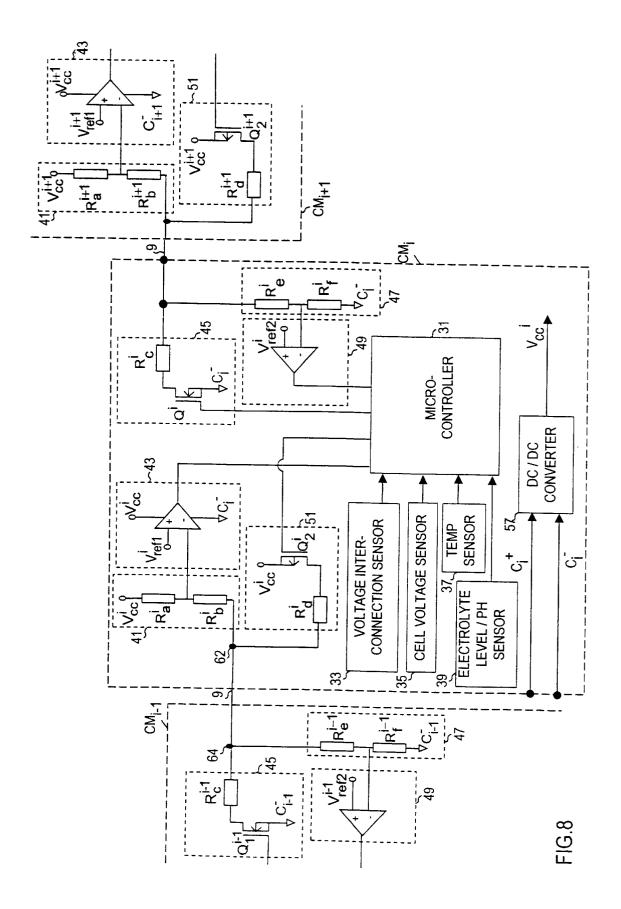


FIG.7



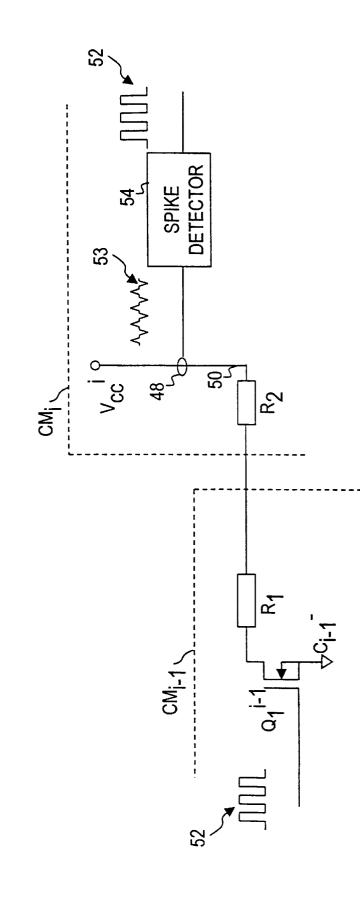
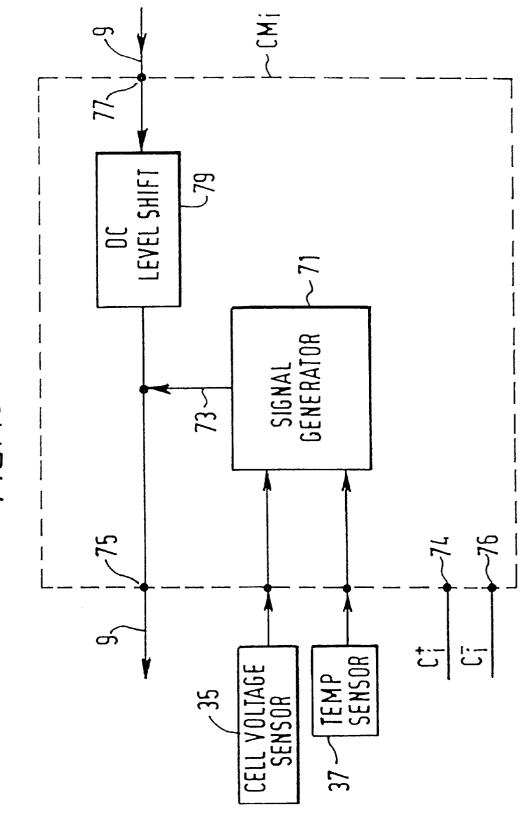
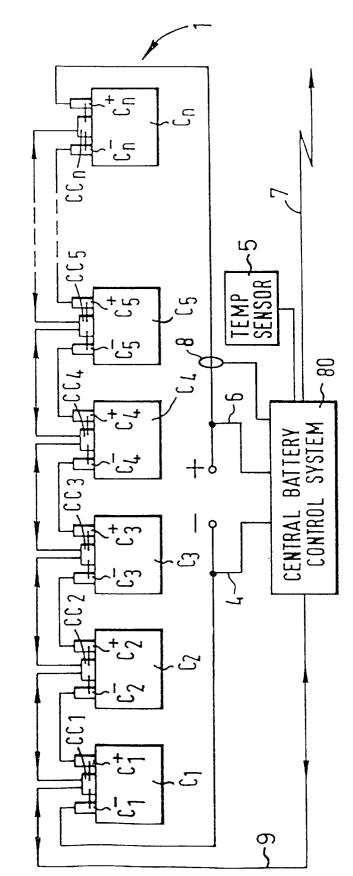


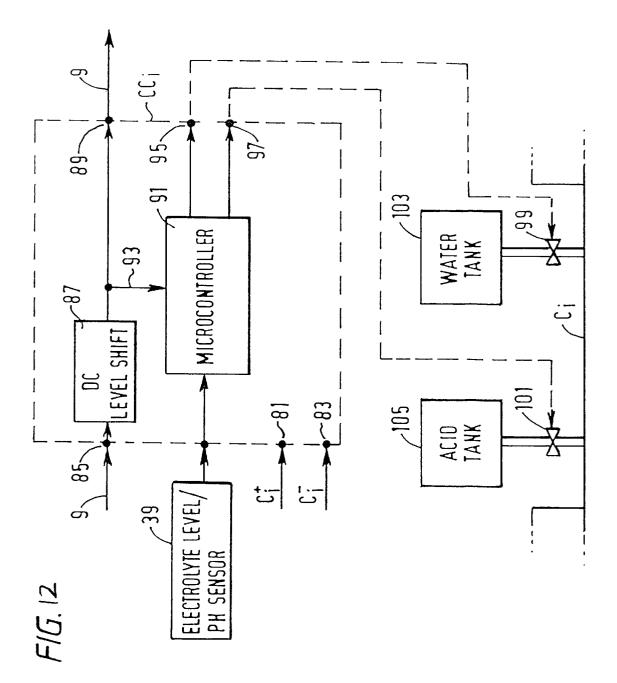
FIG.9

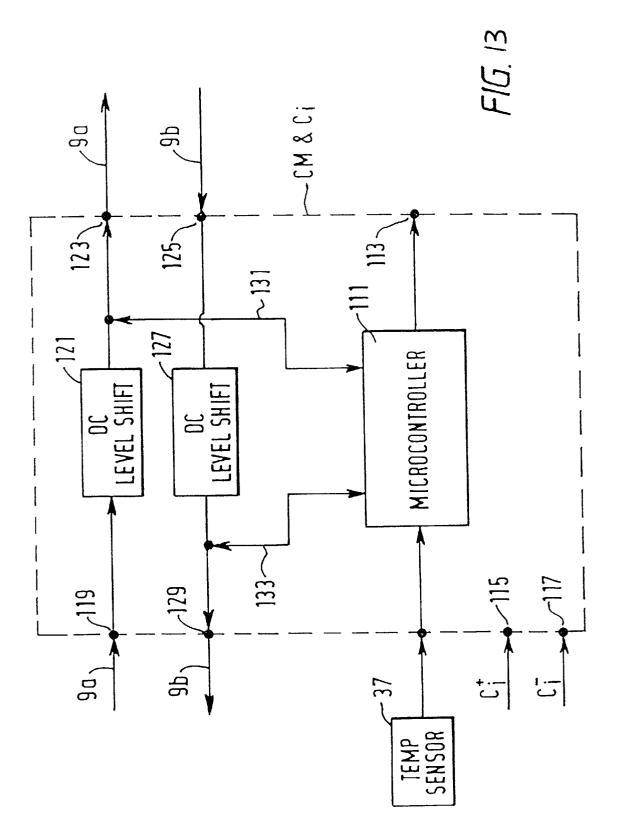


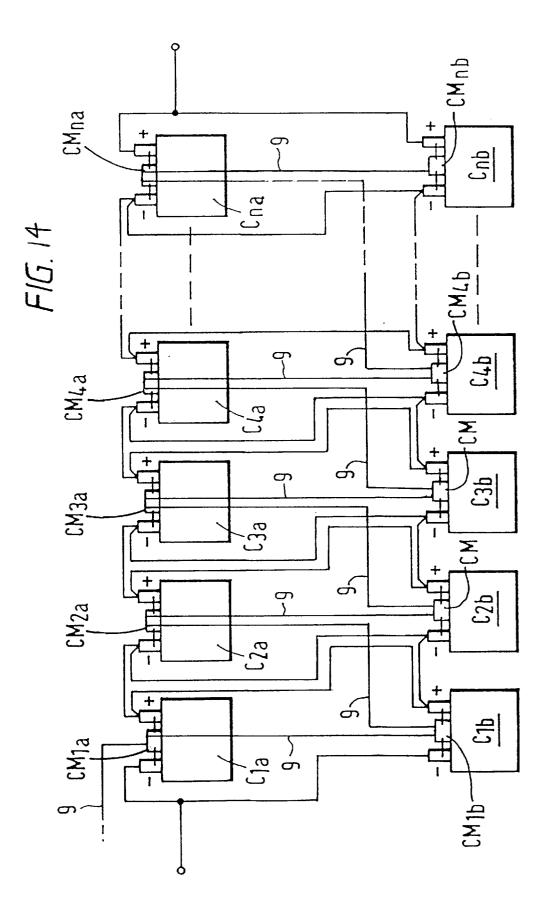
*FIG.* 10

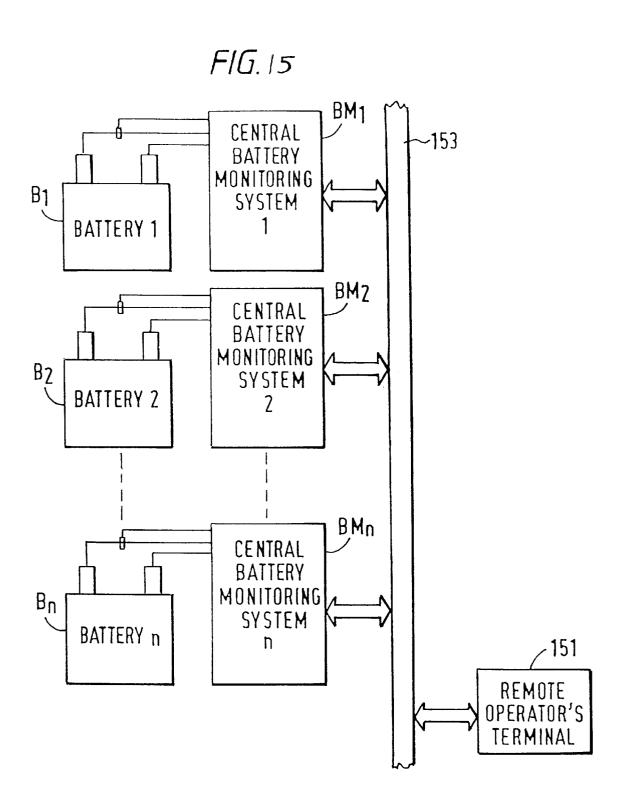


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### FIELD OF THE INVENTION

The present invention relates to a signalling system. The invention is applicable for use in a system for monitoring and/or controlling the cells of an industrial battery.

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### BACKGROUND OF THE INVENTION

Industrial batteries comprise a number of rechargeable battery cells which can be electrically connected in various series and series-parallel combinations to provide a 15 rechargeable battery having a desired output voltage. To recharge the battery, a current is passed through the cells in the opposite direction of current flow when the cells are working. There are many different types of battery cells available, but those most commonly used in industrial 20 applications are lead acid battery cells, each of which provides 2 volts, and nickel-cadmium (Nicad) battery cells, each of which provides 1.2 volts.

The batteries are usually used as a back-up power supply 25 for important systems in large industrial plants, such as off-shore oil rigs, power stations and the like. Since the batteries are provided as back-up in the event of a fault with the main generators, they must be constantly monitored and maintained so that they can provide power to the important systems for a preset minimum amount of time.

Many battery monitoring systems have been proposed which monitor the battery as a whole and provide an indication of the battery voltage. However, only a few systems have been proposed which can also monitor the individual cells which make up the battery. These systems use a number of monitoring devices, some of which are powered by the battery cell or cells which they monitor and send status information indicative of the cell voltage back to a central battery monitory system which monitors the battery as a whole.

However, since the cells are connected in series and since each cell monitoring device is powered by the cell which it is monitoring, the ground or reference voltage of each cell monitoring device is different. For example, in an industrial battery which has sixty lead acid cells connected in series, the negative terminal, i.e. the ground, of the fifth cell will be at a potential of approximately 8 volts and the positive terminal will be at a potential of approximately 10 volts, whereas the negative terminal of the seventh cell will be at a potetial of approximately 12 volts and the positive terminal will be at a potential of approximately 14 volts. This has lead to the common misconception in the art that the cell monitoring devices have to be electrically isolated from each other and from the central battery monitoring system.

In one known cell monitoring system, each cell is independently linked to its own electrically isolated input at the central monitoring system. The problem with this system is that a large number of connectors are needed to link the individual cell monitoring devices to the central monitoring system. Consequently, in practice, it is seldom used for permanent real-time monitoring of the battery cells.

In another known cell monitoring system, each cell monitoring device is serially linked to its neighbours in a daisychain configuration, either by using optical links between 65 the cell signalling device. the monitoring devices or by using transformers which have no DC path. The problem with this system is that to operate,

each of the cell monitoring devices requires either an electrical to optical and an optical to electrical converter or a modulator and a demodulator, which makes them relatively expensive and inefficient since this additional circuitry requires more power from the cell.

There is therefore a need to provide a simple cell monitoring device which can monitor and report on the status of the cells of the battery, but which consumes minimal power from the cell which it is monitoring.

As mentioned above, existing battery monitoring systems monitor the battery and provide an indication of the battery voltage. However, battery voltage is not an indication of the capacity of the battery, i.e. the ability of the battery to provide energy. There is therefore also a need to provide a battery monitoring system which can give the user a fairly accurate estimate of how much load he can place on a battery and over what period of time.

### SUMMARY OF THE INVENTION

The inventor has realised that it is possible to overcome the problem of having the cell monitoring devices operating at different voltages using simple electronic components and that therefore, there is no need for electrical isolation between the individual cell monitoring devices and the central monitoring system.

According to a first aspect, the present invention provides a signalling system for use with a plurality of series connected battery cells, comprising: a plurality of cell signalling devices, each to be powered by a respective one or more of the plurality of battery cells; and a communication link connecting the plurality of cell signalling devices in series; wherein each cell signalling device comprises a level shift circuit which is operable to receive signals transmitted from an adjacent cell signalling device to shift the level of the received signal and to output the level shifted signal for transmission to the communication link. By providing a level shift circuit in each cell signalling device, the cell signalling devices can be linked together in a communication link without the need for electrical isolation between the signalling devices.

The signalling system can be used as part of a battery monitoring and/or control system which is used to monitor and/or control the series connected battery cells. By providing the level shift circuit in each cell signalling device, the signalling system obviates the need for electrical isolation between individual cell signalling devices. Consequently, the communication link can be a simple one-wire communication bus.

Preferably each of the cell signalling devices is able to 50 receive communications from and transmit communications to the communication link so that they can communicate with, for example, the battery monitoring and/or control system. In which case, each cell signalling device can 55 comprise two DC level shift circuits, one for increasing the level of the received signals for transmission to a cell signalling device having a higher ground potential than that of the receiving cell signalling device, and one for reducing the level of the received signals for transmission to a cell signalling device which has a lower ground potential than 60 that of the receiving cell signalling device.

Each level shift circuit can comprise a simple electronic device, such as a comparator, which consumes a relatively small amount of power from the battery cell which powers

The first aspect of the present invention also provides a cell signalling device for use in the above defined signalling system, comprising: a power input terminal connectable to the cell or cells which is or are to power the cell signalling device; and at least one DC level shift circuit for receiving signals from an adjacent cell signalling device, for shifting the level of the received signal, and for outputting the level shifted signal for transmission to the communication link.

The first aspect of the present invention also provides a signalling kit comprising a plurality of the cell signalling devices defined above. The kit may also comprise the communication link for connecting the cell signalling 10 devices in series.

The first aspect of the present invention also provides a signalling method using a plurality of series connected battery cells, comprising the steps of: providing a plurality of cell signalling devices and powering them with a respec-15 tive one or more of the plurality of battery cells; providing a communication link which connects the plurality of cell signalling devices in series; receiving signals transmitted from an adjacent cell signalling device; shifting the level of the received signals; and outputting the level shifted signals 20 to the communication link.

#### BRIEF DESCRIPTION OF THE DRAWINGS

The present invention will now be described, by way of example only, with reference to the accompanying drawings,  $_{25}$ in which:

FIG. 1 schematically shows a battery comprising a number of battery cells connected in series, a central battery monitoring system for monitoring the condition of the battery as a whole and individual cell monitoring devices for monitoring the cells of the battery;

FIG. 2 is a schematic diagram showing more detail of the central battery monitoring system shown in FIG. 1;

FIG. 3 is a schematic diagram of one of the cell monitoring devices shown in FIG. 1;

FIG. 4 is a plot showing the battery-cell voltage distribution:

FIG. 5a is a circuit diagram of a first comparator forming part of the cell monitoring device shown in FIG. 3;

FIG. 5b is a circuit diagram of a second comparator forming part of the cell monitoring device shown in FIG. 3;

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FIG. 5c is a schematic representation showing part of the battery-cell staircase voltage distribution and example data pulses which are applied to the input of the comparators shown in FIGS. 5a and 5b;

FIG. 6 is a schematic diagram of one of the cell monitoring devices according to a second embodiment of the present invention:

FIG. 7 schematically illustrates the way in which signals 50 are passed between adjacent cell signalling devices in the second embodiment of the invention;

FIG. 8 schematically illustrates the form of a cell monitoring device according to a third embodiment and the way in which it is connected to neighbouring cell monitoring 55 battery cell monitoring device CM, mounted on top of the devices;

FIG. 9 schematically illustrates an alternative way in which signals can be transmitted between adjacent cell signalling devices:

FIG. 10 is a schematic diagram of a battery cell moni-60 toring device for use in a battery monitoring system according to a second embodiment of the present invention;

FIG. 11 schematically shows a battery comprising a number of battery cells connected in series, a central battery control system for controlling the battery as a whole and 65 individual battery cell controllers for controlling the cells of the battery;

FIG. 12 is a schematic diagram of one of the battery cell control devices shown in FIG. 11;

FIG. 13 is a schematic diagram of a battery cell monitoring and control device for use in a battery monitoring and control system embodying the present invention;

FIG. 14 is a schematic representation of an industrial battery in which the cells of the battery are connected in a series-parallel configuration; and

FIG. 15 is a schematic diagram of a system for monitoring a plurality of industrial batteries.

### DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENT

A first embodiment of the present invention will now be described with reference to FIGS. 1 to 5. FIG. 1 schematically shows an industrial battery, generally indicated by reference numeral 1, comprising a number of lead acid battery cells  $C_1, C_2, C_3 \dots C_n$  connected so that the negative terminal  $C_i^-$  of cell  $C_i$  is connected to the positive terminal  $C_{i-1}^{+}$  of preceding cell  $C_{i-1}$  and the positive terminal  $C_i^{+}$  of cell  $C_i$  is connected to the negative terminal  $C_{i+1}^{-}$  of the succeeding cell  $C_{i+1}$ , whereby the negative terminal  $C_{i+1}^{-}$  of the first cell  $C_1$  is the negative terminal of the battery and the positive terminal  $C_n^+$  of the last cell  $C_n$  is the positive terminal of the battery. Since the battery cells are lead acid, they each provide approximately 2 volts and the voltage of the battery as a whole will be approximately 2n volts. For industrial applications a voltage of 120 volts is often required. Therefore, 60 series connected lead acid or 100 series connected Nicad battery cells would be required. Sometimes, each cell in the series connection is connected in parallel with one or more similar cells, so as to provide redundancy, so that the battery will not fail if a single cell fails. 35

FIG. 1 also shows a central battery monitoring system 3 which is powered by the battery 1 via connectors 4 and 6, which connect the central battery monitoring system 3 to the negative terminal  $C_1^-$  and the positive terminal  $C_n^+$  of the battery 1, respectively. The battery monitoring system 3 monitors the status of the industrial battery 1 as a whole, based on charging and discharging characteristics of the battery (determined by monitoring the battery voltage from connectors 4 and 6 and the current being drawn from or 45 supplied to the battery 1, which is sensed by current sensor 8, whilst the battery is being charged and subsequently discharged), the ambient temperature (input from temperature sensor 5) and on information relating to the efficiency characteristics of the battery cells (provided by the battery cell manufacturer). The monitoring results can be stored in the central battery monitoring system 3 or they can be transmitted to a remote user (not shown) via the telephone line 7.

Each of the battery cells  $C_i$ , shown in FIG. 1, also has a cell between its positive and negative terminals Ci+and  $C_i^{-1}$ respectively, which monitors the status of the cell C<sub>i</sub>. Each cell monitoring device CM, is powered by the cell C, which it monitors and communicates with the central battery monitoring system 3 via a simple one-wire communication link 9. The communication link 9 links the cell monitoring devices CM<sub>i</sub>. in series in a daisy chain configuration to the central battery monitoring system 3, so that communications from the central battery monitoring system 3 to the cell monitoring devices CM<sub>i</sub> pass from left to right along the communication link 9 and communications from the cell monitoring devices CM<sub>i</sub> to the central battery monitoring system 3 pass from right to left along the communication link 9. Each cell monitoring device  $CM_i$  has its own cell identification or address, which, in this embodiment, is set in advance using DIP-switches mounted in the device. This allows communications from the central battery monitoring system 3 to be directed to a specific cell monitoring device and allows the central battery monitoring system 3 to be able to identify the source of received communications.

The battery monitoring system shown in FIG. 1 operates in two modes. In the first mode, the central battery monitoring system 3 monitors the condition of the industrial battery 1 as a whole and polls each of the cell monitoring devices  $CM_i$  in turn. During this mode, each of the cell monitoring devices  $CM_i$  listens to communications from the central battery monitoring system 3 on the communication link 9 and responds when it identifies a communication directed to it. When polled, each cell monitoring device  $CM_i$ performs a number of tests on the corresponding battery cell  $C_i$  and returns the results of the tests back to the central battery monitoring system 3 via the communication link 9.

In the second mode of operation, the central battery monitoring system **3** listens for communications on the communication link **9** from the cell monitoring devices  $CM_i$  indicating that there is a faulty condition with one of the battery cells  $C_i$ . In this second mode of operation, each cell monitoring device  $CM_i$  continuously monitors the corresponding battery cell  $C_i$  and, upon detection of a faulty condition, checks that the communication link **9** is free and then sends an appropriate message back to the central battery monitoring system **3** via the communication link **9**.

FIG. 2 is a schematic diagram of the central battery monitoring system 3 shown in FIG. 1. As shown, the central battery monitoring system 3 comprises a CPU 11 for controlling the operation of the central battery monitoring system 3. The CPU 11 is connected, via data bus 12, to a main memory 13 where data from the input sensors is stored and where test programs are executed, to a display 15 which displays the battery's current status and to a mass storage unit 17 for storing the sensor data and the results of the battery tests. The mass storage unit 17 can be fixed within the central battery monitoring system 3, but is preferably a floppy disk or a PCMIA memory card which can be withdrawn and input into an operator's personal computer for analysis. An operator can also retrieve the stored data and results and control the set up and initialisation of the central battery monitoring system 3 via the RS-232 serial interface 18. As mentioned above, instead of storing the test results in the mass storage unit 17, they can be transmitted via a modem 21 and telephone line 7 to a remote computer system (not shown) for display and/or analysis.

The central battery monitoring system measures the total battery capacity in Amp-hours (Ahr) or Watt-hours (Whr), the actual or remaining battery capacity as a percentage of the total battery capacity and the internal resistance of the  $_{55}$  battery 1 as a whole. The central battery monitoring system 3 can also measure the internal resistance of the individual cells from the data received from the individual cell monitoring devices CM<sub>i</sub> received via the communication link 9 and the communication circuit 19. 60

In order to be able to measure the total battery capacity, i.e. the maximum amount of charge which can be stored in the battery, and the actual or remaining battery capacity at a given time point as a percentage of the total battery capacity, the central battery monitoring system **3** monitors how much 65 charge is fed into the battery and how much charge is drawn from the battery. Unfortunately, since the charging and

discharging characteristics of the battery are not one hundred percent efficient. Therefore, the estimated capacity derived by monitoring the charge alone is not very accurate. In fact, various factors affect the amount of charge which is input to or drawn from a battery during charging/ discharging, including the ambient temperature, the magnitude of the charging/discharging current, the algorithm used for charging etc. Fortunately, many of these characteristics are known to the battery manufacturer and, in this embodiment the specific characteristics of the battery 1 are programmed into the central battery monitoring system 3. With this information, it is possible to determine more accurately how much charge has been stored in or withdrawn from the battery 1.

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For example, if the battery 1 is charged with a charging current of 10 amps over a period of two hours at an ambient temperature of 20° C., and it is known that the efficiency characteristic of the battery is 95% for such a level of charging current and for that ambient temperature, then the total charge supplied to the battery is 19 Ahr. In the general case, for a current (I(t)) drawn from or supplied to the battery, the capacity (CP) added to or removed from (depending on whether the current is negative or positive) the battery from time  $t_0$  to time  $t_1$  is given by:

$$CP[t_0, t_1] = (k_1 \times k_2 \times k_3) \cdot \int_{t_0}^{t_1} I(t) dt$$
<sup>(1)</sup>

where  $k_i$  are the efficiency characteristics for the battery 1 for the sensed conditions and where I(t) is negative when the current is being drawn from the battery 1.

In order to determine the initial total battery capacity (TCP), the battery 1 is initially fully charged by charging the battery for a long period of time using a small charging current. Then the battery 1 is discharged through a load (not shown) until the battery voltage drops below an end of discharge voltage limit (EODV) which is specified by the battery manufacturer. During this discharging period, the central battery monitoring system 3 monitors the discharge current via current sensor 8, and once the EODV limit is reached, it calculates the capacity (in Amp-hours) which has been removed from the battery using equation 1 above, with to being the time that the discharge is initiated and time t, is the time that the total battery capacity (TCP).

In this embodiment, the central battery monitoring system 3 periodically determines the remaining battery capacity (RCP) as a percentage of the total battery capacity (TCP) by monitoring the amount of current which is drawn from and/or supplied to the battery 1 since the last time the remaining battery capacity was determined and then by using the following equation:

$$RCP[t_1] = RCP[t_0] + \frac{100 \cdot CP[t_0, t_1]}{TCP}$$
(2)

Where  $CP[t_0,t_1]$  is calculated using equation 1 above. The initial estimate for the remaining battery capacity is set equal to the total working capacity of the battery after the battery 60 has been fully charged.

To determine the internal resistance of the battery as a whole, the battery is connected to two different loads and the central battery monitoring system **3** monitors the current through the loads from which it determines the internal resistance of the whole battery.

As mentioned above, in addition to determining the total battery capacity, the remaining battery capacity and the battery internal resistance, the central battery monitoring system 3 also monitors data received from the cell monitoring devices  $CM_i$  via the communication circuit 19 and the communication link 9. If there is a fault with one of the battery cells  $C_i$  or if there is some other faulty condition, the CPU 11 can trigger a local alarm 23 to alert a technician that there is a fault with the battery 1 or with one or more of the battery cells Ci. In this embodiment, the conditions which define a fault and their thresholds are user definable and set in advance.

Although the central battery monitoring system 3 continuously monitors the battery 1, the sensor data and the other battery data, i.e. the remaining battery capacity etc., are only stored periodically in the mass storage unit 17 in order to save storage space. The period is specified in 15 advance by the user and in this embodiment is set at ten seconds. Furthermore, when the samples are stored, they are time and date stamped so that the battery charging and discharging behaviour can be monitored and used to detect the cause of an eventual battery failure. In this embodiment, 20 the data which is to be stored is also filtered in order to try to identify and highlight important events, and the filtered data is also stored in the mass storage unit 17. What counts as an important event is user definable, but can be, for instance, a temperature increase of 2° C. or a change in 25 remaining battery capacity of greater than 1% of the total battery capacity.

As mentioned above, the status data of the battery, i.e. the battery voltage, the discharge/charge current, the battery temperature and the remaining and total battery capacities, -30 are displayed on display 15. For simplicity, since the display 15 does not need to be continuously updated, it is only updated using the samples of the status data which are to be stored in mass storage unit 17. Therefore, in this embodiment, the display 15 is updated every ten seconds.

In this embodiment, the central battery monitoring system 3 is also used to control the battery charger (not shown) which is used to charge the battery 1. In particular, the central battery monitoring system 3 monitors the charging current, the remaining battery capacity, the ambient temperature etc. and controls the operation of the charger (not shown) so that the battery charging is in accordance with the specific charging procedures recommended by the battery manufacturer for the battery 1.

Since the total battery capacity also decreases with time 45 (due to ageing), the central battery monitoring system 3 is programmed to perform regular (for example daily or monthly) automated measurements of the total battery capacity and the battery internal resistance using the procedures outlined above. This allows the central battery moni-50 toring system 3 to be able to build up a picture of the battery life characteristics and to be able to predict the battery end of life and the early detection of faulty conditions.

FIG. 3 is a schematic diagram showing, in more detail, one of the cell monitoring devices CM. As shown, cell 55 monitoring device  $CM_i$  comprises a microcontroller 31 for controlling the operation of the cell monitoring device CM<sub>i</sub> and for analysing sensor data received from voltage interconnection sensor 33, cell voltage sensor 35, temperature sensor 37 and electrolyte level/PH sensor 39.

The voltage interconnection sensor 33 measures the voltage drop between the cell being monitored and its neighbouring cells, by measuring the potential difference between each terminal of the cell C<sub>i</sub> and the respective terminal connections which connects cell  $C_i$  with its neighbouring 65 cells. Ideally, there should be no voltage drop between each terminal and the corresponding terminal connection.

However, due to chemical deposits accumulating at the cell terminals with time, or because of cell malfunction, a difference in potential between the cell terminals and the corresponding connectors sometimes exists, indicating that there is a fault, either with the battery cell  $C_i$  or with the interconnection with a neighbouring cell.

The cell voltage sensor 35 is provided for sensing the potential difference between the positive terminal  $C_i^+$  and the negative terminal  $C_i^-$  of the cell  $C_i$  which it is monitor-10 ing. The temperature sensor 37 senses the cell temperature locally at the cell C<sub>i</sub>. By monitoring the local temperature at each cell  $C_i$ , it is possible to identify quickly faulty cells or cells which are not operating efficiently. The electrolyte level/PH sensor senses the electrolyte level and/or the electrolyte PH of the battery cell  $C_i$  which it is monitoring.

The microcontroller **31** analyses the data input from the sensors and monitors for faulty conditions and reports to the central battery monitoring system 3 via the communication link 9. Since the microcontroller 31 processes digital data, and since the signals received from the sensors and the messages received from the battery monitoring system 3 are analogue signals, the microcontroller 31 has a built-in analogue to digital convertor (not shown) so that it can convert the sensor data and the received messages into corresponding digital signals.

Since the cell monitoring devices are connected in series by the communication link 9, each cell monitoring device CM, will either receive communications originating from the central battery monitoring system 3, from the left hand side of the communication link 9 for transmission to the next cell monitoring device  $CM_{i+1}$ , or they will receive communications from cell monitoring device  $CM_{i+1}$  from the right hand side of the communication link 9 for transmission back to the central battery monitoring system 3. In order to compensate for the difference in reference voltages between each of the cell monitoring devices CM<sub>i</sub>, each cell monitoring device CM, has an up-link 41 for transmitting data received from cell monitoring device  $CM_{i-1}$  to cell monitoring device  $CM_{i+1}$ , and a down-link 43 for transmitting data received from cell monitoring device  $CM_{i+1}$  to cell monitoring device  $CM_{i-1}$ .

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The up-link 41 has a transceiver 45 for increasing the reference voltage of the data signal so that it can be received by the next cell monitoring device  $CM_{i+1}$ , while the downlink 43 has a transceiver 47 which reduces the reference voltage of the received data so that it can be received by the cell monitoring device  $CM_{i-1}$ . The up-link 41 and the down-link 43 are connected to the one wire communication link 9 via switches 49 and 51 which are controlled by microcontroller 31, as represented by arrows 52. The way in which the microcontroller 31 controls the position of the switches 49 and 51 for the above described two modes of operation will be apparent to those skilled in the art and will not be described here. The microcontroller 31 is connected to the up-link 41 by connection 53 so that it can listen for communications sent from the central battery monitoring system 3 which are directed to it. Similarly, the microcontroller 31 is connected to the down-link 43 by connection 55 so that the microcontroller **31** can send messages back to the 60 central battery monitoring system 3, either upon being polled or upon detection of a fault.

In order to power the cell monitoring device  $CM_i$ , the positive terminal  $C_i^+$  and the negative terminal  $C_i^-$  of cell  $C_i$ are connected to the input of a DC to DC convertor 57, which generates, relative to the ground or reference voltage  $V_{REF}^{1}$  of cell C<sub>i</sub> (which equals the voltage potential of the negative terminal  $C_i^-$  of cell  $C_i$ ) the voltages  $V_{REF}^{i} \pm 5V$ , which are used to power the microcontroller 31 and the transceivers 45 and 47.

FIG. 4 shows the voltage characteristic of the industrial battery showing each cell's terminal potential versus the cell's position in the series. As shown in FIG. 4, this voltage characteristic has a staircase shape, with each stair having a height equal to the voltage  $V_{CELL}$  of the respective battery cell  $C_i$ . Each cell monitoring device  $CM_i$  uses the fact that there is only a small difference between the reference voltages of adjacent cells and that therefore the transceivers **45** and **47** only have to increase or decrease the reference voltage of the received data by this voltage difference.

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In this embodiment, the transceivers 45 and 47 comprise voltage comparators and the messages transmitted to and from the central battery monitoring system 3 are encoded 15 within the transitions of a square wave signal. FIG. 5*a* is a circuit diagram of a voltage comparator 61 forming part of the transceiver 45 provided in the up-link 41 shown in FIG. 3. The limits of the comparator 61 are  $V_{REF}^{i}$ +5V and  $V_{REF}^{i}$ -5V, which are generated by the DC to DC converter 20 57. FIG. 5*b* is a circuit diagram of a voltage comparator 63 forming part of the transceiver 47 provided in the down-link 43 shown in FIG. 3. As with comparator 61, the limits of comparator 63 are  $V_{REF}^{i}$ +5V and  $V_{REF}^{i}$ -5V.

FIG. 5c shows part of the battery-cell voltage distribution 25 shown in FIG. 4 and, superimposed thereon, data pulses for illustrating the way in which data is passed along the communication link 9. The left-hand side of FIG. 5c shows the ground or reference voltage  $V_{REF}^{i-1}$  for cell  $C_{i-1}$  and shows that data pulses **65** output by cell monitoring device 30  $CM_{i-1}$ , vary between  $V_{REF}^{i-1}$ +5V and  $V_{REF}^{i-1}$ -5V. In this embodiment, when the data is originally transmitted from the central battery monitoring system 3, the data pulses 65 will be transmitted from cell  $C_i^{-1}$  to cell  $C_i$  and will be applied to the positive input of the comparator 61 on the 35 up-link 41 of cell monitoring device  $CM_i$  via switch 49. As shown in FIG. 5a, the received pulses are compared with  $V_{REF}^{i}$ -2V (which is an approximation of the reference voltage  $V_{REF}^{i-1}$  of the cell  $C_i^{-1}$  which generated the received pulses 65, since the cells are lead acid battery cells which provide approximately 2 volts each) and the data pulses 67 output by comparator 61 will correspond with the received data pulses 65 but will vary between  $V_{REF}^{i}$ +5V and  $V_{REF}^{i}$ -5V, as shown in the middle of FIG. 5c. Therefore, the DC level of the square wave pulses has been increased by 45 passing it through the comparator 61.

The output data pulses 67 are transmitted to the next cell monitoring device  $CM_{i+1}$  via switch 51 and communications link 9. The data pulses 67 output from comparator 61 are also input to the microcontroller 31 via connection 53, so 50 that the microcontroller 31 can identify whether or not the communication from the central battery monitoring system 3 is directed to it. If the communication is directed to it, the microcontroller 31 processes the request, performs the necessary tests and transmits the appropriate data back to the 55 central battery monitoring system 3.

When data pulses **69** are transmitted to cell monitoring device  $CM_i$  from cell monitoring device  $CM_{i+1}$  for transmitting back to the central battery monitoring system **3**, the received data pulses **69**, which vary between  $V_{REF}^{i+1}+5V$  60 and  $V_{REF}^{i+1}-5V$ , are applied to the positive input of comparator **63** on the down-link **43** of cell monitoring device  $CM_i$  via switch **51**. As shown in FIG. **5***b*, the received pulses **69** are compared with  $V_{REF}^{i+2}V$  (which is an approximation of the reference voltage  $V_{REF}^{i+1}$  of the cell  $C_{i+1}$  which 65 generated the received pulses **69**, since the cells are lead acid battery cells which provide approximately 2 volts each). As

shown in FIG. 5*c*, this comparison results in a series of pulses 67 corresponding to the received pulses 65 but which vary between  $V_{REF}^{i}\pm5V$  which are transmitted to cell  $C_{i-1}$  via switch 49. Therefore, the DC level of the square wave pulses has been reduced by passing it through the comparator 63.

Each of the cell monitoring devices  $CM_i$  operate in a similar manner. However, it should be noted that the first cell monitoring device  $CM_i$  has the same ground or reference voltage as the central battery monitoring system **3**. Therefore, it is not necessary to use a transceiver **45** in the up-link **41** of the first cell monitoring device  $CM_i$ , although one is usually used in order to buffer the received signals and in order to standardise each of the cell monitoring devices  $CM_i$ . Similarly, the last cell monitoring device  $CM_n$  will not receive data pulses from a subsequent cell monitoring device and therefore, does not need a transceiver **47** in its downlink. However, one is provided so that all the cell monitoring devices  $CM_i$  are the same, and is used for buffering the data sent from microcontroller **31** of cell monitoring device  $CM_n$  back to the central battery monitoring device **3**.

The battery monitoring system described above has the following advantages:

(1) There is no need for voltage isolation between the cell monitoring devices  $CM_i$  or between the first cell monitoring device  $CM_1$  and the central battery monitoring system **3**. Therefore, each cell monitoring device  $CM_i$  will only consume a few milli-amps and only requires very inexpensive and readily available DC to DC converters for converting the battery cell voltage to the supply voltage needed by the microcontroller **31** and the transceivers **45** and **47**.

(2) Since electrical isolation is not required between the cell monitoring devices  $CM_i$ , there is no longer a need for relatively expensive voltage isolated links between the cell monitoring devices. In the embodiment described, each cell monitoring device  $CM_i$  is linked to its neighbours by a simple wire. The cost of the battery monitoring system is therefore low and system installation is simplified.

(3) Continuous monitoring of all the cells C<sub>i</sub> in battery 1
40 becomes economical and practical, and the user can be informed in real-time if one or more of the battery cells C<sub>i</sub> is under performing or is faulty.

(4) The internal resistance of each cell  $C_i$  can be determined in real-time and without having to disconnect the cell from the battery, since the central battery monitoring system **3** is capable of measuring battery charging and discharging current (which is the same as the cell current) and can correlate it with individual cell voltages (determined by the cell monitoring devices) in order to calculate each cell's internal resistance.

(5) Each cell monitoring device  $CM_i$  is able to measure the voltage drop on cell to cell interconnections and indicate a faulty interconnection condition, usually due to chemical deposits accumulating at the cell terminals with time or because of cell malfunction.

(6) Since each cell monitoring device  $CM_i$  is able to measure the cell voltage and the cell temperature, it is possible to increase the probability of detecting a faulty cell. Therefore, the industrial battery need only be serviced when required.

(7) Since each cell monitoring device  $CM_i$  can read the corresponding cell voltage, cell temperature etc. at the same time as the other cell monitoring devices, the data produced by each cell monitoring device is less likely to be corrupted by changes in load and/or changes in ambient temperature which occur with time, as compared with prior art systems which take readings from the individual cells one at a time.

A number of alternative embodiments will now be described, which operate in a similar manner to the first embodiment. Accordingly, the description of these alternative embodiments will be restricted to features which are different to those of the first embodiment.

FIG. 6 is a schematic diagram showing, in more detail, one of the cell monitoring devices  $CM_i$  shown in FIG. 1 according to a second embodiment of the present invention. As shown, cell monitoring device CM<sub>i</sub> comprises a microcontroller **31** for controlling the operation of the cell monitoring device CM<sub>i</sub> and for analysing sensor data received from voltage interconnection sensor 33, cell voltage sensor 35, temperature sensor 37 and electrolyte level/PH sensor 39, which all operate in the same manner as in the first embodiment.

15 In order to power the cell monitoring device CM<sub>i</sub>, the positive terminal  $C_i^+$  and the negative terminal  $C_i^-$  of cell  $C_i$ are connected to the input of a DC to DC convertor 57, which generates, relative to the ground of cell C, (which equals the voltage potential of the negative terminal  $C_i^-$  of cell  $C_i$ ) the voltage  $V_{cc}^{i}$ , which is used to power the 20 microcontroller 31 and the other components in the device and which in this embodiment is  $C_i^{-}+3$  Volts.

Since the cell monitoring devices are connected in series by the communication link 9, each cell monitoring device CM<sub>i</sub> is operable (i) to receive up-link communications 25 originating from the central battery monitoring system 3 on wire 9a of the communication link 9 for reception by itself and/or for onward transmission to the next cell monitoring device  $CM_{i+1}$ ; (ii) to receive down-link communications from cell monitoring device  $CM_{i+1}$  on wire 9b of the 30 communication link 9 for transmission back to the central battery monitoring system 3; and (iii) to transmit down-link communications generated by itself back to the central battery monitoring system 3 on wire 9b of the communication link 9.

As shown in FIG. 6, in this embodiment, the microcontroller 31 receives up-link communications originating from the central battery monitoring system 3 via wire 9a, potential divider 41 and comparator 43. The microcontroller 31 identifies whether or not the received message from the central battery monitoring system is for it or if it is for onward transmission to the next cell monitoring device  $CM_{i+1}$ . If the message is for the current cell monitoring device  $CM_i$ , then the microcontroller 31 decodes the message and takes the appropriate action. If the received mes-45 sage is for onward transmission, then the microcontroller 31 regenerates the message and outputs it to wire 9a via output block 45. In this embodiment, the messages transmitted are square-wave signals representing digital data. The signals are encoded for error correction purposes and the microcontroller 31 checks for errors in the received messages. Since the microcontroller 31 regenerates the messages for transmission to the next cell monitoring device  $CM_{i+1}$ , the timing between the transitions in the signal levels can be resynchronised, thereby reducing any errors caused by the transmission of the pulses along the communication link 9.

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In a similar manner, down-link messages received from cell monitoring device  $CM_{i+1}$  on wire 9b are passed via potential divider 47 and comparator 49 to the microcontroller 31. In this embodiment, all down-link communications 60 are for transmission back to the central battery monitoring system 3. Therefore, the micro controller 31 checks for any errors in the received data and, if possible, corrects them. The microprocessor 31 then regenerates the message and outputs it to wire 9b via output block 51 for onward 65 transmission back to the central battery monitoring system 3.

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The way in which the messages are transmitted between the cell monitoring devices will now be described in more detail with reference to FIG. 7. Up-link messages originating from the central battery monitoring system 3 which are re-transmitted by the microcontroller 31 of cell monitoring device  $CM_{i-1}$  are applied on line 32 to the gate electrode of the MOSFET  $Q_1^{i-1}$ . The source electrode of the MOSFET  $Q_1^{i-1}$  is connected to the ground  $C_{i-1}^{--1}$  of cell monitoring device  $CM_{i-1}$  and the drain electrode is connected, via 10 resistors  $R_a^i$ ,  $R_b^i$  and  $R_c^{i-1}$ , to  $V_{cc}^i$  output by the DC/DC converter 57 in cell monitoring device CM<sub>i</sub>. In operation, when the microcontroller **31** in the cell monitoring device  $CM_{i-1}$  outputs a voltage low (representing a binary 0) on line **32**, the MOSFET  $Q_1^{i-1}$  does not allow current to flow from the drain electrode to the source electrode and therefore effectively open circuits the connection between  $V_{i}^{i}$ and  $C_{i-1}$ . Therefore, when a voltage low is applied to the gate electrode of MOSFET  $Q_1^{i-1}$ , a voltage of approximately  $V_{cc}^{i}$  is applied on line **34** to the comparator **43**, where it is compared with reference voltage  $V_{ref1}^{i}$ . When the microprocessor **31** of cell monitoring device  $CM_{i-1}$  outputs a voltage high (representing a binary 1) on line 32, the MOSFET  $Q_1^{i-1}$  is switched on and current can flow from the drain electrode to the source electrode. Therefore, current will flow from  $V_{cc}^{i}$  through resistors  $R_a^{i}$ ,  $R_b^{i}$  and  $R_c^{i-1}$ through the MOSFET  $Q_1^{i-1}$  to the ground  $C_{i-1}^{-}$  of cell monitoring device  $CM_{i-1}$ . As a result,  $V_{cc}^{i}$ -X volts will be applied on line 34 to the comparator 43, where it is comapplied on line 34 to the comparator 43, where it is com-pared with the reference voltage  $V_{ref1}^{i}$ . The value of X depends upon the difference between  $V_{cc}^{i}$  and  $C_{i-1}^{-}$  and the values of the resistors  $R_a^{i}$ ,  $R_b^{i}$  and  $R_c^{i-1}$ . Provided the value of the reference voltage  $V_{ref1}^{i}$  is between the voltage levels applied to the comparator 43 on line 34 when the MOSFET  $Q_1^{1_{i-1}}$  is switched on and when it is switched off, the output 35 of the comparator 43 will be a square-wave signal 36varying between the ground potential  $C_i^-$  and  $V_{cc}^{-i}$  of cell monitoring device CM, in synchronism with the variation of the square-wave signal 38 applied to the gate electrode of MOSFET  $Q_1^{i-1}$  of cell monitoring device  $CM_{i-1}$ . Therefore, messages encoded within the variation of the signal applied to the MOSFET  $Q_1^{i-1}$  in cell monitoring device  $CM_{i-1}^{11}$  are transferred from cell monitoring device  $CM_{i-1}$  to cell monitoring device CM<sub>i</sub>.

In a similar manner, down-link messages output by the microcontroller 31 in cell monitoring device CM, for transmission back to the central battery monitoring system 3 are applied to the gate electrode of MOSFET  $Q_2^{i}$  on line 40. The drain electrode of MOSFET  $Q_2^{i}$  is raised to the potential  $V_{ce}^{i}$  and the source is connected, via resistors  $R_d^{i}$ ,  $R_e^{i-1}$  and  $R_f^{i-1}$ , to the ground potential  $C_{i-1}^{-1}$  of cell monitoring device 50  $CM_{i-1}$ . In operation, when the microcontroller **31** in the cell monitoring device CM, outputs a voltage low on line 40, the MOSFET  $Q_2^{\prime}$  does not allow current to flow from the drain electrode to the source electrode and therefore effectively open circuits the connection between  $V_{cc}^{i}$  and  $C_{i-1}^{-}$ . Therefore, when a voltage low is applied to the gate electrode of MOSFET  $Q_2^{i}$ , approximately zero volts is applied on line 42 to the comparator 49, where it is compared with reference voltage  $V_{ref2}^{i-1}$  When the microcontroller **31** of cell monitoring device CM<sub>i</sub> outputs a voltage high on line **40**, the MOSFET  $Q_2^{i}$  is switched on and current flows from  $V_{cc}^{i}$  through resistors  $R_d^{i}$ ,  $R_e^{i-1}$  and  $R_f^{i-1}$  to the ground  $C_{i-1}^{i-1}$  of the cell monitoring device  $CM_{i-1}$ . As a result,  $V_{cc}^{i}$ -Y volts will be applied on line 42 to the comparator 49, where it is compared with the reference voltage  $V_{ref2}^{i-1}$ . The value of Y depends upon the difference between  $V_{cc}^{i}$  and  $C_{i-1}^{-}$  and the values of resistors  $R_d^{i}$ ,  $R_e^{i-1}$  and  $R_f^{i-1}$ . Again, provided

the value of the reference voltage  $V_{ref2}^{ii-1}$  is between the voltage levels applied to the comparator **49** on line **42** when the MOSFET  $Q_2^{i}$  is switched on and when it is switched off, the output of the comparator 49 will be a square-wave signal 44 varying between the ground potential  $C_{i-1}^{--1}$  and  $V_{cc}^{i-1}$  of cell monitoring device  $CM_{i-1}$  in synchronism with the variation of the signal 46 applied to the gate electrode of MOSFET  $Q_2^{i}$  of cell monitoring device  $CM_i$ . Therefore, messages encoded within the variation of the signal applied to MOSFET  $Q_2^{i}$  are transferred from cell monitoring device  $CM_i$  to cell monitoring device  $CM_{i-1}$ .

The values of the resistors  $R_a^{i}$  to  $R_f^{i}$  in each cell monitoring device  $CM_i$  are chosen in order (i) to buffer the input and output terminals of the cell monitoring devices CM<sub>i</sub>; (ii) to reduce power consumption of the cell monitoring devices CM;; and (iii) to provide the necessary voltage division with respect to the difference in voltage between adjacent cell monitoring devices.

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As those skilled in the art will appreciate, the above technique for transferring up-link and down-link data 20 between cell monitoring devices  $\mathrm{CM}_{i-1}$  and  $\mathrm{CM}_i$  will only work provided the difference between  $V_{cc}^{i}$  and  $C_{i-1}^{-}$  does not exceed the operating range of the MOSFET switches  $Q_1^{i-1}$  and  $Q_2^{i}$ . In this embodiment, each battery cell  $C_i$  is provided with a cell monitoring device CM<sub>i</sub>, the difference 25 in operating potentials of adjacent cell monitoring devices is approximately two volts and  $V_{cc}^{i}$  is three volts more than the ground potential of the cell monitoring device. Therefore, in this embodiment, the difference between  $V_{cc}^{i}$  and  $C_{i-1}^{-}$  is approximately five volts. MOSFET transistors Q which can 30 operate with such a loading are readily available. Indeed, there are some commercially available MOSFETs which can operate with a loading of up to sixty volts. Therefore, this technique of transmitting data between adjacent cell monitoring devices can be used in most practical situations, even 35 in an embodiment where, for example, one cell monitoring device is provided for every tenth battery cell C<sub>i</sub>.

Each of the cell monitoring devices  $CM_i$  operate in a similar manner. However, it should be noted, that in this embodiment, the first cell monitoring device CM<sub>1</sub> has the 40 same ground or reference voltage as the central battery monitoring system 3. Therefore, in this embodiment, it is not necessary to use the potential divider 41 and the comparator 43 of the up-link nor the output block 51 in the down-link of the first cell monitoring device CM<sub>1</sub>, although these are 45 usually provided in order to standardise each of the cell monitoring devices CM<sub>i</sub>. Similarly, the last cell monitoring device CM<sub>n</sub> will not transmit data to nor receive data from a subsequent cell monitoring device. Therefore, cell monitoring device CM<sub>n</sub> does not need the output block 45 in the up-link nor the potential divider 47 and the comparator 49 in the down-link. However, these are usually provided so that all the cell monitoring devices  $CM_i$  are the same.

As those skilled in the art will appreciate, in the above embodiment, solid state switches were used to, effectively, 55 shift the DC level of the received signals for onward transmission. FIG. 8 illustrates a further embodiment which uses analogue switches to transmit messages uplink as well as downlink between adjacent cell monitoring devices CM<sub>i</sub>, via a single communication wire 9 which connects the cell 60 monitoring devices in series. As shown in FIG. 8, each cell monitoring device  $CM_i$  has the same potential dividers 41 and 49, comparators 43 and 47, and output blocks 45 and 51 as in the second embodiment. The difference between this embodiment and the second embodiment is that the uplink 65 and the downlink between adjacent cell monitoring devices share a common communication wire 9. This is achieved by

connecting, at connection 62, the output of block 51 in cell monitoring device CM, to the connection between potential divider 41 in cell monitoring device  $CM_i$  and the output block 45 in cell monitoring device  $CM_{i-1}$  and by connecting, at connection 64, the potential divider 49 in cell monitoring device  $CM_{i-1}$  to the connection between potential divider 41 in cell monitoring device CM<sub>i</sub> and the output block 45 in cell monitoring device CM<sub>i-1</sub>.

Since both the uplink and the downlink connection 10 between adjacent cell monitoring devices is via the same wire 9, communications between cell monitoring devices can be in one direction only at any given time. To achieve this, during a downlink communication, MOSFET  $Q_1^{i-1}$  is switched off so that messages transmitted by switching the state of MOSFET  $Q_2^i$  pass via the potential divider 49 and the comparator 47 into the microcontroller (not shown) in cell monitoring device  $CM_{i-1}$ . Similarly, for uplink communications from, for example, cell monitoring device  $CM_i$  to cell monitoring device  $CM_{i+1}$ , the MOSFET  $Q_2^{i+1}$  is switched off so that the output block 51 has a high impedance and does not affect the operation of the potential divider 41. Further, as those skilled in the art will appreciate, in order for the potential dividers and output blocks to operate in the same way as in the second embodiment, the resistances of resistors R<sub>a</sub>, R<sub>b</sub>, R<sub>e</sub> and R<sub>f</sub> must be relatively large compared to the resistances of resistors  $R_c$  and  $R_d$ . In this embodiment, with 2 volt battery cell voltages,  $R_a = R_b = R_e$  $R_f = 10K$  ohms,  $R_d = R_c = 2K$  ohms,  $V_{ref1}^i = V_{ref2}^i = V_{cc}^i/2$  and  $c^{i} = C_{i}^{-} + 5V$ . As those skilled in the art will appreciate, the cell monitoring device of this embodiment can easily be adapted to operate with any battery cell voltage, the only changes that are required are the resistor values, the reference voltage values and the maximum allowable drain to source voltage of the MOSFETs.

As those skilled in the art will appreciate, in the above embodiments, data was transmitted between adjacent cell monitoring devices by varying the output impedance of an output block in the cell monitoring device which is to transmit the message and by detecting this variation in the cell monitoring device which is to receive the message. For example, when up-link message data is to be transmitted from cell monitoring device  $CM_{i-1}$  to cell monitoring device CM<sub>i</sub>, the impedance of output block 45 in cell monitoring device  $CM_{i-1}$  is varied in dependence upon the data to be transmitted-when a voltage low is to be transmitted, the impedance of output block 45 is made to be very high, whereas when a voltage high is to be transmitted, the impedance of output block 45 is made to be relatively low. This variation of the impedance of output block 45 is detected by the potential divider 41 and the comparator 43 in the cell monitoring device  $CM_i$  which is to receive the transmitted up-link message. As those skilled in the art will appreciate, there are various ways of varying an output impedance of a cell monitoring device in dependence upon the data to be transmitted and various ways of detecting that variation in the adjacent cell monitoring device. FIG. 9 shows one of these alternative embodiments.

In the embodiment shown in FIG. 9, the output impedance of cell monitoring device  $CM_{i-1}$  is varied in the same way as it was varied in the second embodiment but this variation is detected in cell monitoring device CM<sub>1</sub> in a different manner. In particular, in this embodiment, a current detector **48** is used to detect the changes in current flowing between the  $V_{cc}^{i}$  terminal in cell monitoring device CM<sub>i</sub> to the ground potential  $C_{i-1}^{-}$  in cell monitoring device  $CM_{i-1}$ . In operation, when a voltage low is applied to the gate elec-trode of MOSFET  $Q_1^{i-1}$ , the MOSFET is open and no current flows down line 50. However, when a voltage high is applied to the gate electrode of MOSFET  $Q_1^{i}$ <sup>-1</sup>. the MOSFET opens and current is drawn down line 50. This change in current is detected by the current sensor 48. More specifically, each time there is a transition from a voltage high to a voltage low (or vice versa) applied to the gate electrode of MOSFET  $\hat{Q}_1^{i-1}$ , a voltage spike is induced in the current detector 48. As illustrated in FIG. 9, in response to up-link message data 52 being applied to the MOSFET  $Q_1^{i-1}$ , a train of voltage spikes 53 are induced in the current 10 detector 48 and passed to a spike detector 54 which regenerates and outputs the up-link message data 52 for transmission to the microcontroller (not shown). The resistors  $R_1$ and  $R_2$  are provided in order to reduce the power consumed by each of the cell monitoring devices  $CM_i$  and in order to 15 buffer the input and output of the respective cell monitoring devices.

In the last three embodiments, MOSFET switches were used as a device whose impedance can be varied in depenskilled in the art will appreciate, these MOSFETs can be replaced by any electronic switches (solid-state relays, electromechanical relays, J-FETs, transistors etc.) and, in embodiments where the up-link and the down-link messages are received and re-transmitted by a microcontroller, can be omitted altogether. This is because the output impedance of the output pin of the microcontroller varies in dependence upon whether it is outputting a voltage high or a voltage low. Therefore, for example, the output pin from the microcontroller in cell monitoring device  $\overline{CM}_{i-1}$  could be directly -30 connected to the potential divider 41 in cell monitoring device CM<sub>i</sub>. However, such an embodiment is not preferred, because the microcontroller can be damaged by the voltage applied to the output pin from the adjacent cell monitoring device.

In the above embodiments, each cell monitoring device CM, has a microcontroller 31 for receiving messages from the central battery monitoring system 3, for analysing data from various sensors and for sending data back to the central battery monitoring system 3 via the communication link 9. FIG. 10 schematically shows an alternative cell monitoring device CM, of an alternative embodiment which does not use a microcontroller 31.

In particular, as shown in FIG. 10, each cell monitoring device  $CM_i$  comprises a signal generator 71 which receives 45 sensor signals from the cell voltage sensor 35 and the temperature sensor 37 and outputs, on line 73, a signal which varies in dependence upon the received sensor signals. The signal generator 71 may comprise a voltage controlled oscillator which outputs an alternating signal whose frequency varies in dependence upon an input voltage from, for example, the cell voltage sensor 35. The signal output from the signal generator 71 is applied to an output terminal 75 for transmission to the central battery monitoring system 3, via the communication link 9. In this embodiment, each cell monitoring device CM, only transmits signals back to the central battery monitoring system 3, they can not receive messages from the central battery monitoring system. Therefore, only a down-link is required to receive signals at input terminal 77, transmitted from cell monitoring device 60  $CM_{i+1}$ .

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As in the first embodiment, each cell monitoring device  $CM_i$  is powered by the cell  $C_i$  which it is monitoring. This is illustrated in FIG. 6 by the connections  $C_i^+$  and  $C_i^-$  which are connected to input terminals 74 and 76 respectively. Since the communication link 9 connects each of the cell monitoring devices CM, in series in a daisy chain

configuration, cell monitoring device CM<sub>i</sub> will receive signals, at input terminal 77, from cell monitoring device  $CM_{i+1}$ . The received signals are applied to a DC level shift circuit 79 which reduces the DC level of the received signals and supplies them to the output terminal 75 for transmission to the next cell monitoring device  $CM_{i-1}$  in the communication link 9.

In the above embodiments, the system described was a battery monitoring system. FIG. 11 schematically shows an embodiment which is a control system for controlling the cells of an industrial battery. As shown, the control system has a similar architecture to the battery monitoring system shown in FIG. 1, except that the central battery monitoring system 3 is now a central battery control system 80 and the cell monitoring devices CM<sub>i</sub> are now battery cell control devices  $CC_i$ . As in the monitoring system of FIG. 1, the central battery control system 80 communicates with each of the cell controlling devices CC<sub>i</sub> via the communication link

FIG. 12 schematically shows an example of one of the dence upon the message data to be transmitted. As those 20 battery cell control devices  $CC_i$  shown in FIG. 11. In this example, each cell controlling device  $CC_i$  is used to control the topping up of acid and water in the respective battery cell C<sub>i</sub>, in response to an appropriate control signal received from the central battery control system 80. As in the above embodiments, each cell control device  $CC_i$  is powered by 25 the cell which it is to control, as represented by inputs  $C_i$ and  $C_i^-$  applied to input power terminals 81 and 85 respectively. In this embodiment, each cell controlling device CC<sub>e</sub> is arranged to receive messages from the central battery controlling system (not shown), but not to transmit messages back. Accordingly, signals received at the input terminal 85 from cell controller  $CC_{i-1}$  are applied to DC level shift circuit 87, which-increases the DC level of the received signals and outputs them to output terminal 89 for transmis-35 sion to the next cell controlling device  $CC_{i+1}$ . The microcontroller 91 monitors the received signals via connection 93 and outputs appropriate control signals to output terminals 95 and 97 when the received signals are directed to it. The control signals output to terminals 95 and 97 are used to control the position of valves 99 and 101 respectively, so 40 as to control the amount of water and acid to be added to the battery cell  $C_i$  from the water tank 103 and the acid tank 105. The microcontroller 91 determines the amount of water and acid to add with reference to the sensor signals received from the electrolyte level/PH sensor 39.

In the above embodiments, a central battery monitoring system or a central battery control system was provided which monitored or controlled the system as a whole. FIG. 13 schematically shows a cell monitoring and control device CM&C, which can be used in a combined battery control and monitoring system in which there is no central battery monitoring and control system and in which each cell monitoring and control device  $CM\&C_i$  communicates directly with the other cell monitoring and control devices. As in the other embodiments, each cell monitoring and control device CM&C, is powered by the cell which it is monitoring and controlling, as represented by inputs  $C_i^+$  and  $C_i^-$  applied to input power terminals 115 and 117 respectively.

As shown in FIG. 13, each cell monitoring and control device CM&C, comprises a microcontroller 111 which receives sensor data from temperature sensor 37 and which outputs control data to output terminal 113 for controlling, for example, a liquid crystal display (not shown) mounted on 65 the respective cell  $C_i$ .

In this embodiment, the communication link comprises two wires 9a and 9b and therefore, switches 49 and 51 are not required to connect the up-link and the down-link to the communication link 9. Wire 9a is used for passing communications up the series communication link 9 from cell monitoring and control device CM&C<sub>i</sub> to cell monitoring and control device  $CM\&C_{i+1}$  and wire 9b is used for transmitting signals down the series communication link 9 from cell monitoring and control device CM&C<sub>i</sub> to cell monitoring and control device  $CM\&C_{i-1}$ . Accordingly, the signals received by cell monitoring and control device CM&C, at input terminal **119** are applied to DC level shift circuit 121 which increases the DC level of the received signals and outputs them to output terminal 123 for transmission to cell monitoring and control device  $CM\&C_{i+1}$ . Similarly, signals received at input terminal 125 are applied to DC level shift circuit 127 which decreases the DC level of the received signals and outputs them to output terminal 129 for transmission to cell monitoring and control device  $CM\&C_{i-1}$ . As shown, microcontroller 111 can receive data from and transmit data to both the up-link 9a and the down-link 9b via connections 131 and 133 respectively.

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Various modifications which can be made to the above described embodiments will now be described.

In the first embodiment, the transceivers 45 and 47 used in the up-link and the down-link within each cell monitoring device CM<sub>i</sub> comprises a voltage comparator. Other types of transceivers could be used. For example, voltage to current and current to voltage comparators could be used. In such an embodiment, the voltage to current comparators and the current to voltage comparators would be arranged alternatively along the communication link 9 so that a voltage to 30 current comparator is connected to the input of a current to voltage comparator, and vice-versa.

In the first embodiment the data transmitted between cells and between the first cell and the central battery monitoring systems varies between  $V_{REF}^{i}$ ±5V. The value of 5 volts was 35 chosen for convenience since the normal operating voltage for the microcontroller 31 is 5 volts above the ground voltage for that cell. Theoretically, where the data transmitted between cells is given by  $V_{REF}^{i} \pm X$  volts, X must be greater than half the cell voltage  $V_{CELL}$  in order for the 40 comparator to be able to regenerate the received data pulses at the increased or decreased potential. Practically, since the battery cells and the comparators are not ideal, X should be at least two and a half times the cell voltage  $V_{CELL}$ .

In the first embodiment, a cell monitoring device was used 45 to monitor each cell of the battery. In a cheaper implementation, each cell monitoring device CM, could be used to monitor two or three series connected battery cells  $C_i$ . However, in such an embodiment, where the data transmitted between cell monitoring device is given by  $V_{REF}^{i} \pm X$ volts, X should be at least two and a half times the difference in the reference potentials between adjacent cell monitoring devices.

In the first embodiment, the received data pulses are compared with an approximation of the ground or reference 55 voltage of the cell which sent the data pulses. Alternatively, the received data pulses could simply be compared with the reference voltage of the cell monitoring device which receives the data pulses.

In the embodiments described, the cells are connected in 60 series. It is possible to connect the battery cells C, in a series-parallel or ladder configuration. FIG. 14 shows such an interconnection of battery cells, in which cell  $C_{ia}$  is connected in parallel with cell Cib and the parallel combinations  $C_{ia}$  and  $C_{ib}$  are connected in series for i=1 to n. In 65 the configuration shown in FIG. 14, a single cell monitoring device CM, is provided for monitoring each of the battery

cells and the communication link 9 connects  $CM_{ia}$  to  $CM_{ib}$ and  $CM_{ib}$  to  $CM_{i+1a}$  etc. Alternatively, a single cell monitoring device could be used to monitor each parallel combination of battery cells Cia and Cib. Additionally, more than two battery cells can be connected in parallel.

In the above embodiments, the central battery monitoring and/or control system was provided at the zero volt reference voltage end of the communication link 9. Alternatively, the central battery monitoring and/or control system could be connected at the high reference voltage end of the communication link 9. Alternatively still, the central battery monitoring and/or control system could be connected at both ends, thereby forming a circular communications path in which messages which are transmitted to and received from the battery monitoring/controlling system are passed in one direction through the cell monitoring/controlling devices. Therefore, each cell monitoring/controlling device only needs either an up-link or a down-link for increasing or decreasing the DC level of the received signals, depending on whether the messages are transmitted up or down the communication staircase.

In the above described embodiments, the communication link 9 comprised either one or two wires. As those skilled in the art will appreciate, the communication link 9 may comprise any number of wires along which data can be transmitted in parallel.

In the above embodiments, a separate central battery monitoring system or a central battery control system was provided. In an alternative embodiment, a combined battery monitoring and control system could be used to both monitor and control the battery.

In the above described embodiments, a single battery comprising a plurality of battery cells, is monitored and/or controlled by a central battery monitoring and/or controlling system. FIG. 15 shows an alternative embodiment where a plurality of batteries  $B_i$  are provided, and wherein each battery B, is monitored by its own central battery monitoring system BM, which communicates with a remote operator's terminal 151 via a data bus 153. The data bus 153 may be a proprietary data link or can be the public telephone exchange. In operation, each of the central battery monitoring systems  $BM_i$  monitors the respective battery  $B_i$  and reports its status back to the remote operator's terminal 151, where the condition of each of the batteries is monitored by a human operator. A similar system could also be provided for controlling or for monitoring and controlling a plurality of batteries.

In the first embodiment, a cell monitoring device was used to monitor each cell of the battery. In an alternative less expensive implementation, each cell monitoring device CM<sub>i</sub> could be used to monitor two or more series connected battery cells  $C_i$ .

In the above embodiments which employ MOSFET switches, the cell signalling devices were connected in series in a daisy-chain configuration, with the position of each cell signalling device in the series communication link corresponding with the position of the cell or cells which are to power the cell signalling device in the series connection of battery cells. This is not essential. The cell signalling devices can be connected in any arbitrary series configuration relative to the series connection of battery cells. This is because the MOSFET switches  $Q_1$  and  $Q_2$  operate in the same manner irrespective of the voltage loading applied across their drain and source electrodes. However, in this case, the values of the resistors  $R_a$  to  $R_f$  in each cell monitoring/ control device will be different and will be chosen so as to provide the necessary voltage division having regard to the difference in voltage between adjacent cell monitoring devices in the communications link.

The present invention is not limited by the exemplary embodiments described above, and various other modifications and embodiments will be apparent to those skilled in 5 the art.

What is claimed is:

1. A signalling system for use with a plurality of series connected battery cells, comprising:

- first and second cell signalling devices, each to be powered by a respective one or more of said plurality of battery cells; and
- a communication link for connecting an output terminal of said first cell signalling device to an input terminal of said second cell signalling device;

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characterised in that at least one of said first and second cell signalling devices comprises a DC level shift circuit which is operable (i) to receive signals transmitted from an adjacent cell signalling device; (ii) to shift the DC level of the received signals; and (iii) to 20 output the level shifted signals for transmission to said communication link.

2. A signalling system according to claim 1, wherein said DC level shift circuit is operable (i) to receive signals from an adjacent cell signalling device which is to be powered by a cell having a higher ground potential than that of the receiving cell signalling device; (ii) to decrease the DC level of the received signals; and (iii) to output the level shifted signals for transmission to said communication link.

**3**. A signalling system according to claim **1**, wherein said 30 DC level shift circuit is operable (i) to receive signals from an adjacent cell signalling device which is to be powered by a cell having a lower ground potential than that of the receiving cell signalling device; (ii) to increase the DC level of the received signals; and (iii) to output the level shifted 35 signals for transmission to said communication link.

4. A signalling system according to claim 1, wherein each cell signalling device comprises at least one sensor input terminal operable to receive a signal from a sensor, which are to power the cell signalling device.

5. A signalling system according to claim 4, wherein each of said cell signalling devices comprises a sensor input terminal operable to receive a signal from an electrolyte level and/or electrolyte pH sensor, which signal is indicative 45 cell signalling device. of the electrolyte level and/or the electrolyte pH of the cell or cells which are to power the cell signalling device.

6. A signalling system according to claim 4, wherein each cell signalling device comprises a sensor input terminal operable to receive a signal from a voltage sensor, which 50 signal is indicative of the voltage of the cell or cells which are to power the cell signalling device.

7. A signalling system according to claim 4, wherein each cell signalling device comprises a sensor input terminal which is operable to receive a signal from a temperature 55 sensor, which signal is indicative of the temperature of the cell or cells which are to power the cell signalling device.

8. A signalling system according to claim 4, wherein each cell signalling device comprises a sensor input terminal operable to receive a signal from a voltage interconnection 60 sensor, which signal is indicative of the voltage drop between the cell which is to power said cell signalling device and its adjacent cells.

9. A signalling system according to claim 1, wherein each cell signalling device comprises two of said DC level shift 65 circuits, one of which is operable (i) to receive signals from an adjacent cell signalling device which is to be powered by

a cell having a higher ground potential than that of the receiving cell signalling device; (ii) to decrease the DC level of the received signals; and (iii) to output the level shifted signals for transmission to said communication link; and the other one of which is operable (i) to receive signals from an adjacent cell signalling device which is to be powered by a cell having a lower ground potential than that of the receiving cell signalling device; (ii) to increase the DC level of the received signals; and (iii) to output the level shifted signals for transmission to said communication link.

10. A signalling system according to claim 9, wherein said communication link comprises a single wire communication bus, and wherein said two DC level shift circuits lie on two separate data transfer paths which are connectable to said single wire communication bus by a switch.

11. A signalling system according to claim 9, wherein said two DC level shift circuits are located on separate data transfer paths, and wherein said communication link comprises a two wire communication bus for connecting the respective data transfer paths with corresponding data transfer paths of an adjacent cell signalling device.

12. A signalling system according to claim 1, further comprising a central battery monitoring system for monitoring the battery as a whole, and wherein each of said cell signalling devices is operable to communicate, via said communication link, with said central battery monitoring system.

13. A signalling system according to claim 12, wherein each cell signalling device comprises:

- at least one sensor input terminal operable to receive a signal from a sensor, which signal is indicative of a condition of the cell or cells which are to power the cell signalling device; and
- a signal generator operable to generate a signal in dependence upon said sensor signal and to output said generated signal for transmission to said central battery monitoring system.

14. A signalling system according to claim 13, wherein said central battery monitoring system is operable to poll signal is indicative of a condition of the cell or cells which 40 each of said plurality of cell signalling devices in turn, and wherein upon being polled, each cell signalling device is operable to return a signal back to said central battery monitoring system via said communication link, which is indicative of said condition of the cell which is to power said

> 15. A signalling system according to claim 13, wherein said condition is the cell voltage and wherein said central battery monitoring system is operable to measure the battery charging and discharging current and to calculate the internal resistance of each battery cell by correlating said charging and discharging current with the cell voltages determined by the respective cell signalling devices.

> 16. A signalling system according to claim 13, wherein said central battery monitoring system is operable to monitor the battery voltage, the battery temperature, the total battery current and the total level of charge.

> 17. A signalling system according to claim 1, wherein each of said cell signalling devices is operable to receive a control signal from said communication link and comprises a signal generator operable to generate an actuation signal in dependence upon said received control signal and to output said generated actuation signal for controlling an actuator.

> 18. A signalling system according to claim 17, further comprising a central battery control system for transmitting said control signal to said communication link.

19. A signalling system according to claim 17, wherein each cell signalling device comprises a sensor input terminal operable to receive a signal from an electrolyte level and/or electrolyte pH sensor, which signal is indicative of the electrolyte level and/or the electrolyte pH of the cell or cells which are to power the cell signalling device, and wherein upon receiving said control signal said cell signalling device is operable to output an actuation signal in dependence upon said sensor signal for controlling the addition of water and acid to the cell in order to control its electrolyte level and/or its electrolyte pH.

20. A signalling system according to claim 17, wherein 10 ling device. said actuation signal is for controlling a display. 35. A sign

**21.** A signalling system according to claim **13**, wherein said signal generator comprises a microcontroller which is operable to receive communications from and to transmit communications to said communication link.

22. A signalling system according to claim 21, wherein the microcontrollers of said signalling devices are independently addressable so that communications can be directed to a selected one or more of said cell signalling devices via said communication link.

**23.** A signalling system according to claim **22**, wherein the microcontrollers of said cell signalling devices are operable to communicate with each other.

**24**. A signalling system according to claim **1**, wherein said DC level shift circuit comprises a comparator.

**25**. A signalling system according to claim **24**, wherein said comparator comprises a voltage comparator.

**26**. A signalling system according to claim **25**, wherein the communications transmitted over said communication link comprise square wave signals, and wherein each of said 30 comparators is arranged to compare said square wave signals with a reference signal which is an approximation of the ground potential of the adjacent cell signalling device which transmitted the received square wave signals and to output a square wave signal in dependence upon whether or not the 35 received square wave signal is greater or less than said reference signal.

27. A signalling system according to claim 24, wherein said comparator comprises a current comparator.

**28.** A signalling system according to claim **24**, wherein 40 alternate voltage to current comparators and current to voltage comparators are used in adjacent cell signalling devices.

**29**. A signalling system according to claim **1**, wherein said DC level shift circuit comprises a switch and wherein said 45 signalling devices comprises means for opening and closing said switch in dependence upon the signal to be transmitted.

**30**. A signalling system according to claim **29**, wherein said switch comprises a transistor.

**31.** A signalling system according to claim **30**, wherein a 50 source electrode of said switch is connected to a ground potential of one of said first and second cell signalling devices, wherein a drain electrode of said switch is connected to a positive terminal of the other one of said first and second cell signalling devices, wherein said means for 55 opening and closing said switch operates on a gate electrode of the switch in dependence upon the signal to be transmitted to the other cell signalling device and wherein said other cell signalling device and wherein said other cell signalling device and wherein said other cell signalling device specification is for sensing the change in impedance of said transistor switch.

**32**. A signalling system according to claim **31**, wherein said sensing means comprises a current sensor for sensing the variation of current drawn by said switch as it is opened and closed.

**33.** A signalling system according to claim **31**, wherein said sensing means comprises a voltage divider connected in series with said switch for sensing the change in voltage across the switch as it is opened and closed.

**34**. A signalling system according to claim **1**, wherein each cell signalling device comprises a DC to DC convertor which is operable to convert the cell voltage of the cell which is to power the cell signalling device, to supply voltages and a ground voltage for powering the cell signalling device.

**35**. A signalling system according to claim **1**, wherein a cell signalling device is provided for each of said series connected battery cells.

**36**. A signalling system according to claim **1**, wherein one or more of said series connected battery cells are connected in parallel with one or more additional battery cells.

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**37**. A cell signalling device for use in a signalling system according to claim 1, comprising:

- a power input terminal connectable to the cell or cells which is or are to power said cell signalling device; and
- at least one DC level shift circuit which is operable (i) to receive signals transmitted from an adjacent cell signalling device; (ii) to shift the DC level of the received signals; and (iii) to output the level shifted signals for transmission to the communication link forming part of said signalling system.

**38**. A signalling kit for use in a signalling system according to claim **1**, comprising a plurality of cell signalling devices each of which comprises:

- a power input terminal connectable to the cell or cells which is or are to power said cell signalling device; and
- at least one DC level shift circuit which is operable (i) to receive signals transmitted from an adjacent cell signalling device; (ii) to shift the DC level of the received signals; and (iii) to output the level shifted signals for transmission to the communication link forming part of said signalling system.

**39**. A signalling kit according to claim **38**, further comprising a communication link for connecting said plurality of cell signalling devices in series.

**40**. A signalling system according to claim **1** in combination with a plurality of series connected battery cells, wherein one or more of said battery cells are connected to a respective one of said plurality of cell signalling devices, for powering said cell signalling devices.

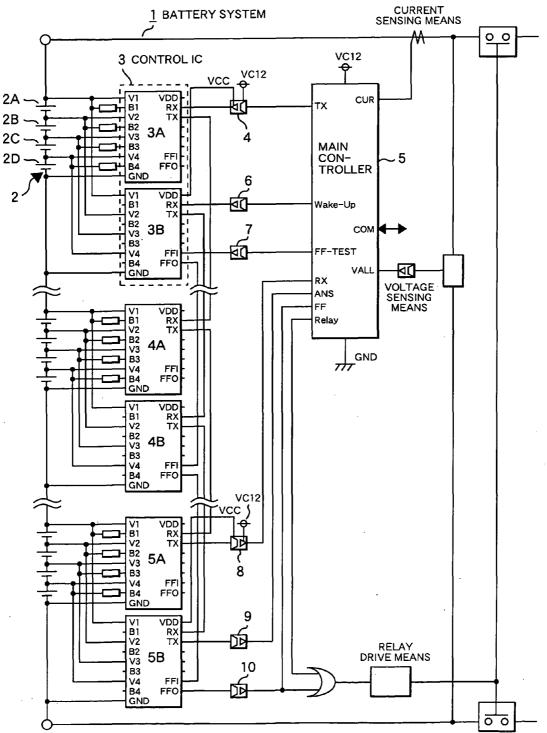
41. A cell signalling device according to claim 37 in combination with a battery cell, wherein the terminals of said battery cell are connectable to said cell signalling device.

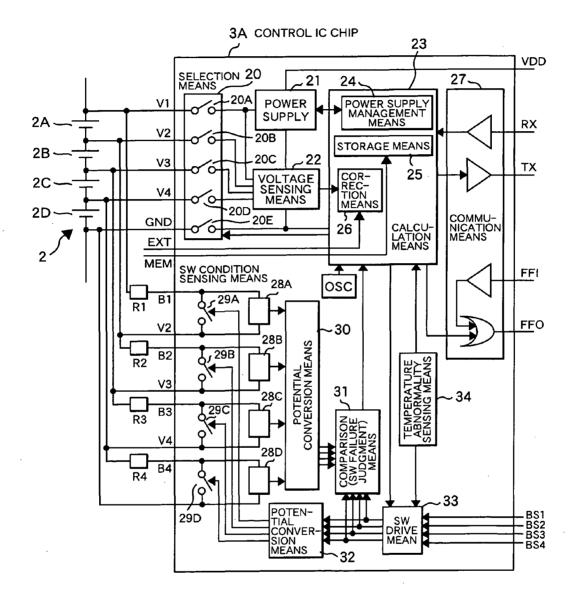
**42**. A signalling method using a plurality of series connected battery cells, comprising the steps of:

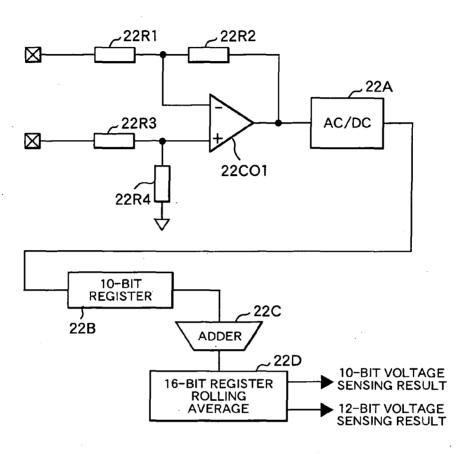
- providing a plurality of cell signalling devices and powering them with a respective one or more of said plurality of battery cells;
- providing a communication link for connecting said plurality of cell signalling devices in series;
- receiving signals transmitted from an adjacent cell signalling device;
- shifting the DC level of the received signals; and
- outputting the level shifted signals to the communication link.

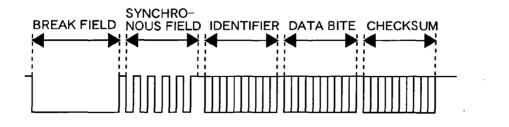
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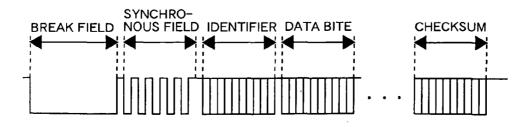
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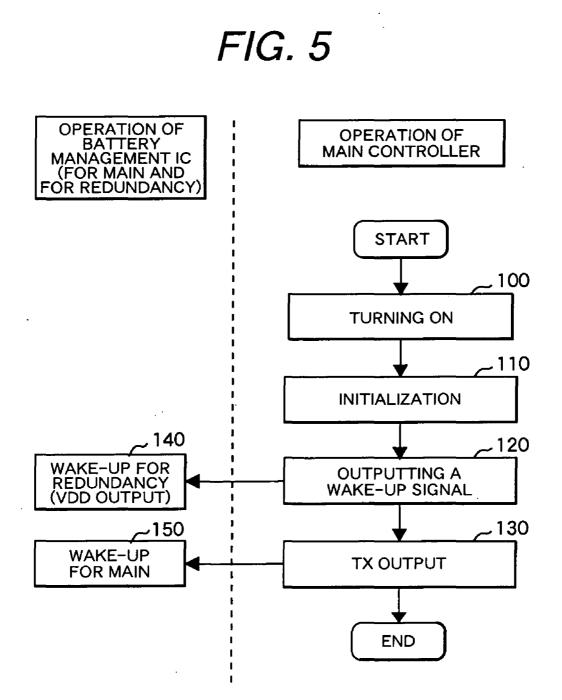


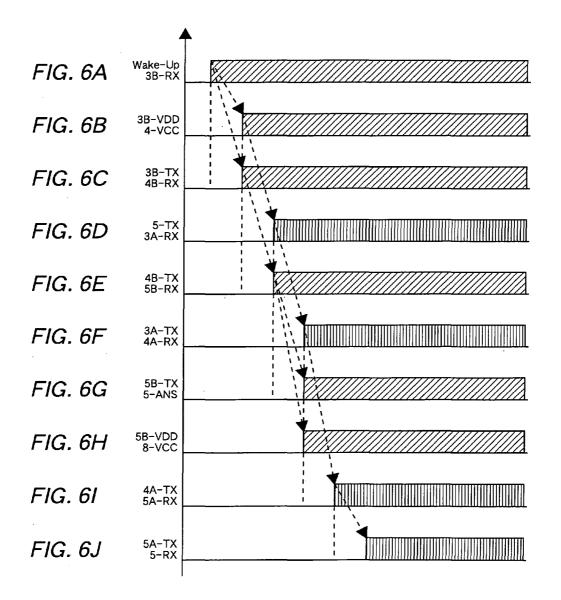


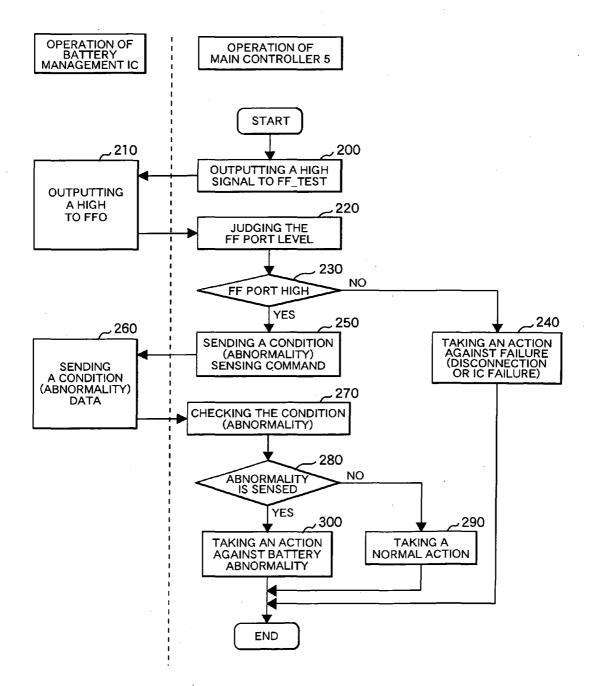


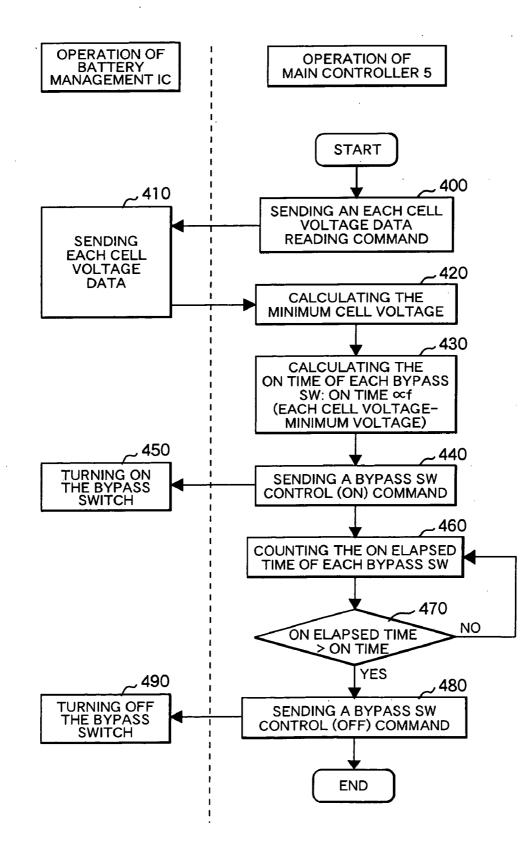


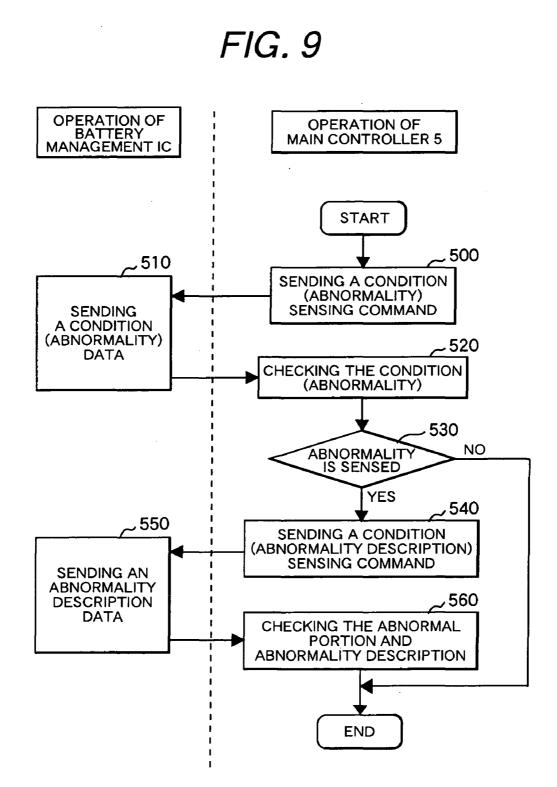


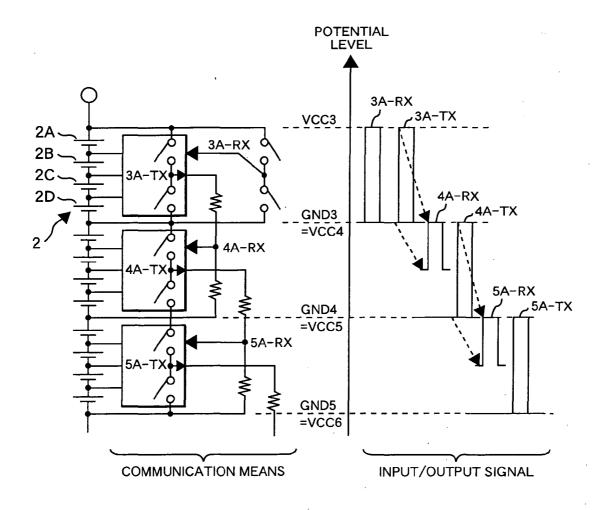


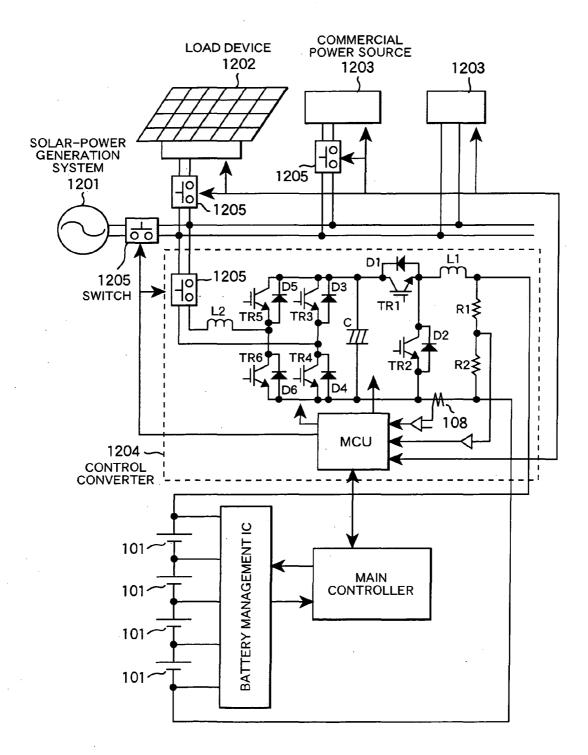


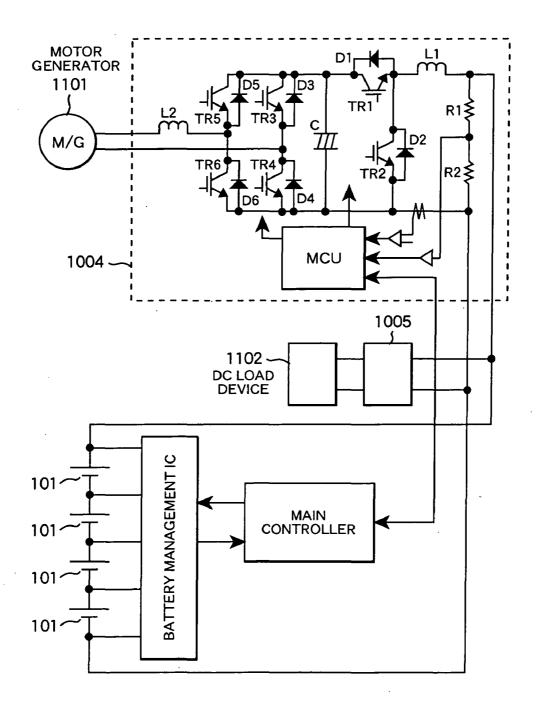












### **MULTI-SERIES BATTERY CONTROL SYSTEM**

CLAIM OF PRIORITY

[0001]

### BACKGROUND OF THE INVENTION

**[0002]** The present invention relates to secondary battery (lithium battery) for vehicle or for power supply, particularly to a multi-series battery control system for managing the condition of the secondary battery (lithium battery).

**[0003]** In order to solve a problem that higher precision voltage sensing involves cost increase of insulation means installed for each of multiple lower control units, there has been proposed a storage battery unit aiming at cost reduction by reducing the number of insulation means, for example, refer to Japanese Patent Laid-open 2003-70179, especially pages 3-4, FIG. 1. This Patent Document aims to lower the effect of disturbance such as noise and thereby improve reliability and enable higher precision voltage sensing.

### SUMMARY OF THE INVENTION

[0004] A conventional storage battery unit is so constructed as to comprise multiple series-connected battery modules, each consisting of multiple series-connected storage batteries, multiple lower control units that are provided corresponding respectively to the multiple battery modules and control the multiple storage batteries constituting the battery module, and upper control unit that controls the multiple lower control units, wherein there are provided an input terminal of the lower control unit positioned at the maximum potential among the multiple lower control units and output terminal of the lower control unit positioned at the minimum potential, insulation means or potential conversion means that connects the upper control unit, and cutout device that is installed between the output terminal of the lower control unit and storage battery in the battery module on the lower potential side and cuts out the discharge current from the storage battery in the battery module, and signal input/output terminals of the multiple lower control units are connected with each other, electrically not insulated.

**[0005]** This conventional storage battery unit is not satisfactory in view of high reliability.

**[0006]** An object of the present invention is to offer a multi-series battery control system that can realize high reliability.

[0007] In an aspect of the invention, a multi-series battery control system comprising: a plurality of unit battery cells of which unit consists of multiple battery cells connected in series; a plurality of control ICs each comprising a control circuit for controlling the unit battery cell; a main controller that sends and receives signal to/from the control ICs via an insulation; means for sending an abnormality signal, which represents the existence or the absence of abnormality of the control ICs, responding to the first signal outputted from the main controller via the insulation; and means for sending an abnormality contents signal of the control ICs or the battery cells, to the main controller from the control ICs, responding to the second signal outputted from the main controller via the insulation.

[0008] In another aspect of the invention, A multi-series battery control system comprising: a plurality of unit battery cells of which unit consists of multiple battery cells connected in series; a plurality of control IC chips comprising a control circuit for controlling the unit battery cell; a plurality of cell monitor IC chips each monitoring the voltage of the unit battery cell; a plurality of control ICs each consisting of the control IC chip and the cell monitor IC chip; a main controller that sends and receives signal to/from the control IC chips via an insulation; means for sending an abnormality signal, which represents the existence or the absence of abnormality of the control IC chips or the battery cells, to the main controller from the control IC chips, responding to the first signal outputted from the main controller via the insulation; and means for sending the abnormality contents signal of the control IC chip or the battery cells, to the main controller from the control IC chips, responding to the second signal outputted from the main controller via the insulation.

**[0009]** According to a preferred embodiment of the present invention, a high reliability multi-series battery control system can be realized.

**[0010]** Other objects and features of the present invention are described hereunder along with preferred embodiments.

### BRIEF DESCRIPTION OF DRAWINGS

**[0011] FIG. 1** is a diagram showing an embodiment of the multi-series battery control system of the present invention.

**[0012]** FIG. 2 is a detailed circuit diagram of the control IC chip shown in FIG. 1.

**[0013]** FIG. 3 is a block diagram showing a concrete embodiment of the voltage sensing means shown in FIG. 2.

**[0014] FIG. 4** is a diagram showing an embodiment of communication command.

**[0015] FIG. 5** is a chart showing the wake-up sequence in starting up the system.

**[0016] FIG. 6** is a chart showing the sent/received wakeup signals on each control IC shown in **FIG. 5**.

[0017] FIG. 7 is a chart showing an FF-TEST subroutine.

**[0018] FIG. 8** is a chart showing a balancing subroutine for switching the balancing switch.

**[0019] FIG. 9** is a flow chart showing an operation flow for testing abnormality of battery cell or IC chip.

**[0020]** FIG. 10 is a chart for explaining how communication signal is sensed in each control IC.

**[0021] FIG. 11** is a diagram showing a case where the invention is employed in combination with a commercial power source.

**[0022]** FIG. 12 is a diagram showing a case where the invention is applied to a motor generator.

### DETAILED DESCRIPTION OF THE INVENTION

**[0023]** The present invention can be realized by monitoring multiple series-connected battery cells as a unit and managing the condition of a battery so as to sense abnormality of IC chip circuit or battery cell and take appropriate measures.

### **EMBODIMENT** 1

**[0024]** An embodiment of the multi-series battery control system according to the present invention is described hereunder in detail.

**[0025] FIG. 1** shows the construction of the multi-series battery control system of the present invention.

[0026] In FIG. 1, a battery system 1 is so constructed that a unit battery cell 2 comprising four series-connected battery cells 2A, 2B, 2C, 2D is provided with a corresponding paired control IC 3 (control IC chip 3A and cell monitor IC chip 3B). The paired control IC 3 consists of two ICs: one is the control IC chip 3A that contains a control circuit and the other is the cell monitor IC chip that monitors the unit battery cell. One end of the control IC chip 3A is connected with each terminal of the battery cells 2A, 2B, 2C, 2D of the unit battery cell 2. The other end of the control IC chip 3A is connected with a main controller 5 via a high-speed insulation means 4. The main controller 5 is connected with one end of the cell monitor IC chip 3B via insulation means 6 and 7. The other end of the cell monitor IC chip 3B is connected with each terminal of the battery cells 2A, 2B, 2C, 2D.

[0027] The paired control IC 3 is provided for each unit battery cell comprising four battery cells. Although FIG. 1 shows only three paired ICs, the number of paired ICs 3 is the same number of units of all battery cells of a lithium battery where one unit comprises four battery cells.

[0028] FIG. 2 shows a detailed circuit of the control IC chip 3A. Although the figure shows the control IC chip 3A only, the other control IC chip 4A, 5A... have the same construction. Moreover, the cell monitor IC chips 3B-5B in FIG. 1 can also completely be constituted using the same IC chip. Then, in the following explanation, FIG. 2 is referred to also to explanation of the cell monitor IC chips 3B-5B.

[0029] In FIG. 2, the (+) terminal of the battery cell 2A of the unit battery cell 2 is connected with a selection means 20 via the V1 input terminal. This selection means 20 is a multiplexer, for example. The selection means is provided with switches 20A, 20B, 20C, 20D, 20E. One end of the switch 20A is connected with the V1 input terminal and the other end of the switch 20A is connected with a power supply 21 and voltage sensing means 22. In addition, the (-) terminal of the battery cell 2A of the unit battery cell 2, which is the (+) terminal of the battery cell 2B, is connected with one end of the switch 20B of the selection means 20 via the V2 input terminal and the other end of the switch 20B is connected with the voltage sensing means 22.

[0030] In addition, the (-) terminal of the battery cell 2B of the unit battery cell 2, which is the (+) terminal of the battery cell 2C, is connected with one end of the switch 20C of the selection means 20 via the V3 input terminal and the other end of the switch 20C is connected with the voltage sensing means 22. Furthermore, the (-) terminal of the battery cell 2C of the unit battery cell 2, which is the (+) terminal of the battery cell 2D, is connected with one end of the switch 20D of the selection means 20 via the V4 input

terminal and the other end of the switch **20**D is connected with the voltage sensing means **22**.

[0031] The (-) terminal of the battery cell 2D of the unit battery cell 2 is connected with one end of the switch 20E of the selection means 20 via the GND (ground) terminal and the other end of the switch 20E is connected with the voltage sensing means.

[0032] The power supply 21 is constructed for example as a DC/DC converter, which is made using the unit battery cells so as to convert the power of the unit battery cell 2 to a specified voltage and supply to the outside through the VDD terminal and also to supply drive power to each circuit in the control IC chip 3A.

[0033] The voltage sensing means 22 senses each terminal-to-terminal voltage between the battery cells 2A, 2B, 2C, 2D of the unit battery cell 2, and the sensed terminalto-terminal voltage between the battery cells 2A, 2B, 2C, 2D is outputted to a calculation means 23. The calculation means comprises a power supply management means 24, storage means 25 and correction means 26. The power supply management means 24 controls ON/OFF of the power supply 21.

[0034] The storage means 25 stores each terminal-toterminal voltage between the battery cells 2A, 2B, 2C, 2D of the unit battery cell 2 sensed by the voltage sensing means 22 separately for each battery cell 2A, 2B, 2C, 2D. To be concrete, the storage means 25 is constructed as a shift register. The correction means 26 corrects each terminal-toterminal voltage between the battery cells 2A, 2B, 2C, 2D of the unit battery cell 2 sensed by the voltage sensing means 22.

[0035] The calculation means 23 is connected with a communication means 27. The communication means 27 receives through the RX terminal via the high-speed insulation means 4 a communication command (such as 8-bit, 10-bit or 12-bit ON/OFF signal) sent from the main controller 5. That is to say, the main controller 5 sends out a command for operating a specified control IC chip 3A to the high-speed insulation means 4, including a communication command for reading the voltage between each battery cell 2A, 2B, 2C, 2D or communication command for adjusting the voltage between each battery cell 2A, 2B, 2C, 2D of a specified unit battery cell 2. The high-speed insulation means 4 does not send out the communication command received from the main controller 5 directly to the communication means 27 but via the insulation means.

[0036] The high-speed insulation means 4 is a transformer type and as small as an IC. Being a transformer type, the high-speed insulation means 4 needs power and is driven by power supplied from the cell monitor chip 3B.

[0037] The communication means 27 generates a communication command (such as 10-bit or 12-bit ON/OFF signal), corresponding to the communication command sent from the main controller 5 via the high-speed insulation means 4, by the voltage of eight battery cells, that is, two seriesconnected unit battery cells and outputs it to the calculation means 23.

[0038] How communication signal is sensed in each control IC chip 3A, 4A, . . . 5A is described hereunder, using FIG. 10.

[0039] In FIG. 10, among the control IC chip 3A, control IC chip 4A, ... control IC chip 5A, communication signal is judged by the control IC chip 3A and control IC chip 4A, and the control IC chip 4A and control IC chip 5A. In the control IC chip 3A in FIG. 10, a VCC3 voltage level signal (Hi/Low signal of the VCC3 voltage level), of which Hi is the total voltage of the sum of each voltage of battery cells 2A, 2B, 2C, 2D of the unit battery cell 2 and Low is the GND (ground) level, is inputted into the RX terminal of the control IC chip 3A. The VCC3 voltage level signal inputted into the RX terminal of the control IC chip 3A is outputted from the TX terminal of the control IC chip 3A but the voltage is divided by a resistor, and so a divided VCC3 voltage level signal (Hi/Low signal of the VCC3 voltage level divided to 1/2 for example) is inputted into the RX terminal of the control IC chip 4A. That is, the communication signal is inputted into the RX terminal of the control IC chip 4A as a signal for example repeating Hi/Low of the VCC3 voltage level divided to ½. If the control IC chip 4A attempts to judge the signal outputted from the TX terminal of the control IC chip 3A using the same threshold as for the control IC chip 3A based on each voltage of the unit battery cell 2 under its control, judgment is impossible because the Low level of the signal outputted from the TX terminal of the control IC chip **3**A is half the total voltage applied to the control IC chip 4A.

[0040] That is to say, in FIG. 10, the RX terminal voltage of the control IC chip 3A ranges VCC3 to GND3. The RX terminal voltage of the control IC chip 4A ranges (VCC3 to GND4)×R/2R because the voltage outputted from the TX terminal of the control IC chip 3A is divided to  $\frac{1}{2}$  by a resistor. The TX terminal voltage of the control IC chip 4A ranges VCC4 to GND4. The RX terminal voltage of the control IC chip 5A ranges (VCC4 to GND5)×R/2R because the voltage outputted from the TX terminal of the control IC chip 4A is divided to  $\frac{1}{2}$  by a resistor.

[0041] Accordingly, both input and output (RX and TX) of the highest control IC chip 3A ranges VCC to GND. The threshold of the highest control IC chip 3A for judging Hi/Low of the input (RX) is therefore  $\frac{1}{2}$  VCC. The output (TX) of other control IC chips (4A, ... 5A) than the highest control IC chip 3A ranges VCC to  $\frac{1}{2}$  VCC. Accordingly, for smooth operation, the threshold of each control IC chip 4A, ... 5A for judging Hi/Low of the input (RX) shall be nothing but  $\frac{3}{4}$  VCC.

[0042] In addition, the (+) terminal of the battery cell 2A of the unit battery cell 2 is connected with the B1 terminal via a resistor R1. This B1 terminal is connected with one end of the SW condition sensing means 28A and the other end of the SW condition sensing means 28A is connected with the (-) terminal of the battery cell 2A of the unit battery cell 2 via the V2 terminal. And, a balancing switch 29A series-connected with the resistor R1 is inserted between the two terminals of the battery cell 2A of the unit battery cell 2.

[0043] In addition, the (+) terminal of the battery cell 2B of the unit battery cell 2 is connected with the B2 terminal via a resistor R2. This B2 terminal is connected with one end of the SW condition sensing means 28B and the other end of the SW condition sensing means 28B is connected with the (-) terminal of the battery cell 2B of the unit battery cell 2 via the V3 terminal. And, a balancing switch 29B series-connected with the resistor R2 is inserted between the two terminals of the battery cell 2B of the unit battery cell 2.

[0044] In addition, the (+) terminal of the battery cell 2C of the unit battery cell 2 is connected with the B3 terminal via a resistor R3. This B3 terminal is connected with one end of the SW condition sensing means 28C and the other end of the SW condition sensing means 28C is connected with the (-) terminal of the battery cell 2C of the unit battery cell 2 via the V4 terminal. And, a balancing switch 29C series-connected with the resistor R3 is inserted between the two terminals of the battery cell 2C of the unit battery cell 2.

[0045] Furthermore, the (+) terminal of the battery cell 2D of the unit battery cell 2 is connected with the B4 terminal via a resistor R4. This B4 terminal is connected with one end of the SW condition sensing means 28D and the other end of the SW condition sensing means 28D is connected with the (-) terminal of the battery cell 2D of the unit battery cell 2. And, a balancing switch 29D series-connected with the resistor R4 is inserted between the two terminals of the battery cell 2D of the unit battery cell 2.

[0046] These SW condition sensing means 28A, 28B, 28C, 28D sense the voltage between both ends of the balancing switches 29A to 29D, respectively. They also sense abnormality of the balancing switches 29A, 29B, 29C, 29D. That is to say, if the terminal voltage of the battery cells 2A, 2B, 2C, 2D is outputted while the balancing switches 29A, 29B, 29C, 29D are ON, the balancing switches 29A, 29B, 29C, 29D can be judged abnormal. These SW condition sensing means 28A, 28B, 28C, 28D are a voltage sensing circuit comprising a differential amplifier.

[0047] These balancing switches 29A, 29B, 29C, 29D are switches that short-circuits each battery cell via the resistor R1, resistor R2, resistor R3, and resistor R4 respectively so as to discharge the series-connected battery cells 2A, 2B, 2C, 2D constituting the unit battery cell 2 and match the battery cell voltage of the four battery cells 2A, 2B, 2C, 2D constituting the unit battery cell with each other. To be concrete, they are constructed as a MOS type FET. In addition, the SW condition sensing means 28A senses whether the balancing switch 29A is operating correctly, SW condition sensing means 28B senses whether the balancing switch 29B is operating correctly, SW condition sensing means 28C senses whether the balancing switch 29C is operating correctly, and SW condition sensing means 28D senses whether the balancing switch 29D is operating correctly. That is, the SW condition sensing means 28A to 28D continuously monitor the voltage of the balancing switches 29A to 29D, and when the balancing switches 29A, 29B, 29C, 29D are turned ON, the SW condition sensing means 28A, 28B, 28C, 28D sense a voltage near 0 (zero), respectively.

[0048] A potential conversion means 30 is connected with these SW condition sensing means 28A, 28B, 28C, 28D. The potential conversion means 30 convert the voltage between each battery cell 2A, 2B, 2C, 2D sensed by the SW condition sensing means 28A, 28B, 28C, 28D to a specific potential (potential suitable for processing) and output it to a comparison means 31. That is, since the potential levels between each battery cell 2A, 2B, 2C, 2D are different, the potential conversion means 30 converts them to such potential levels that can be compared with each other.

[0049] The comparison means 31, into which a drive signal of a SW drive means 33 is inputted, compares the drive signal with the voltage, which is the voltage between

each balancing switches **29A**, **29B**, **29C**, **29D** sensed by the SW condition sensing means **28A**, **28B**, **28C**, **28D**, and converted into a specific voltage (potential suitable for processing) and outputted from the potential conversion means **30**, and judges whether the balancing switches **29A**, **29B**, **29C**, **29D** are normal or abnormal.

[0050] On the other hand, a signal for driving the balancing switch 29A is inputted via the BS1 terminal, signal for driving the balancing switch 29B is inputted via the BS2 terminal, signal for driving the balancing switch 29C is inputted via the BS3 terminal, and signal for driving the balancing switch 29D is inputted via the BS4 terminal from the main controller 5 into the SW drive means 33, respectively. The SW drive means 33 converts the switch signal sent from the main controller 5 into each switch drive signal and outputs it to the comparison means 31 connected with the SW drive means and potential conversion means 32.

[0051] The potential conversion means 32 receives the switch drive signal sent from the SW drive means 33, converts it to a drive voltage signal (to be concrete, a gate signal) for turning ON/OFF the balancing switches 29A, 29B, 29C, 29D, and supplies it (to be concrete, supplies a gate voltage) to the balancing switches 29A, 29B, 29C, 29D

[0052] When abnormality of the balancing switches 29A, 29B, 29C, 29D is sensed by the comparison means 31, it identifies which balancing switch 29A, 29B, 29C, 29D is abnormal based on the switch drive signal outputted from the SW drive means 33 and outputs the result to the calculation means 23. When abnormality is sensed by the comparison means 31, the calculation means 23 identifies an abnormal balancing switch and sends a signal informing the abnormality is sent to the main controller 5 from the FFO terminal of the communication means 27.

[0053] In FIG. 2, the BS1 to BS4 terminals in the SW drive means 33 are used to input a signal for turning ON the balancing switches 19A to 19D from the outside, and the signal inputted from these terminals BS1 to BS4 drives the SW drive means 33 and the SW drive means 33 sends out an ON signal of the balancing switches 19A to 19D to the potential conversion means 32. The potential conversion means 32 receives the switch drive signal sent from the SW drive means 33, converts it to a drive voltage signal (to be concrete, a gate signal) for turning ON the balancing switches 29A, 29B, 29C, 29D, and supplies it (to be concrete, supplies a gate voltage) to the balancing switches 29A, 29B, 29C, 29D.

[0054] 34 in FIG. 2 is a temperature abnormality sensing means, and the temperature abnormality sensing means 34 senses the temperature of the control IC chip 3A, checking whether it reaches a preset temperature. If the temperature abnormality sensing means 34 senses a temperature in excess of the preset temperature, it sends out a signal to the SW drive means 33 so as to stop supplying current to the balancing switches 29A to 29D and terminate the charging control by the balancing switches 29A, 29B, 29C, 29D so that no more heat is generated.

[0055] FIG. 3 shows a concrete embodiment of the voltage sensing means 22 shown in FIG. 2.

[0056] The voltage sensing means 22 in FIG. 3 is connected with the selection means 20. The voltage sensing

means 22 is provided with a resistor 22R1 connected with the (+) terminal of the battery cells 2A, 2B, 2C, 2D of which connection is switched by the switches 20A, 20B, 20C, 20D, 20E of the selection means 20. The other end of the resistor 22R1 is connected with one end of a resistor 22R2 and the (-) input terminal of the operation amplifier 22OP1. The other end of the resistor 22R2 is connected with an AC/DC converter 22A.

[0057] On the other hand, there is provided a resistor 22R3 connected with the (-) terminal of the battery cells 2A, 2B, 2C, 2D of which connection is switched by the switches 20A, 20B, 20C, 20D, 20E of the selection means 20, and the other end of the resistor 22R3 is connected with one end of a resistor 22R4 and the (+) input terminal of the operation amplifier 22OP1. The output terminal of the operation amplifier 22OP1 is connected with the AC/DC converter 22A. The other end of the resistor 22R4 is connected with the ground.

[0058] The output terminal of the AC/DC converter 22A is connected with an adder 12C via a 10-bit resistor 22B, and the adder 12C is connected with a 16-bit resistor rolling average 22D.

**[0059]** Because a duplex integration type is employed as explained above, noise content in the input voltage can be filtered. In addition, because a 16-bit resistor rolling average is employed, resolution can be improved and sensed value can be filtered.

**[0060] FIG. 4** shows an embodiment of communication command. This communication command is sent from the main controller **5** and inputted to the RX terminal of the communication means **27** shown in **FIG. 2**. A unit data of this communication command comprises 8 bits and a communication command contains 5 bytes. The first 8 bits of the communication command are a break field informing an incoming signal, second 8 bits are a synchronous field as a signal for synchronization, third 8 bits are an identifier equivalent to an address showing which control IC chip **3A** applies, fourth 8 bits are a data byte showing the communication detail (control detail), and fifth 8 bits are a check-sum. These communication commands consisting of 5 bytes are sent in series.

[0061] FIG. 5 shows the wake-up sequence in starting up the system. That is, this wake-up sequence is the operation flow for actuating the control IC chip 3A and cell monitor IC chip 3B in turning on the main controller 5.

[0062] In FIG. 5, when the main controller 5 shown in FIG. 2 is turned on (key-switch is turned on) in step 100, the main controller 5 is initialized in step 110. After the initialization of the main controller 5 in step 110, a wake-up signal shown in FIG. 6 (A) is outputted from the wake-up terminal of the main controller 5 to the RX terminal of the cell monitor IC chip 3B via the insulation means 6 in step 120. The RX terminal of the cell monitor IC chip 3B and so, when a wake-up signal is inputted to the RX terminal of the cell monitor IC chip 3B, the cell monitor IC chip 3B is actuated (wakes up). When this cell monitor IC chip 3B wakes up, the power VCC supplied from the battery cells 2A, 2B, 2C, 2D as shown in FIG. 6 (B) is outputted from the VDD terminal of the cell monitor IC chip 3B.

**[0063]** When a wake-up signal shown in **FIG. 6** (A) is outputted from the wake-up terminal of the main controller

5 to the RX terminal of the cell monitor IC chip 3B via the insulation means 6 in step 120, the cell monitor IC chip 3B supplies the power VCC shown in FIG. 6 (B) from the VDD terminal to the high-speed insulation means 4 (VDD output) in step 140. When the power VCC shown in FIG. 6 (B) is supplied from the VDD terminal of the cell monitor IC chip 3B to the high-speed insulation means 4, the high-speed insulation means 4 wakes up. When the high-speed insulation means 5 wakes up, the wake-up signal shown in FIG. 6 (D) outputted from the TX terminal of the control IC chip 3A.

[0064] As explained above, the cell monitor IC chip 3B wakes up when a wake-up signal shown in FIG. 6 (A) outputted from the wake-up terminal of the main controller 5 via the insulation means 6 is received at the RX terminal in step 120, and a wake-up signal shown in FIG. 6 (D) for waking up the control IC chip 3A is outputted from the TX terminal of the main controller 5 to the RX terminal of the control IC chip 3A via the high-speed insulation means 4 in step 130. In step 130, a wake-up signal shown in FIG. 6 (D) for waking up the control IC chip 3A is outputted from the TX terminal of the main controller 5 to the RX terminal of the control IC chip 3A via the high-speed insulation means 4 in step 130, and when it is received at the RX terminal of the control IC chip 3A, the control IC chip 3A wakes up in step 150.

[0065] When the cell IC chip 3B is woken up by a wake-up signal shown in FIG. 6 (A) outputted from the wake-up terminal of the main controller 5 to the RX terminal of the cell monitor IC chip 3B via the insulation means 6, the cell IC chip 3B copies the wake-up signal shown in FIG. 6 (A) as a wake-up signal shown in FIG. 6 (C) and outputs it from the RX terminal of the cell monitor IC chip 3B to the RX terminal of the cell monitor IC chip 4B in the next stage. The cell IC chip 4B is woken up by a wake-up signal shown in FIG. 6 (C) outputted from the TX terminal of the cell monitor IC chip 3B, and the cell IC chip 4B copies the wake-up signal shown in FIG. 6 (C) outputted from the TX terminal of the cell monitor IC chip 3B and outputs it as a wake-up signal shown in FIG. 6 (E) from the TX terminal of the cell monitor IC chip 4B to the RX terminal of the cell monitor IC chip 5B in the last stage. There are multiple cell IC chips provided between the cell monitor IC chip 4B and the cell monitor IC chip 5B in the last stage, but they are omitted in FIG. 1.

[0066] On the other hand, the power supplied from the battery cells 2A, 2B, 2C, 2D is outputted from the VDD terminal of the cell monitor IC chip 3B to the high-speed insulation means 4, the high-speed insulation means 4 is turned on, and a wake-up signal shown in FIG. 6 (D) is outputted from the TX terminal of the main controller 5 to the RX terminal of the control IC chip 3A via the high-speed insulation means 4. When this wake-up signal shown in FIG. 6 (D) is inputted to the RX terminal of the control IC chip 3A, the control IC chip 3A wakes up. When the control IC chip 3A wakes up, the control IC chip 3A copies the wake-up signal shown in FIG. 6 (D) sent from the TX terminal of the main controller 5 and outputs it as a wake-up signal shown in FIG. 6 (F) from the TX terminal of the control IC chip 3A to the RX terminal of the control IC chip 4A in the next stage.

[0067] The cell IC chip 4A is woken up by a wake-up signal shown in FIG. 6 (F) outputted from the TX terminal of the cell monitor IC chip 3A, and the cell IC chip 4A copies the wake-up signal shown in FIG. 6 (F) outputted from the TX terminal of the cell monitor IC chip 3A and outputs it as a wake-up signal shown in FIG. 6 (I) from the TX terminal of the cell monitor IC chip 4A to the RX terminal of the cell monitor IC chip 5A in the last stage. There are multiple cell IC chips provided between the cell monitor IC chip 5A in the last stage, but they are omitted in FIG. 1.

[0068] The control IC chip 3A, control IC chip 4A, ... control IC chip 5A and the cell monitor IC chip 3B, cell monitor IC chip 4B, ... cell monitor IC chip 5B wake up as explained above, and the battery management IC for managing the battery cells 2A to 2N wakes up. In addition, there is provided a VDD terminal on each control IC chip 3A, control IC chips 4A to 5A, cell monitor IC chip 3B, and cell monitor IC chips 4B to 5B, and so power can be supplied to the outside by appropriate utilization of these VDD terminals.

[0069] After the cell monitor IC chip 3B wakes up as above, the cell monitor IC chip 4B and remaining cell monitor IC chips wake up by repeating a similar operation. The cell IC chip 5B wakes up as it receives a wake-up signal shown in FIG. 6 (E) outputted from the TX terminal of the cell monitor IC chip 4B at the RX terminal. When the cell monitor IC chip 5B wakes up as it receives a wake-up signal shown in FIG. 6 (E) outputted from the TX terminal of the cell monitor IC chip 5B, the cell monitor IC chip 5B to the RX terminal of the cell monitor IC chip 5B, the cell monitor IC chip 5B to the ANS terminal of the main controller 5.

[0070] When the cell monitor IC chip 5B wakes up, it supplies the power VCC shown in FIG. 6 (H) from the VDD terminal to the high-speed insulation means 8 (VDD output). When the power VCC shown in FIG. 6 (H) is supplied from the VDD terminal of the cell monitor IC chip 5B to the high-speed insulation means 8, the high-speed insulation means 8 wakes up. When the high-speed insulation means 8 wakes up, the high-speed insulation means 8 is turned ON and so the TX terminal of the control IC chip 5A can communicate with the RX terminal of the main controller 6. That is, when the high-speed insulation means 8 wakes up, a wake-up signal shown in FIG. 6 (J) is sent from the TX terminal of the control IC chip 5A to the RX terminal of the main controller 5. When the wake-up signal shown in FIG. 6 (J) from the TX terminal of the control IC chip 5A is received at the RX terminal of the main controller 5, the main controller 5 confirms that the control IC chip 3A, control IC chip 4A, . . . control IC chip 5A and the cell monitor IC chip 3B, cell monitor IC chip 4B, . . . cell monitor IC chip 5B have woken up and the battery management IC for managing the battery cells 2A to 2N has woken up correctly. Whether the cell monitor IC chip 3B, cell monitor IC chip 4B, ... cell monitor IC chip 5B have woken up is judged by confirming that the high-speed insulation means 8 is turned ON and a wake-up signal shown in FIG. 6 (J) is sent from the TX terminal of the control IC chip 5A to the RX terminal of the main controller 5 is the cell monitor IC chip 5B has woken up.

[0071] The battery management IC is provided with a high-speed insulation means 4 on its top stage and high-speed insulation means 8 on its bottom stage for the purpose of insulation and so it is not grounded to chassis (power supply is lifted from the chassis).

[0072] FIG. 7 shows an FF-TEST subroutine. That is, the FF-TEST subroutine is a processing flow for inputting a test signal from the FFI of the cell monitor IC chip 3B and sensing abnormality in the circuits of the cell monitor IC chips 3B to 5B.

[0073] In FIG. 7, a High signal is sent from the FF-TEST terminal of the main controller 5 as shown in FIG. 1 to the FFI terminal of the communication means of the cell monitor IC chip 3B shown in FIG. 1 via an insulation means 7 in step 200. When a High signal is sent to the FFI terminal of the communication means of the cell monitor IC chip 3B in step 200, the cell monitor IC chip 3B outputs the High signal, without adding any processing, from the FFO terminal to the FFI terminal of the cell monitor IC chip 4B in the next stage. Similarly, when a High signal is sent to the FFI terminal of the communication means of the cell monitor IC chip 4B, the cell monitor IC chip 4B outputs the High signal, without adding any processing, from the FFO terminal to the FFI terminal of the cell monitor IC chip 5B in the next stage. Then, when a High signal is sent to the FFI terminal of the communication means of the cell monitor IC chip 5B, the cell monitor IC chip 5B outputs the High signal, without adding any processing, from the FFO terminal to the main controller 5. When a signal is outputted from the FFO terminal of the communication means 27, the FF port level is judged based on the signal sent from the FFO terminal to the main controller 5 in step 220. When the FF port level is judged based on the signal outputted from the FFO terminal and sent to the main controller 5 in step 220, the main controller 5 judges whether the FF port level is High or not in step 230.

**[0074]** If the main controller **5** judges that the FF port level is not High (is Low) in step **230**, it takes an action needed in a case the circuit is disconnected somewhere or the cell monitor IC chip itself is abnormal.

[0075] If the main controller 5 judges the FF port level is High in step 230, it is necessary in step 250 to check if the returned High signal is a signal representing normality (High signal) that has been inputted by chance in spite of overcharging or over-discharging. That is, in step 250, it sends a condition (abnormality) sensing command, which is a command for sensing other abnormality (abnormality of battery cell), to the RX terminal of the communication means 27 provided on the control IC chip 3A. When this condition (abnormality) sensing command is sent to the controller 5 and RX terminal of the communication means 27 of the control IC chip 3A, a condition (abnormality) data, which is a data showing the current condition, is sent from the TX terminal of the communication means 27 of the control IC chip 3A to the main controller 5 in step 260. When this condition (abnormality) data is sent from the TX terminal of the communication means 27 to the main controller 5, the main controller 5 checks the condition (abnormality) in step 270 and judges whether the condition (abnormality) data sent from the TX terminal of the communication means 27 is a signal indicating abnormality in step 280. In a similar manner, it checks the condition (abnormality) of the control

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IC chip 4A and control IC chip 5A and judges whether the condition (abnormality) data sent from the TX terminal of the communication means 27 is a signal indicating abnormality. If the condition (abnormality) data sent from the TX terminal of the communication means 27 is judged to be a signal indicating no abnormality in step 280, it takes a normal action and finishes the flow. If the condition (abnormality) data sent from the TX terminal of the communication means 27 is judged to be a signal indicating abnormality in step 280, it takes a normal action and finishes the flow. If the condition (abnormality in step 280, it takes an action against battery abnormality and finishes the flow.

[0076] FIG. 8 shows a balancing subroutine for switching the balancing switches 29A, 29B, 29C, 29D. That is, this balancing subroutine is a processing flow for discharging the series-connected battery cells 2A, 2B, 2C, 2D constituting the unit battery cell 2 and matching the battery cell voltage of the four battery cells 2A, 2B, 2C, 2D constituting the unit battery cell with each other.

[0077] In step 400 in FIG. 8, the main controller 5 sends each battery cell voltage reading command, which is a command for reading the voltage data of each battery cell 2A to 2D, to the RX terminal of the communication means 27 in FIG. 2. When the each battery cell voltage reading command is sent in step 400, the each battery cell voltage reading command judges control particulars and reads the battery cell voltage of each battery cell 2A, 2B, 2C, 2D, periodically updated and stored in the storage means, in the calculation means 23 of the control IC chip 3A and sends in series each battery cell voltage data from the TX terminal to the main controller 5. When each battery cell voltage data from the control IC chip 3A is received, the main controller 5 finds the minimum battery cell voltage out of each battery cell voltage data received and calculates the minimum cell voltage so as to calculate the discharging time of each battery cell in step 420. After calculating the minimum cell voltage in step 420, it calculates the ON time of each balancing switch 29A, 29B, 29C, 29D in step 430. The ON time of each balancing switch 29A, 29B, 29C, 29D is calculated by subtracting the minimum cell voltage from each battery cell voltage.

[0078] In step 440, a bypass SW control (ON) command for ON control of each balancing switch 29A, 29B, 29C, 29D is sent from the main controller 5 to the RX terminal of the communication means 27 shown in FIG. 2. When the bypass SW control (ON) command is sent in step 440, the bypass control (ON) command judges control particulars in the calculation means 23 of the control IC chip 3A and drives the SW drive means 33 so that a switch drive signal (a signal specifying which switch to drive) is outputted from the SW drive means 33 to the potential conversion means 32, and a selected balancing switch out of 29A, 2529B, 29C, 29D is turned ON in step 450. When the selected balancing switch out of 29A, 29B, 29C, 29D is turned ON, one of the battery cells 2A, 2B, 2C, 2D discharges.

[0079] When the selected balancing switch out of 29A, 29B, 29C, 29D is turned ON in step 450, the main controller 5 counts the ON elapsed time of each bypass SW (balancing switch) 29A, 29B, 29C, 29D in step 460. When the ON elapsed time of each bypass SW is counted in step 460, whether the ON elapsed time of each bypass SW (balancing switch) 29A, 29B, 29C, 29D becomes greater than the ON time in step 470. That is, in step 470, the main controller

waits until the ON elapsed time of each bypass SW (balancing switch) **29A**, **29B**, **29C**, **29D** becomes greater than the ON time.

[0080] When the ON elapsed time of each bypass SW (balancing switch) 29A, 29B, 29C, 29D is judged greater than the ON time in step 470, a bypass SW control (OFF) command for OFF control of each balancing switch 29A, 29B, 29C, 29D is sent from the main controller 5 to the RX terminal of the communication means 27 shown in FIG. 2 in step 480. When the bypass SW control (OFF) command is sent in step 480, the bypass control (OFF) command judges control particulars in the calculation means 23 of the control IC chip 3A and controls the SW drive means 33 so that a switch drive signal (a signal specifying which switch to drive) is outputted from the SW drive means 33 to the potential conversion means 32, and a selected balancing switch out of 29A, 29B, 29C, 29D is turned OFF in step 490. When the selected balancing switch out of 29A, 29B, 29C, 29D is turned OFF, one of the battery cells 2A, 2B, 2C, 2D stops discharging. A similar operation applies to the control IC chip 4A and control IC chip 5A.

[0081] FIG. 9 shows an operation flow for checking whether the control IC chips 3A to 5A or each battery cell is normal or not.

[0082] To start with, in step 500, a condition (abnormality) sensing command (the first signal) is sent from the TX terminal of the main controller 5 to the RX terminal of the control IC chip 3A. When a condition (abnormality) sensing command is sent from the TX terminal of the main controller 5, the control IC chip 3A receives the condition (abnormality) sensing ity) sensing command.

[0083] When the condition (abnormality) sensing command is sent from the TX terminal of the main controller 5 in step 500, the control IC chip 3A, control IC chip 4A, . . . . control IC chip 5A receives it in turn and the cell monitor IC chip 5B on the last stage sends it to the main controller 5.

[0084] That is, the control IC chip 3A that has received the condition (abnormality) sensing command adds an abnormality signal representing the existence or the absence of abnormality in own range, and sends the condition (abnormality) sensing command to the RX terminal of the next control IC chip 4A. When the condition (abnormality) sensing command is outputted from the TX terminal of the control IC chip 3A, the control IC chip 4A receives the condition (abnormality) sensing command and sends the condition (abnormality) sensing command to the TX terminal of the next control IC chip 5A. Consequently, when the control IC chip 5A on the last stage receives the condition (abnormality) sensing command sent from the TX terminal of the control IC chip 4A, it sends the condition (abnormality) sensing command received through the TX terminal of the control IC chip 5A to the RX terminal of the main controller 5 via the insulation means 10.

[0085] When the control IC chip 3A, control IC chip 4A, ... control IC chip 5A receives the command in turn and the cell monitor IC chip 5B on the last stage sends it to the main controller 5 in step 510, the main controller 5 that has received the condition (abnormality) sensing command from the control IC chip 5A checks the condition (abnormality) in step 520. Which of the control IC chip 3A, control IC chip

 $4A, \ldots$  control IC chip 5A or corresponding battery cells is abnormal can be judged from the condition (abnormality) sensing command returned to the main controller 5.

[0086] After checking the condition (abnormality) of the control IC chip 3A, control IC chip 4A, ... control IC chip 5A in step 520, the main controller 5 judges whether abnormality is found on any of the control IC chips or corresponding battery cells in step 530. If it judges no abnormality is found on any of the control IC chips or corresponding battery cells in step 530, it finishes the flow. If the main controller 5 judges abnormality is found on any of the control IC chips or corresponding battery cells in step 530, it finishes the flow. If the main controller 5 judges abnormality is found on any of the control IC chip 5A in step 530, a condition (abnormality detail) sensing command (the second signal) for specifying the address of the control IC chip on which abnormality is sensed and identifying the abnormality detail is sent from the TX terminal of the main controller 5 to the RX terminal of the control IC chip 3A via the insulation means 7 in step 540.

[0087] When the condition (abnormality detail) sensing command is sent from the TX terminal of the main controller 5 in step 540, the control IC chip 3A receives it in step 550 and then a control IC chip having different address than the specified sends the condition (abnormality detail) sensing command as it is to the control IC chip on the next stage. This sending and receiving is performed sequentially as follows: the control IC .chip 3A receives the condition (abnormality detail) sensing command through the RX terminal and sends it from the TX terminal to the RX terminal of the control IC chip 4A, and the control IC chip 4A sends it from the TX terminal to the RX terminal of the control IC chip 5A, and the control IC chip 5A sends the condition (abnormality detail) sensing command received from the control IC chip 4A from the TX terminal of the control IC chip 5A to the RX terminal of the main controller 5 via the insulation means 10.

[0088] When abnormality is sensed based on the condition (abnormality detail) sensing command that is received from the control IC chip 4A and sent from the TX terminal of the control IC chip 5A to the RX terminal of the main controller 5 via the insulation means 9, a signal is outputted from the Relay terminal of the main controller 5 so as to drive a relay drive circuit and turn OFF the relay.

[0089] When the control IC chip 3A, control IC chip 4A, ... control IC chip 5A receives the command in turn and the cell monitor IC chip 5A on the last stage sends it to the main controller 5 in step 550, the main controller 5 having received the condition (abnormality detail) sensing command from the control IC chip 5A checks the abnormal portion and abnormality detail in the control IC chip 3A, control IC chip 4A, ... control IC chip 5A in step 560 and finishes the flow.

[0090] The main controller 5 first sends an alert signal for sending a signal (break field) from the TX terminal of the main controller 5 to the RX terminal of the control IC chip 3A and then sends a synchronous signal for receiving an incoming signal synchronously, and after that, sends out the first signal for sensing abnormality continuously. Responding to the first signal for sensing abnormality, an abnormality sensed signal showing abnormality is sensed on one of the control IC chip 3A, control IC chip 4A, . . . control IC chip 5A or corresponding battery cells is sent back with an identified address of abnormal control IC chip. When this

abnormality sensed signal which represents the existence or the absence of abnormality of the control IC chip or the battery cells is received, the main controller **5** sends out the second signal for identifying the abnormality detail based on the abnormal sensed signal. This signal for identifying the abnormality detail specifies which control IC chip shall send what type of information, and the abnormality detail specifies the address and type of data (overcharging, battery cell voltage, etc.).

**[0091]** As explained above, the main controller **5** collects individual voltage of the battery cells and performs cell balancing control upon start-up, and then sends a signal for sensing abnormality of each control IC chip and, if abnormality is sensed, sends a signal for identifying the abnormality detail.

[0092] In this embodiment, a multi-series battery control system comprises a plurality of unit battery cells (2) of which unit consists of multiple battery cells (2A-2D) connected in series; a plurality of control IC chips (3A-5A) comprising a control circuit for controlling the unit battery cell (2); a plurality of cell monitor IC chips (3B-5B) each monitoring the voltage of the unit battery cell (2); a plurality of control ICs (3) each consisting of the control IC chip (3A-5A) and the cell monitor IC chip (3B-5B); a main controller (5) that sends and receives signal to/from the control IC chips (3A-5A) via an insulation (4,8); means (process in 510) for sending an abnormality signal, which represents the existence or the absence of abnormality of the control IC chips or the battery cells, to the main controller (5) from the control IC chips (3A-5A), responding to the first signal (abnormality sensing command) outputted from the main controller (5) via the insulation (4,8); means (process in 540) for sending the abnormality contents signal of the control IC chip or the battery cells, to the main controller (5) from the control IC chips (3A-5A), responding to the second signal (condition sensing command) outputted from the main controller (5) via the insulation (4,8); and means (process in 410) for sending voltage signals of the battery cells, to the main controller (5) from the control IC chips (3A-5A), responding to a voltage sensing command outputted from the main controller (5) via the insulation (4,8).

[0093] The main controller 5 periodically senses the total voltage of the battery cells by the voltage sensing means and collects it through the VALL terminal of the main controller 5 via the insulation means. It also senses the total current through the battery cells by the current sensing means and collects it through the CUR terminal of the main controller 5. In addition, the main controller 5 periodically sums up each cell voltage and compares the total voltage so as to accomplish conformity diagnosis by checking if the differential voltage is within a specified range. Since whether this differential voltage is within a specified range or not is always checked, nothing more is needed to adjust the balancing but turning ON/OFF the balancing switches according to the voltage of each battery cell.

**[0094] FIG. 11** shows a case where this embodiment is employed in combination with a commercial power source.

[0095] In the figure, 1201 is a commercial power source, 1202 is a solar-power generation system, 1203 is a load device, 1204 is a control converter, and 1205 is a switch.

[0096] Multiple battery cells 101 are connected in series, a battery management IC is connected with each battery cell

101, and the output of the battery management IC is connected with the main controller 5 via an insulation coupler. In addition, the control converter 1204 is connected to both ends of the row of the battery cells 101, and the main controller 5 is connected with the MCU in the control converter 1204.

[0097] Furthermore, the solar-power generation system 1202, load device 1203 and control converter 1204 are connected with the common commercial power source 1201 each via a switch 1205. At the same time, the solar-power generation system 1202, load device 1203, control converter 1204, switch 1205 and main controller 5 are connected with each other in both directions.

**[0098]** The solar-power generation system **1202** is a system that converts the sunlight to DC current using solar cells and outputs AC current using an inverter.

[0099] The load device 1203 includes home electric appliances such as air-conditioner, refrigerator, microwave range, and lighting, and electric appliances such as motor, computer, and medical devices. The control converter 1204 is a charging/discharging device that converts AC current to DC current or DC current to AC current. This converter also functions as a controller for controlling the charging and discharging as well as for controlling the above solar-power generation system 1202 and load device 1203.

**[0100]** In the construction as above, if power needed for the load device **1203** cannot be fully supplied by the commercial power source **1201** and solar-power generation system **1202**, power is supplied from the battery cell **101** via the control converter **1204**. When the power supplied from the commercial power source **1201** and solar-power generation system **1202** becomes excessive, it is stored in the battery cell **101** via the control converter **1204**.

[0101] If the terminal-to-terminal voltage of the battery cell 101 reaches a level requiring charging or discharging to be ceased in the course of the above operation, the main controller 5 sends a relevant signal to the control converter 1204 and the control converter 1204 controls charging and discharging accordingly.

**[0102]** With the above construction, it becomes possible to lower the contract demand and power demand of the commercial power source **1201** and generation rating of the solar-power generation system **1202**, and hence equipment cost and running cost decrease.

**[0103]** In addition, if power is supplied from the battery cell **101** to the commercial power source **1201** when the power demand concentrates to a specific time zone and stored into a storage battery when the power demand is low, the concentration of power demand can be moderated and the power demand can be leveled

**[0104]** Furthermore, since the control converter **1204** monitors the power demand of the load device **1203** and controls the load device **1203** accordingly, energy saving and effective utilization of power can be realized.

**[0105] FIG. 12** shows a case where the embodiment is applied to a motor generator.

[0106] If the figure, 1101 is a motor generator, 1004 is a control converter, 1005 is a voltage regulator, and 1102 is a

DC load device (for example, power steering, electric brake, and suction/exhaust valve timing device).

[0107] Multiple battery cells 101 are connected in series, a battery management IC is connected with each battery cell 101, and the output of the battery management IC is connected with the main controller 5 via an insulation coupler. In addition, the main controller 5 is connected with the MCU in the control converter 1004.

**[0108]** The motor generator **1101** is a motor that converts the generated AC power to DC power and outputs.

[0109] With the above construction, while an automobile is driven by engine and is moving, power is generated by the motor generator 1101 that is driven by the automobile movement via a drive belt or directly driven by actuating an electromagnetic clutch. The power generated by the motor generator 1101 is supplied and charged into the battery cell 101 via the control converter 1004. Charging and discharging of the battery cell 101 is controlled by the motor generator 1101 through the battery management IC and via the main controller 5. In case of discharging, power is supplied through the battery management IC to the motor to drive the tires to rotate. The MCU in the control converter 1004 and the system are also connected with each other.

[0110] The main controller 5 is grounded with the ground (chassis ground) but both ends of the battery cells 2A, 2B, 2C, 2D of the unit battery cell 2 are lifted from the ground. The control converter 1004 is not grounded, either but lifted from the ground. In short, the power related circuit is lifted from the ground.

**[0111]** If the system is actually abnormal while the main controller **5** becomes out of control and mistakenly judges normal, the relay cannot be turned off because the main controller **5** is out of control. If this happens, a signal is outputted from an analog system so as to drive the relay drive circuit and turn OFF the relay.

**[0112]** With this embodiment, the number of components constituting the multi-series battery control system can be decreased.

**[0113]** In addition, with this embodiment, lower cost can be realized in constructing the multi-series battery control system.

**[0114]** Furthermore, with this embodiment, higher reliability of the multi-series battery control system can be realized.

**[0115]** Furthermore, with this embodiment, operability of the multi-series battery control system can be improved.

**[0116]** Furthermore, with this embodiment, the multi-series battery control system can be further generalized.

**[0117]** With this embodiment, higher-speed communication in the multi-series battery control system can also be realized.

**[0118]** In addition, with this embodiment, the multi-serial battery control system can be easily constructed and can be simplified.

**[0119]** According to the proper embodiments of the present invention, high reliability can be achieved.

What is claimed is:

1. A multi-series battery control system comprising:

- a plurality of unit battery cells of which unit consists of multiple battery cells connected in series;
- a plurality of control ICs each comprising a control circuit for controlling the unit battery cell;
- a main controller that sends and receives signal to/from the control ICs via an insulation;
- abnormality sensing means for sending an abnormality signal, which represents the existence or the absence of abnormality of the control ICs or the battery cells, to the main controller from the control ICs, responding to an abnormality sensing command outputted from the main controller via the insulation; and
- voltage sensing means for sending voltage signals of the battery cells, to the main controller from the control ICs, responding to a voltage sensing command outputted from the main controller via the insulation.

**2**. A multi-series battery control system according to claim 1, wherein the control ICs each comprising:

a balancing circuit which discharges so that the voltage of each cell constituting the unit battery cell connected with the control IC becomes stable at a specified voltage.

**3**. A multi-series battery control system according to claim 1, wherein the control ICs each comprising:

- means for sensing the voltage of each cell constituting the unit battery cell connected with the control IC; and
- means for controlling the supply of power from the unit battery cell.

**4**. A multi-series battery control system according to claim 1, wherein the main controller is grounded with chassis but the unit battery cells and the control ICs are not grounded.

**5**. A multi-series battery control system according to claim 2, wherein the balancing circuit in the control IC is driven when the voltage variation of each cell constituting the unit battery cell sensed on the control IC is greater than a specified voltage.

6. A multi-series battery control system comprising:

- a plurality of unit battery cells of which unit consists of multiple battery cells connected in series;
- a plurality of control ICs each comprising a control circuit for controlling the unit battery cell;
- a main controller that sends and receives signal to/from the control ICs via an insulation;
- means for sending an abnormality signal, which represents the existence or the absence of abnormality of the control ICs or the battery cells, to the main controller from the control ICs, responding to the first signal outputted from the main controller via the insulation; and
- means for sending an abnormality contents signal of the control ICs or the battery cells, to the main controller from the control ICs, responding to the second signal outputted from the main controller via the insulation.

**7**. A multi-series battery control system according to claim 6, wherein the control ICs each comprising:

a balancing circuit which discharges so that the voltage of each cell constituting the unit battery cell connected with the control IC becomes stable at a specified voltage.

**8**. A multi-series battery control system according to claim 6, wherein the control ICs each comprising:

- means for sensing the voltage of each cell constituting the unit battery cell connected with the control IC; and
- means for controlling the supply of power from the unit battery cell.

**9**. A multi-series battery control system according to claim 6, wherein communication between the main controller and the control ICs are made duplex communication system so as to send the first signal for requesting abnormality of the control ICs or the battery cells and send the second signal for requesting abnormality contents to the control ICs in case the existence of abnormality is sensed by the abnormality requesting.

**10.** A multi-series battery control system according to claim 6, wherein the main controller is grounded with chassis but the unit battery cells and the control ICs are not grounded.

**11.** A multi-series battery control system according to claim 7, wherein the balancing circuit in the control IC is driven when the voltage variation of each cell constituting the unit battery cell sensed on the control IC is greater than a specified voltage.

12. A multi-series battery control system comprising:

- a plurality of unit battery cells of which unit consists of multiple battery cells connected in series;
- a plurality of control IC chips each comprising a control circuit for controlling the unit battery cell;
- a plurality of cell monitor IC chips each monitoring the voltage of the unit battery cell;
- a plurality of control ICs each consisting of the control IC chip and the cell monitor IC chip;
- a main controller that sends and receives signal to/from the control IC chips via an insulation;
- means for sending an abnormality signal, which represents the existence or the absence of abnormality of the control IC chips or the battery cells, to the main controller from the control IC chips, responding to the first signal outputted from the main controller via the insulation; and

means for sending the abnormality contents signal of the control IC chip or the battery cells, to the main controller from the control IC chips, responding to the second signal outputted from the main controller via the insulation.

**13**. A multi-series battery control system according to claim 12, wherein the cell monitor IC chips each comprising:

a balancing circuit which discharges so that the voltage of each cell constituting the unit battery cell connected with the control IC chip becomes stable at a specified voltage.

14. A multi-series battery control system according to claim 12, wherein the control IC chips each comprising:

- means for sensing the voltage of each cell constituting the unit battery cell connected with the control IC chip; and
- means for controlling the supply of power from the unit battery cell.

15. A multi-series battery control system according to claim 12, wherein communication between the main controller and the control IC chips are made duplex communication system so as to send the first signal for requesting abnormality of the control IC chips or the battery cells and send the second signal for requesting abnormality contents to the control IC chips in case the existence of abnormality is sensed by the abnormality requesting.

16. A multi-series battery control system according to claim 12, wherein the main controller is grounded with chassis but the unit battery cells, the control IC chips and the cell monitor IC chips are not grounded.

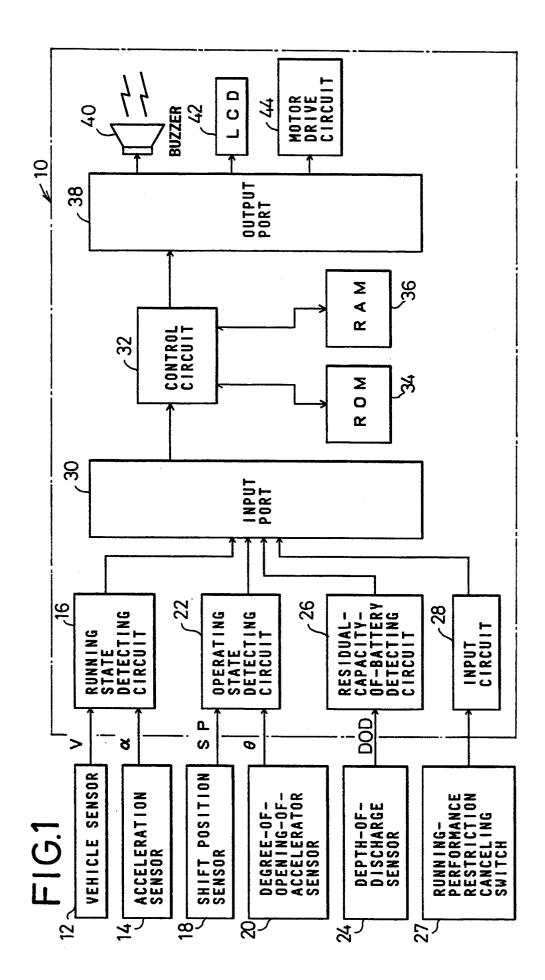
17. A multi-series battery control system according to claim 13, wherein the balancing circuit in the cell monitor IC chip is driven based on a command from the control IC chip when the voltage variation of each cell constituting the unit battery cell sensed on the control IC chip is greater than a specified voltage.

**18**. A multi-series battery control system according to claim 13, wherein the cell monitor IC chips each comprising:

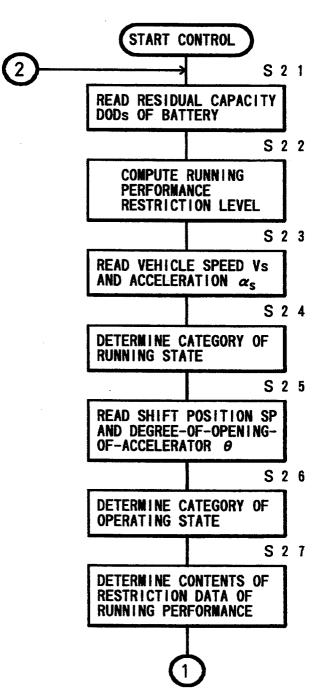
means for driving the balancing circuit independently even without receiving a command for driving the balancing circuit from the control IC chip when the voltage variation of each cell constituting the unit battery cell is greater than a specified voltage.

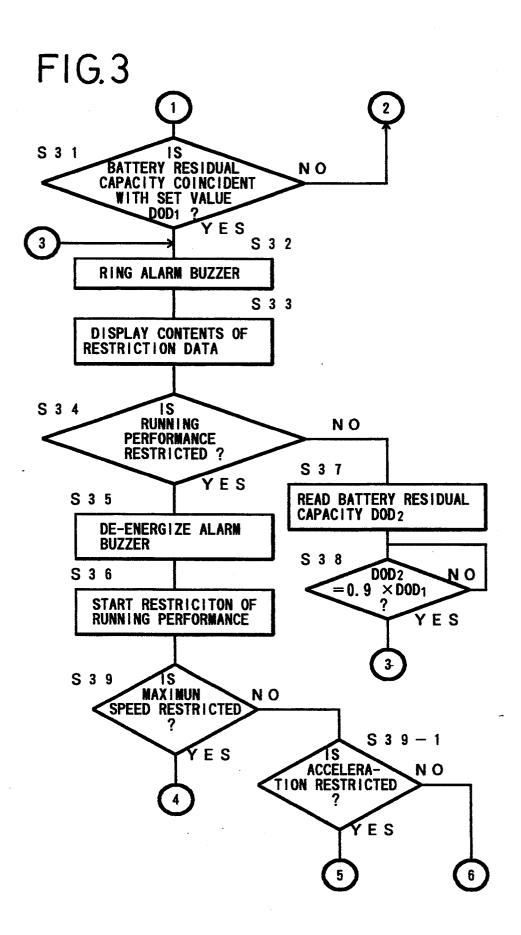
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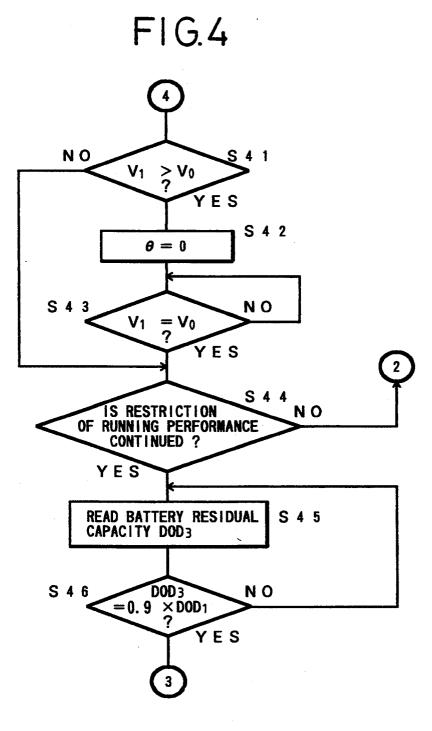
# IE4

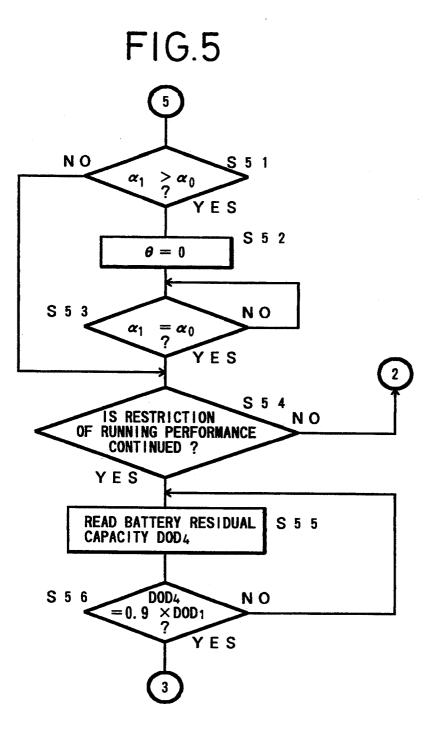


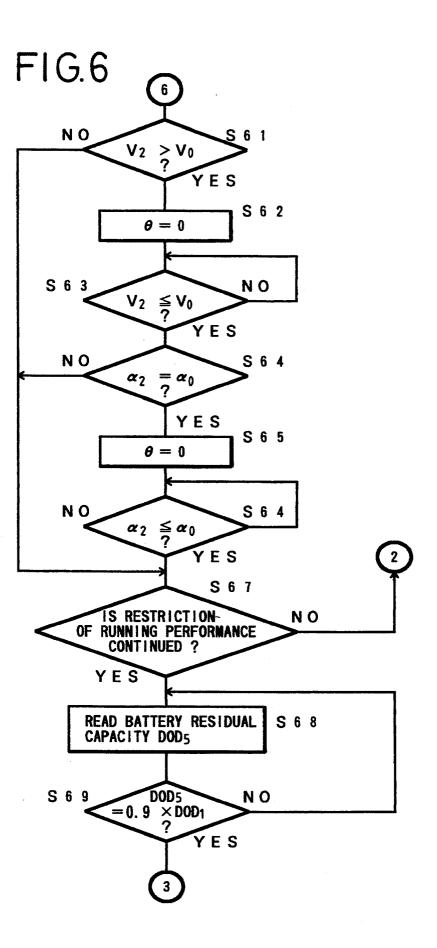


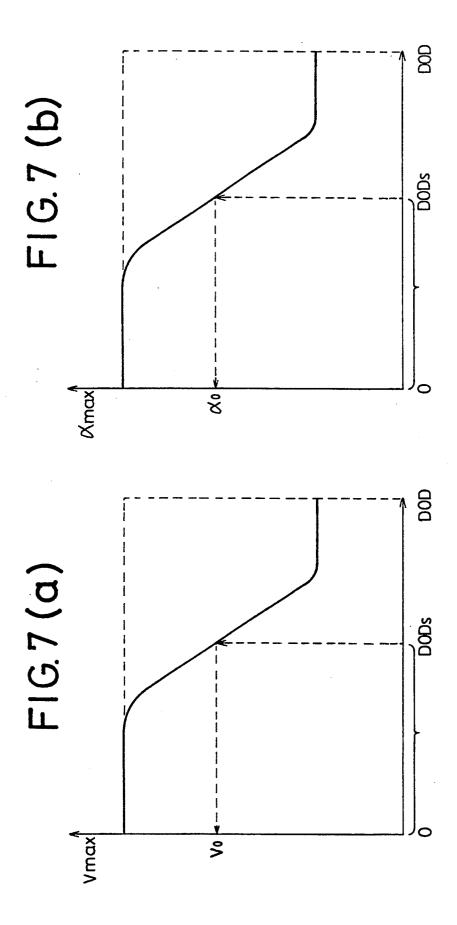


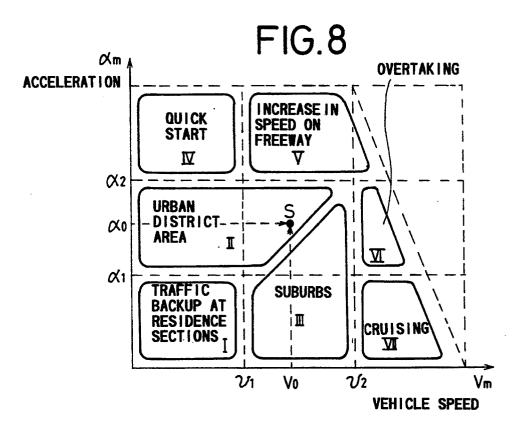


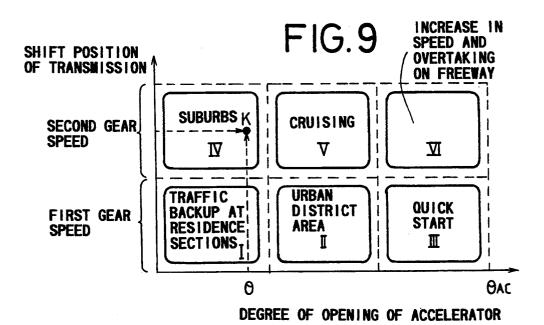












# FIG.10

				A : RESTRICTION OF MAXIMUM SPEED	B : RESTRICTION OF MAXIMUM ACCELERATION	C : RESTRICTION OF BOTH MAXIMUM SPEED AND	MAXIMUM ACCELERATION	X : IMPROBABLE ASPECTS
OPERATING STATE OF DRIVEN VEHICLE	١١	×	×	1	×	1	1	ပ
	>	×	×	A	×	1	1	1
	N	×	A	A	×	1	t	1
	III	B	B	×	B	B	×	×
	II	1	1	ł	×	×	×	×
		l	1	l	×	×	×	×
		-	Ξ	Ħ	N	>	N	⋝
RUNNING STATE								

- : NON-RESTRICTION ASPECTS

### **RUNNING PERFORMANCE CONTROL** APPARATUS AND METHOD FOR AN ELECTRIC VEHICLE

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# BACKGROUND OF THE INVENTION

### 1. Field of the Invention

The present invention relates to a running performance control apparatus suitable for use in an electric vehicle, which is capable of determining, based on a 10 driver's choice, whether the running performance of the vehicle while being driven should be restricted when the capacity of a battery used for the running of the electric vehicle is reduced to a predetermined value.

2. Description of the Related Art

It has recently been pointed out that the environment is deteriorating due to noise and exhaust gases produced by gasoline vehicles. Therefore, a strong demand is now arising for wide use of electric vehicles as a countermea-20 sure for solving such environmental problems.

Since this type of electric vehicle is normally driven by electric energy stored in a battery, the vehicle is provided with a large battery. However, the continuous running distance over which tile electric vehicle can be driven by one charge level is shorter than that obtained <sup>25</sup> by a gasoline vehicle. Therefore, a method of efficiently using the capacity of a battery is an important consideration.

When a battery develops trouble and electric power outputted from the faulty battery is reduced, for exam- 30 ple, an electric vehicle is apt to abruptly stop running before it reaches a desired destination when the electric vehicle runs under the action of the electric power outputted from the battery which is at fault.

In order to solve such a problem, a technique for 35 comparing a normal battery voltage with an actual battery voltage in view of improper variations in the battery voltage due to a failure in a valve, and for determining a torque command value for a motor based on the result of the comparison so as to reduce the torque 40 outputted from the motor, thereby making it possible to increase a vehicle drivable distance under the action of the defective battery, has been disclosed in Japanese Patent Application Laid-Open Publication No. 63-77302, for example.

However, the method which is employed in the above prior art has the problem that the running state is not controlled to provide the running state corresponding to the running environment at all times because whether the running environment of a running vehicle 50 falls into either an express highway or an urban district area has not been taken into consideration.

# SUMMARY OF THE INVENTION

vide a running performance control apparatus suitable for use in an electric vehicle, which is capable of restricting running performance to the optimum level based on a running state immediately preceding the present running state when the capacity of the battery is 60 reduced to the preset residual capacity during running of the vehicle and the restriction of the running performance begins, and determining, based on a driver's choice, whether the restriction of the running performance should begin, thereby making it possible to re- 65 output of the depth-of-discharge sensor into data on the strict the running state to the most suitable level.

It is a principal object of the present invention to provide a running performance control apparatus suit-

able for use in an electric vehicle, the running performance control apparatus being used to restrict the running performance of the vehicle when the capacity of a battery employed in the vehicle is reduced to the preset residual capacity, the running performance control apparatus comprising detecting means for detecting that the battery capacity is reduced to the preset residual capacity, running state detecting means for detecting the vehicle running state including the speed of the running vehicle and the acceleration of the vehicle. operating state detecting means for detecting an operating state including shift positions of the transmission of the vehicle and the degree of opening of the vehicle accelerator, storing means for storing therein the con-15 tents of a plurality of restriction data each used to restrict the running performance of the vehicle, determining means for selectively determining whether the running state of the vehicle should be restricted when the battery capacity is reduced to the preset residual capacity, and running performance controlling means for reading information indicative of the contents of one restriction datum for restricting the running performance from the storing means based on information indicative of the running state detected by the running state detecting means and information indicative of the operating state detected by the operating state detecting means when the determining means determines that the running state of the vehicle should be restricted, thereby restricting the running performance of the vehicle.

It is another object of the present invention to provide a running performance control apparatus suitable for use in an electric vehicle, the running performance control apparatus being used to restrict the running performance of the vehicle when the capacity of a battery employed in the vehicle is reduced to the preset residual capacity, the running performance control apparatus comprising detecting means for detecting that the battery capacity is reduced to the preset residual capacity, running state detecting means for detecting the vehicle running state including the speed of a running vehicle and the acceleration of the vehicle, operating state detecting means for detecting an operating 45 state including shift positions of the transmission of the vehicle and the degree of opening of the vehicle accelerator, storing means for storing therein the contents of a plurality of restriction data each used to restrict the running performance of the vehicle, and running performance controlling means for reading the contents of one restriction datum for restricting the running performance from the storing means based on information indicative of the running state detected by the running state detecting means add information indicative of the It is a general object of the present invention to pro- 55 operating state detected by the operating state detecting means when the battery capacity detected by the detecting means is reduced to the preset residual capacity, thereby restricting the running performance of the vehicle.

> It is a further object of the present invention to provide a running performance control apparatus of this type wherein the detecting means comprises a depth-ofdischarge sensor for detecting the depth of discharge of the battery and a converter circuit for converting the residual capacity of the battery.

> It is a still further object of the present invention to provide a running performance control apparatus of this

type wherein the converter circuit can detect a plurality of residual capacities which differ from one another.

It is a still further object of the present invention to provide a running performance control apparatus of this type wherein the running state detecting means com- 5 prises a vehicle speed sensor for detecting the running speed of the vehicle and an acceleration sensor for detecting the acceleration of the running vehicle.

It is a still further object of the present invention to provide a running performance control apparatus of this 10 type wherein the operating state detecting means comprises a shift position sensor for detecting each of the shift positions of the transmission and a degree-of-opening-of-accelerator sensor for detecting the degree of opening of the accelerator. 15

It is a still further object of the present invention to provide a running performance control apparatus of this type wherein each of the contents stored in the storing means represents any one of an aspect of the restriction of the maximum speed of the vehicle, an aspect of the 20 restriction of the maximum acceleration of the vehicle, an aspect of the restriction of both the maximum speed and the maximum acceleration, and an aspect of the non-restriction of the maximum speed and the maximum acceleration. 25

It is a still further object of the present invention to provide a method of controlling a running performance of an electric vehicle having a battery and a transmission, when the capacity of the battery is reduced to a preset residual capacity, comprising the steps of detect- 30 ing that the capacity of the battery is reduced to the preset residual capacity, detecting a vehicle running state including a speed of the vehicle while being driven and an acceleration of the vehicle, detecting an operating state including shift positions of a transmission of the 35 vehicle and tile degree of opening of the vehicle accelerator, storing the contents of a plurality of restriction data each used to restrict the running performance of the vehicle, selectively determining whether the running state of the vehicle should be restricted when the 40 battery capacity is reduced to the preset residual capacity, reading information indicative of the stored contents of one restriction datum for restricting the running performance based on information indicative of the running state and reading information indicative of the 45 operating state when it has been selectively determined that the running state of the vehicle should be restricted, and restricting the running performance of the vehicle in response to the read information.

It is a still further object of the present invention to 50 provide a method of controlling the running performance of an electric vehicle having a battery and transmission, wherein the detecting of the capacity of the battery includes detecting the depth of discharge of the battery and converting the detected depth-of-discharge 55 into data on the residual capacity of the battery.

It is a still further object of the present invention to provide a method of controlling the running performance of an electric vehicle having a battery and transmission, wherein the converting step converts a plural-60 ity of residual capacities of the battery which differ from one another.

It is a still further object of the present invention to provide a method of controlling the running performance of an electric vehicle having a battery and trans- 65 mission, wherein each of the contents stored in the storing step represents any one of an aspect of the restriction of the maximum speed of the vehicle, an aspect

of the restriction of the maximum acceleration of the vehicle, an aspect of the restriction of both the maximum speed and the maximum acceleration, and an aspect of the non-restriction of the maximum speed and the maximum acceleration.

It is a still further object of the present invention to provide a running performance control apparatus suitable for use in an electric vehicle, the running performance control apparatus being used to restrict the running performance of the vehicle when the capacity of a battery employed in the vehicle is reduced to a preset residual capacity, the running performance control apparatus comprising means for determining that the battery is reduced to the preset residual capacity, means for determining a vehicle running state including a speed of the vehicle and an acceleration of the vehicle, means for storing restriction data used to restrict the running performance of the vehicle, and control means for reading one of the restriction data from the storing means based on information indicative of the running state determined by the running state determining means when the battery capacity determined by the battery capacity determining means is reduced to the preset residual capacity, and for restricting the running performance of the vehicle in response to the one restriction data.

It is a still further object of the present invention to provide a running performance control apparatus, wherein operating state determining means are provided for determining each of the shift positions of a transmission and a degree of opening of an accelerator of the vehicle, and the control means reading is also based on information determined by the operating state determining means.

It is a still further object of the present invention to provide a running performance control apparatus, wherein means are provided for a driver of the vehicle to selectively prevent the restriction on the running performance of the vehicle.

It is a still further object of the present invention to provide a running performance control apparatus, wherein the restriction data stored in the storing means represents any one of an aspect of the restriction of the maximum speed of the vehicle, an aspect of the restriction of the maximum acceleration of the vehicle, an aspect of the restriction of both the maximum speed and the maximum acceleration, and an aspect of the non-restriction of the maximum speed and the maximum acceleration.

It is a still further object of the present invention to provide a running performance control apparatus, wherein means are provided for indicating to a driver of the vehicle which aspect of the restriction on the running of the vehicle has been imposed by the control means.

It is a still further object of the present invention to provide a running performance control apparatus, wherein means are provided for the driver to selectively prevent the restriction on the running performance of the vehicle.

The above and other objects, features and advantages of the present invention will become apparent from the following description and the appended claims, taken in conjunction with the accompanying drawings in which a preferred embodiment of the present invention is shown by way of illustrative example.

# BRIEF DESCRIPTION OF THE DRAWINGS

FIG. 1 is a block diagram showing the structure of a running performance control apparatus according to one embodiment of the present invention;

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FIG. 2 is a flowchart for describing a procedure for restricting running performance of an electric vehicle while being driven, by the running performance control apparatus shown in FIG. 1;

FIG. 3 is a flowchart for describing another proce-10 dure for restricting the running performance of the vehicle by the running performance control apparatus;

FIG. 4 is a flowchart for describing a further procedure for restricting the running performance of the vehicle by the running performance control apparatus; 15

FIG. 5 is a flowchart for describing a still further procedure for restricting the running performance of the vehicle by the running performance control apparatus:

FIG. 6 is a flowchart for describing a still further 20 procedure for restricting the running performance of the vehicle by the running performance control apparatus

FIGS. 7(a) and 7(b) are graphs for describing a lookup table (LUT) used to read or determine the maximum 25 limit speed and the maximum limit acceleration from data indicative of the remaining capacity of the battery of the running vehicle by the running performance control apparatus:

FIG, 8 is a diagram for describing a LUT used to 30 determine a running state of the vehicle from data indicative of both the speed and the acceleration of the vehicle by the running performance control apparatus;

FIG. 9 is a diagram for describing a LUT used to determine an operating state of the running vehicle 35 from data indicative of the degree of opening of the accelerator and the shift positions of the transmission by the running performance control apparatus; and

FIG. 10 is a diagram for describing a LUT used to read data for restricting the running performance from 40 data indicative of the vehicle's running state determined in FIG. 8 and data indicative of the vehicle's operating state determined in FIG. 9.

### DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

FIG. 1 is a block diagram showing the structure of a running performance control apparatus according to one embodiment of the present invention.

In FIG. 1, reference numeral 10 indicates a running 50 performance control apparatus for an electric vehicle (not shown). The running performance control apparatus 10 comprises a running state detecting circuit 16 for reading or detecting signals output from a vehicle speed sensor 12 and an acceleration sensor 14, an operating 55 state detecting circuit 22 for detecting signals output from a shift position sensor 18 and a degree-of-openingof-accelerator sensor 20, a residual-capacity-of-battery detecting and converting circuit 26 for detecting a signal output from a depth-of-discharge (hereinafter called 60 "DOD") sensor 24 used to detect a discharged state of the main vehicle battery (not shown) and for converting the detected signal into data on the remaining capacity of the battery, an input circuit 28 for receiving a signal output from a running-performance restriction cancel- 65 restriction data A for restricting the running perforing switch 27 used to input information about a selective determination to be made by the driver of the vehicle as to whether the running performance of the electric

vehicle should be restricted, and an input port 30 used for these circuits.

Further, the running performance control apparatus 10 comprises a control circuit 32 for restricting the running performance of the vehicle, a ROM 34 for storing therein a program used to restrict the running performance of the vehicle, and a plurality of look-up tables (hereinafter called "LUTs"), a RAM 36 for temporarily storing therein data under the control of the control circuit 32, an output port 38 used for the control circuit 32, a buzzer 40 serving as a warning sound producing means, which is electrically connected to the output port 38, a liquid crystal display (LCD) 42 serving as a display device, and a motor drive circuit 44 for energizing an unillustrated motor used to run or drive the vehicle.

Operations and effects for restricting the running performance of the vehicle by the running performance control apparatus 10 constructed as described above will be described in detail with reference to FIGS. 1 through 10.

When the vehicle starts running, the control circuit 32 reads the remaining capacity DODs (DOD at start) of a battery from the residual-capacity-of-battery detecting and converting circuit 26 via the input port 30 (Step S21, see FIG. 2), and determines a running performance restriction level from the remaining capacity DODs.

More specifically, the maximum limit speed Vocorresponding to the residual capacity DODs of the battery is read from a LUT (see FIG. 7(a)) representing a relationship between the residual capacity DOD of the battery and the maximum speed Vmax previously stored in the ROM 34. The maximum limit speed  $V_0$ thus read is then stored in the RAM 36. In addition, the maximum limit acceleration  $\alpha_0$  corresponding to the residual capacity DODs is read from a LUT (see FIG. 7(b)) showing a relationship between the remaining capacity DOD of the battery and the maximum acceleration  $\alpha_{max}$ . The maximum limit acceleration a<sub>0</sub> thus read is then stored in the RAM 36 (Step S22).

Next, the control circuit 32 reads a vehicle speed Vs of a running vehicle detected by the vehicle speed sensor 12 and an acceleration  $\alpha_s$  of the running vehicle 45 detected by the acceleration sensor 14 via the running state detecting circuit 16 and the input port 30 (Step S23). Thereafter, one desired LUT previously stored in the ROM 34 is read based on the vehicle speed Vs and the acceleration  $\alpha_s$ , thereby determining that, for example, the vehicle is running in an urban district area (see a point S in FIG. 8) indicated by the number II (Step S24).

Further, the control circuit 32 reads a shift position SP of a lever for shifting the transmission (not shown), e.g., a second gear speed, and the degree of opening of an accelerator  $\theta$  from the operating state detecting circuit 22 via the input port 30 (Step S25). Then, the control circuit 32 reads a desired LUT based on both the shift position SP and the degree of the opening of the accelerator  $\theta$  and determines that, for example, the operating state of the running vehicle falls under the category of a suburb (see a point K in FIG. 9) indicated by the number IV (Step S26).

Next, the control circuit 32 determines or calculates mance of the vehicle from the LUT stored in the ROM 34 based on the number II as data indicative of the running state of the vehicle, which has been read in Step S24 and the number IV as data indicative of the operating state, which has been read in Step S26 (see FIG. 10). The restriction data A is used to restrict only the maximum speed of the vehicle without restricting the maximum acceleration of the vehicle. That is, the maximum 5 speed of the vehicle is restricted to the maximum limit speed V<sub>0</sub> determined in Step S22 (Step S27).

Thus, the restriction data A for restricting the running performance of the vehicle has been obtained. However, the calculation of the restriction data A is 10 carried out repeatedly until the remaining capacity DODs of the battery, which has been determined from the depth-of-discharge sensor 24 by the residual-capacity-of-battery detecting and converting circuit 26, is reduced to a set value DOD1 which has been previously 15 set in the ROM 34. The restriction data A is brought up to date for each calculation and stored in the RAM 36.

When the remaining capacity DODs detected by the residual-capacity-of-battery detecting and converting circuit 26 is reduced to the set value DOD<sub>1</sub> (Step S31, 20 see FIG. 3), the control circuit 32 is activated to energize the buzzer 40 via the output port 38, thereby producing a warning sound (Step S32). The contents of the latest restriction data read from the LUT shown in FIG. 10 and stored in the RAM 36 are displayed and blinked 25 on the LCD 42 (Step S33).

Next, the control circuit 32 determines whether the running-performance restriction canceling switch 27 has been operated by the driver (Step S34). When the running-performance restriction canceling switch 27 is 30 operated during a predetermined period, the control circuit 32 stops the buzzer 40 from ringing (Step S35) and starts restricting the running performance (Step S36)

If the answer is determined to be No in Step S34, then 35 the control circuit 32 is activated to cause the residualcapacity-of-battery detecting and converting circuit 26 to read the present remaining capacity DOD<sub>2</sub> from the depth-of-discharge sensor 24 (Step S37). When the read remaining capacity DOD<sub>2</sub> is reduced to a value corre- 40 sponding to 90% of the remaining capacity DOD<sub>1</sub> read in Step S31 (Step S38), the procedure is returned to Step S32 where the buzzer 40 is energized to ring.

On the other hand, the control circuit 32 determines whether the contents of the latest restriction data deter- 45 mined in Step S27 represent information indicative of the restriction of the maximum speed (Step S39). If the answer is determined to be Yes (see "A" in FIG. 10), then the control circuit 32 reads the present vehicle speed  $V_1$  from the running state detecting circuit 16 and 50 determines whether the present vehicle speed  $V_1$  is faster than the maximum limit speed  $V_0$  (Step S41, see FIG. 4). If the answer is determined to be Yes (i.e.,  $V_1 > V_0$ ), then the degree of opening of the accelerator  $\theta$  is set to "0" to restrain the speed of the vehicle (Step 55 Step S21 as the initial control step. S42). The running speed of the vehicle is restrained until the present vehicle speed  $V_1$  coincides with the maximum limit speed  $V_0$  (i.e.,  $V_1 = V_0$ ) (Step S43).

When the running speed  $V_1$  of the vehicle is restricted to the maximum limit speed  $V_0$ , the control 60 circuit 32 is activated to cause the input circuit 28 to read the state of the running-performance restriction canceling switch 27 again (Step S44). If the state of the running-performance restriction canceling switch 27 represents that the restriction of the running perfor- 65 mance continues, then the control circuit 32 reads the present remaining capacity DOD<sub>3</sub> of the battery (Step S45) and determines whether the present remaining

capacity DOD<sub>3</sub> is of the value corresponding to 90% of the remaining capacity DOD<sub>1</sub> read in Step S31 (Step S46). If the answer is determined to be Yes in Step S46, then the routine procedure is returned to Step S32 where the buzzer 40 is energized to buzz.

If the answer is determined to be No in Step S41, then the control circuit 32 determines in Step S44 whether the restriction of the running performance should be carried out continuously. If the state of the running-performance restriction canceling switch 27, which has been read by the control circuit 32 in Step S44, represents that the restriction of the running performance be discontinued, then the routine procedure is returned to Step S21 as the initial control step.

If, on the other hand, the result of the determination in Step S27 does not represent the restriction of the maximum speed, it is then determined whether the latest restriction data of the running performance, which has been determined in Step S27, represents data used to restrict an acceleration  $\alpha$  (Step S39-1). If the answer is determined to be Yes (see "B" in FIG. 10), then tile control circuit 32 reads the present acceleration  $\alpha_1$  from the running state detecting circuit 16 and determines whether the present acceleration  $\alpha_1$  is greater than the maximum limit acceleration  $\alpha_0$  (Step S51, see FIG. 5). If the answer is determined to be Yes (i.e.,  $\alpha_1 > \alpha_0$ ), then the degree of opening of the accelerator  $\theta$  is set to "0" to restrain the speed of the vehicle (Step S52). The running speed of the vehicle is restrained until the present acceleration  $\alpha_1$  coincides with the maximum limit acceleration  $\alpha_0$  (i.e.,  $\alpha_1 = \alpha_0$ ) (Step S53).

When the acceleration  $\alpha_1$  of the vehicle is restricted to the maximum limit acceleration  $\alpha_0$ , the control circuit 32 is activated to cause the input circuit 28 to read the state of the running-performance restriction canceling switch 27 again (Step S54). If the state of the running-performance restriction canceling switch 27 represents that the restriction of the running performance continues, then the control circuit 32 reads the present remaining capacity DOD4 of the battery (Step S55) and determines whether the read remaining capacity DOD<sub>4</sub> is of the value corresponding to 90% of the remaining capacity DOD<sub>1</sub> read in Step S31 (Step S56). If tile answer is determined to be Yes in Step S56, then the routine procedure is returned to Step S32 where the buzzer 40 is energized to ring.

If the answer is determined to be No in Step S51, then the control circuit 32 determines in Step S54 whether the restriction of the running performance should be carried out continuously. If the state of the running-performance restriction canceling switch 27, which has been read by the control circuit 32 in Step S54, represents that the restriction of the running performance be discontinued, then the routine procedure is returned to

If the result of the determination in Step S39-1 does not represent the restriction of the acceleration  $\alpha$ , it is then determined that the latest restriction data of the running performance, which has been determined in Step S27, represents data used to restrict the maximum speed and the maximum acceleration (see "C" in FIG. 10).

The present vehicle speed  $V_2$  is reduced to the maximum limit speed  $V_0$  in a manner similar to the process executed in each of Steps S41 through S43 (Steps S61 through S63, see FIG. 6). Next, the present acceleration  $\alpha_2$  of the vehicle is brought into agreement with the maximum limit acceleration  $\alpha_0$  in a manner similar to

the process executed in each of Steps S51 through S53 (Steps S64 through S66). It is then determined, based on a driver's choice, whether these restrictions should be carried out continuously (Step S67). If the answer is determined to be Yes in Step S67, then the present 5 remaining capacity DOD<sub>5</sub> of the battery is read (Step S68). When the read remaining capacity DOD<sub>5</sub> coincides with the value corresponding to 90% of the remaining capacity DOD<sub>1</sub> read in Step S31 (Step S69), the routine procedure is returned to Step S32 where the 10 buzzer 40 is energized to produce a buzzing sound.

According to the present embodiment, as described above, when the remaining capacity DODs of the battery of the vehicle while being driven is reduced to the preset remaining capacity DOD<sub>1</sub> (Step S31), the buzzer <sup>15</sup> 40 is activated to ring (Step S32) and the contents of the restriction data are displayed on the LCD 42 (Step S33). Then, the driver is urged to make a decision as to whether the restriction of the running performance should be carried out (Step S34). If the answer is determined to be Yes in Step S34, then the restriction of the running performance is made.

Further, when the running performance of the vehicle is limited to a target restriction state, it is determined, based on a driver's choice, whether the restriction of the running performance should be carried out continuously. Thus, the driver's circumstantial judgments made depending on running environments can be adopted for control without automatically and uniformly restricting the running state according to the residual capacity of the battery, thereby making it possible to provide the restriction of the running performance most suitable for the running state.

In the running performance control apparatus according to the present invention, which is used in the electric vehicle, the running performance of the vehicle while being driven is restricted based on the contents of the restriction data of the running performance, which are obtained from information indicative of the running state of the vehicle and information indicative of the operating state. It is therefore possible to provide the restriction of the most suitable running performance based on the running state just before the present running state. 45

Further, when the running performance of the vehicle is restricted, information can be read indicative of a decision to be made as to whether the running performance, input via the inputting means by the driver, should be restricted, and a selective decision can be  $_{50}$ made in accordance with the read information as to whether the restriction of the running performance should be carried out. Therefore, the restriction of the optimum running performance can be made, to which a consideration as to the running environments has been  $_{55}$ given.

Having now fully described the invention., it will be apparent to those skilled in the art that many changes and modifications can be made without departing from the spirit or scope of the invention as set forth herein. 60

What is claimed is:

1. A running performance control apparatus for use in an electric vehicle, wherein said running performance control apparatus restricts the running performance of the vehicle when a capacity of a battery em- 65 ployed in the vehicle is reduced to a preset residual capacity, said running performance control apparatus comprising:

- battery capacity detecting means for detecting that said battery capacity is reduced to said preset residual capacity;
- speed detecting means for detecting a speed of the vehicle while being driven;
- acceleration detecting means for detecting an acceleration of the vehicle;
- running state determining means for determining a vehicle running state based on detected values from said speed detecting means and said acceleration detecting means;
- shift position detecting means for detecting a shift position of a transmission of the vehicle;
- accelerator opening detecting means for detecting a degree of opening of an accelerator of the vehicle;
- operating state determining means for determining an operating state based on detected values from said shift position detecting means and said accelerator opening detecting means;
- storing means for storing therein a plurality of restriction data each used to restrict the running performance of the vehicle;
- restriction determining means for selectively determining whether the running state of the vehicle should be restricted when said battery capacity is reduced to said preset residual capacity; and
- running performance controlling means for reading information indicative of at least one restriction datum for restricting the running performance from said storing means based on information indicative of the running state determined by said running state determining means and on information indicative of the operating state determined by said operating state determining means when said restriction determining means determines that the running state of the vehicle should be restricted, and for restricting the running performance of the vehicle in response to said read information.

2. A running performance control apparatus according to claim 1, wherein said battery capacity detecting means comprises a depth-of-discharge sensor for detecting the depth of discharge of said battery and a converter circuit for converting the output of said depth-ofdischarge sensor into data on the residual capacity of the battery.

3. A running performance control apparatus according to claim 2, wherein said converter circuit can detect a plurality of residual capacities of said battery which differ from one another.

4. A running performance control apparatus according to claim 1, 2, or 3, wherein said running state determining means comprises said speed detecting means and said acceleration detecting means.

5. A running performance control apparatus according to claim 1, 2, or 3, wherein said operating state determining means comprises said shift position detecting means and said accelerator opening detecting means.

6. A running performance control apparatus according to claim 1, 2, or 3, wherein each of said plurality of restriction data stored in said storing means represents any one of an aspect of the restriction of the maximum speed of the vehicle, an aspect of the restriction of the maximum acceleration of the vehicle, an aspect of the restriction of both the maximum speed and the maximum acceleration, and an aspect of the non-restriction of the maximum speed and the maximum acceleration.

7. A running performance control apparatus for use in an electric vehicle, wherein said running performance control apparatus restricts the running performance of the vehicle when capacity of a battery employed in the vehicle is reduced to preset residual ca- 5 pacity, said running performance control apparatus comprising:

battery capacity detecting means for detecting that said battery capacity is reduced to said preset residual capacity:

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- speed detecting means for detecting a speed of the vehicle while being driven;
- acceleration detecting means for detecting an acceleration of the vehicle;
- running state determining means for determining a 15 vehicle running state based on the detected values from said speed detecting means and said acceleration detecting means;
- shift position detecting means for detecting a shift position of a transmission of the vehicle; 20
- accelerator opening detecting means for detecting a degree of opening of an accelerator of the vehicle;
- operating state determining means for determining an operating state based on the detected values from said shift position detecting means and said acceler- 25 ator opening detecting means;
- storing means for storing therein a plurality of restriction data each used to restrict the running performance of the vehicle; and
- running performance controlling means for reading 30 at least one restriction datum for restricting the running performance from said storing means based on information indicative of the running state determined by said running state determining means and on information indicative of the operat- 35 ing state determined by said operating state determining means when said battery capacity detected by said battery capacity detecting means is reduced to said preset residual capacity, and for restricting the running performance of the vehicle in response 40 to said at least one restriction datum.

8. A running performance control apparatus according to claim 7, wherein said battery capacity detecting means comprises a depth-of-discharge sensor for detecting the depth of discharge of the battery and a con- 45 verter circuit for converting the output of said depth-ofdischarge sensor into data on the residual capacity of the battery.

9. A running performance control apparatus according to claim 8, wherein said converter circuit can detect 50 a plurality of residual capacities of said battery which differ from one another.

10. A running performance control apparatus according to claim 7, 8, or 9, wherein said running state determining means comprises said speed detecting means and 55 mance control apparatus restricts the running perforsaid acceleration detecting means.

11. A running performance control apparatus according to claim 7, 8, or 9, wherein said operating state determining means comprises said shift position detecting means and said accelerator opening detecting 60 means.

12. A running performance control apparatus according to claim 7, 8, or 9, wherein each of said plurality of restriction data stored in said storing means represents any one of an aspect of the restriction of the maximum 65 speed of the vehicle, an aspect of the restriction of the maximum acceleration of the vehicle, an aspect of the restriction of both the maximum speed and the maxi-

mum acceleration, and an aspect of the non-restriction of the maximum speed and the maximum acceleration.

13. A method of controlling the running performance of an electric vehicle having a battery and a transmission, when the capacity of the battery is reduced to a preset residual capacity, comprising the steps of:

detecting that the capacity of said battery is reduced to said preset residual capacity;

detecting a speed of the vehicle while being driven; detecting an acceleration of the vehicle;

- determining a vehicle running state based on the detected speed of the vehicle while being driven and the detected acceleration of the vehicle;
- detecting a shift position of a transmission of the vehicle:
- detecting a degree of opening of an accelerator of the vehicle;
- determining an operating state of the vehicle based on the detected shift position and the detected degree of accelerator;
- storing a plurality of restriction data each used to restrict the running performance of the vehicle;
- selectively determining whether the running state of the vehicle should be restricted when said battery capacity is reduced to said preset residual capacity;
- reading information indicative of said stored restriction data for restricting the running performance based on information indicative of the running state and on information indicative of the operating state when it has been selectively determined that the running state of the vehicle should be restricted; and

restricting the running performance of the vehicle in response to the read information.

14. A method according to claim 13, wherein said detecting of the capacity of the battery includes detecting the depth of discharge of said battery and converting said detected depth-of-discharge into data on the residual capacity of the battery.

15. A method according to claim 14, wherein said converting step converts a plurality of residual capacities of said battery which differ from one another.

16. A method according to claim 13, 14, or 15, wherein each of said plurality of restriction data stored in said storing step represents any one of an aspect of the restriction of the maximum speed of the vehicle, an aspect of the restriction of the maximum acceleration of the vehicle, an aspect of the restriction of both the maximum speed and the maximum acceleration, and an aspect of the non-restriction of the maximum speed and the maximum acceleration.

17. A running performance control apparatus for use in an electric vehicle, wherein said running performance of the vehicle when capacity of a battery employed in the vehicle is reduced to a preset residual capacity, said running performance control apparatus comprising:

- means for determining that said battery is reduced to said preset residual capacity;
- means for detecting a speed of the vehicle while being driven:

means for detecting an acceleration of the vehicle;

running state determining means for determining a vehicle running state based on the detected speed of the vehicle and the detected acceleration of the vehicle;

means for detecting a shift position of a transmission of the vehicles;

means for detecting a degree of opening of an accelerator of the vehicle;

operating state determining means for determining an 5 operating state of the vehicle based on the detected shift position and the detected degree of opening of the accelerator;

means for storing restriction data used to restrict the running performance of the vehicle; and

control means for reading said restriction data from said storing means based on information indicative of the running state determined by said running state determining means and information indicative of the operating state determined by said operating state determining means when said battery capacity determined by said battery capacity determining means is reduced to said preset residual capacity, and for restricting the running performance of the vehicle in response to said restriction data.

18. A running performance control apparatus according to claim 17, wherein means are provided for a driver of the vehicle to selectively prevent the restriction on the running performance of the vehicle.

19. A running performance control apparatus according to claim 17 wherein said restriction data stored in said storing means represents any one of an aspect of the restriction of the maximum speed of the vehicle, an aspect of the restriction of the maximum acceleration of the vehicle, an aspect of the restriction of both the maximum speed and the maximum acceleration, and an aspect of the non-restriction of the maximum speed and the maximum acceleration.

of the running state determined by said running state determining means and information indicative of the operating state determined by said operating state determining means when said battery capacity determined by said battery capacity determining

21. A running performance control apparatus according to claim 20, wherein means are provided for the20 driver to selectively prevent the restriction on the running performance of the vehicle.

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